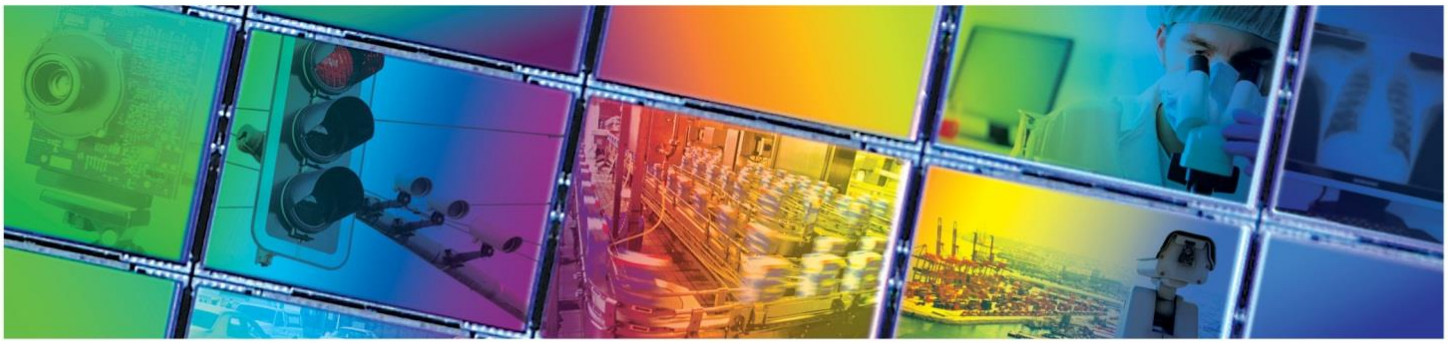
ON Semiconductor®

KAI-16070 IMAGE SENSOR

4864 (H) X 3232 (V) INTERLINE CCD IMAGE SENSOR



JUNE 4, 2014

DEVICE PERFORMANCE SPECIFICATION

REVISION 2.1 PS-0010

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KAI-16070 Image Sensor

DESCRIPTION

The KAI-16070 Image Sensor is a 16-megapixel CCD in a 35 mm optical format. Based on the TRUESENSE 7.4 micron Interline Transfer CCD Platform, the sensor provides very high smear rejection and up to 82 dB linear dynamic range through the use of a unique dual-gain amplifier. Flexible readout architecture enables use of 1, 2, or 4 outputs for full resolution readout up to 8 frames per second, while a vertical overflow drain structure suppresses image blooming and enables electronic shuttering for precise exposure control.

The sensor is available with the TRUESENSE Sparse Color Filter Pattern, a technology which provides a 2x improvement in light sensitivity compared to a standard color Bayer part.

The sensor shares common pin-out and electrical configurations with a full family of Truesense Imaging Interline Transfer CCD image sensors, allowing a single camera design to be leveraged in support of multiple devices.

FEATURES

- Superior smear rejection
- Up to 82 dB linear dynamic range
- Bayer Color Pattern, TRUESENSE Sparse Color Filter Pattern, and Monochrome configurations
- Progressive scan & flexible readout architecture.
- High frame rate
- High sensitivity - Low noise architecture
- Package pin reserved for device identification

APPLICATIONS

- Industrial Imaging and Inspection
- Traffic
- Aerial Photography

Parameter	Typical Value
Architecture	Interline CCD; Progressive Scan
Total Number of Pixels	4932 (H) x 3300 (V)
Number of Effective Pixels	4888 (H) x 3256 (V)
Number of Active Pixels	4864 (H) x 3232 (V) (15.7M)
Pixel Size	7.4 µm (H) x 7.4 µm (V)
Active Image Size	36.0 mm (H) x 23.9 mm (V) 43.2 mm (diag) 35mm optical format
Aspect Ratio	3:2
Number of Outputs	1, 2, or 4
Charge Capacity	44,000 electrons
Output Sensitivity	9.7 µV/e^- (low), 33 µV/e^- (high)
Quantum Efficiency Mono (-ABA) R, G, B (-CBA)	 48% 32%, 41%, 39%
Base ISO -ABA -CBA, -PBA	 350 130, 310 (respectively)
Read Noise (f= 40MHz)	12 electrons rms
Dark Current Photodiode / VCCD	 1 / 145 electrons/s
Dark Current Doubling Temp Photodiode / VCCD	 7 °C / 9 °C
Dynamic Range High gain amp (40 MHz) Dual amp, 2x2 bin (40 MHz)	 70 dB 82 dB
Charge Transfer Efficiency	0.999999
Blooming Suppression	> 1000 X
Smear	-115 dB
Image Lag	< 10 electrons
Maximum Pixel Clock Speed	40MHz
Maximum Frame Rates Quad/Dual/Single Output	 8 / 4 / 2 fps
Package	72 pin PGA
Cover Glass	AR Coated, 2 Sides

All parameters are specified at T = 40 °C unless otherwise noted.

Ordering Information

Catalog Number	Product Name	Description	Marking Code
4H2212	KAI-16070-AXA-JD-B1	Monochrome, Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Grade 1	KAI-16070-AXA Serial Number
4H2213	KAI-16070-AXA-JD-B2	Monochrome, Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Grade 2	
4H2189	KAI-16070-AXA-JD-AE	Monochrome, Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Engineering Grade	
4H2214	KAI-16070-CXA-JD-B1	Color (Bayer RGB), Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Grade 1	KAI-16070-CXA Serial Number
4H2215	KAI-16070-CXA-JD-B2	Color (Bayer RGB), Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Grade 2	
4H2185	KAI-16070-CXA-JD-AE	Color (Bayer RGB), Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Engineering Grade	
4H2216	KAI-16070-PXA-JD-B1	Color (TRUESENSE Sparse CFA), Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Grade 1	KAI-16070-PXA Serial Number
4H2217	KAI-16070-PXA-JD-B2	Color (TRUESENSE Sparse CFA), Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Grade 2	
4H2187	KAI-16070-PXA-JD-AE	Color (TRUESENSE Sparse CFA), Special Microlens, PGA Package, Sealed Clear Cover Glass with AR coating (both sides), Engineering Grade	

See Application Note *Product Naming Convention* for a full description of the naming convention used for image sensors. For reference documentation, including information on evaluation kits, please visit our web site at www.truesenseimaging.com.

Please address all inquiries and purchase orders to:

Truesense Imaging, Inc.
1964 Lake Avenue
Rochester, New York 14615

Phone: (585) 784-5500
E-mail: info@truesenseimaging.com

Device Description

ARCHITECTURE

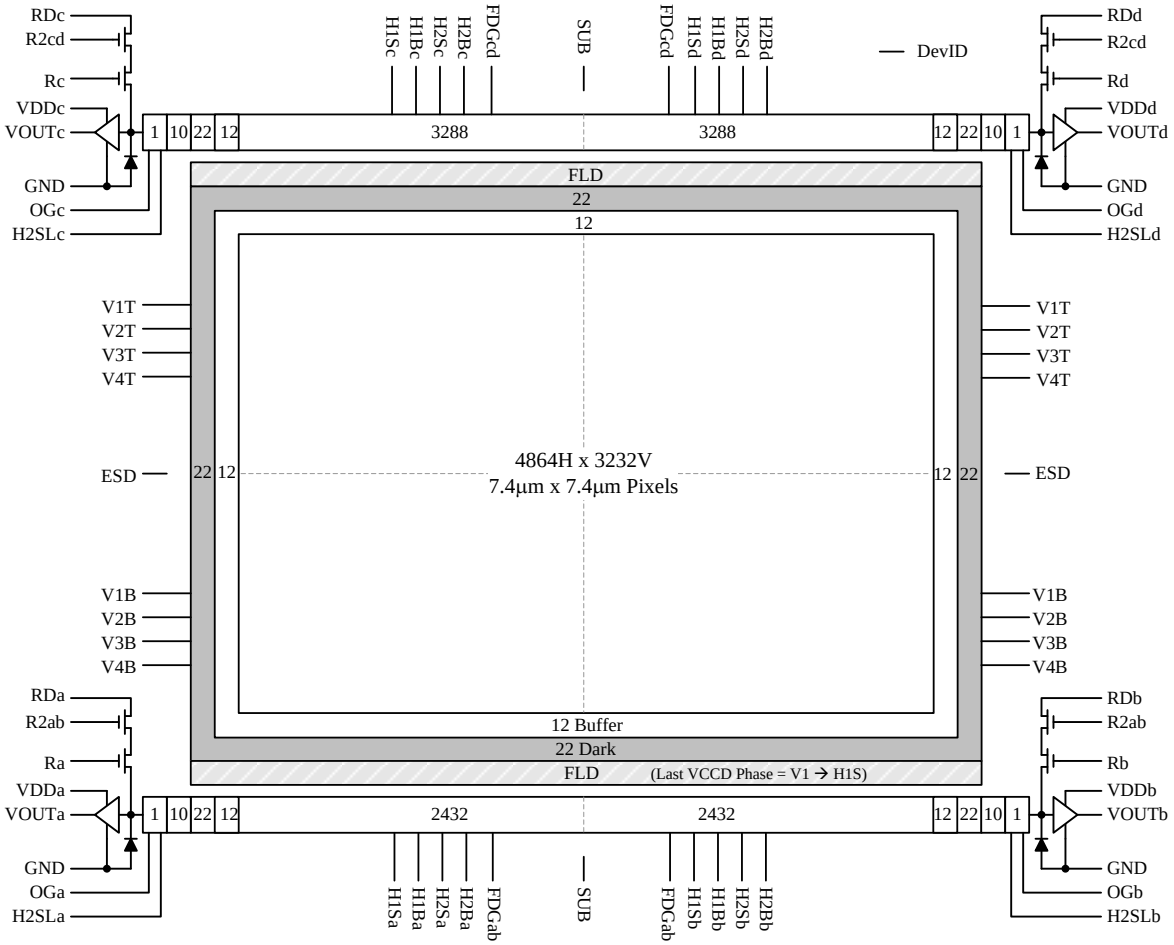


Figure 1: Block Diagram

Dark Reference Pixels

There are 22 dark reference rows at the top and 22 dark rows at the bottom of the image sensor. The dark rows are not entirely dark and so should not be used for a dark reference level. Use the 22 dark columns on the left or right side of the image sensor as a dark reference.

Under normal circumstances use only the center 20 columns of the 22 column dark reference due to potential light leakage.

Dummy Pixels

Within each horizontal shift register there are 11 leading additional shift phases. These pixels are designated as dummy pixels and should not be used to determine a dark reference level.

In addition, there is one dummy row of pixels at the top and bottom of the image.

Active Buffer Pixels

12 unshielded pixels adjacent to any leading or trailing dark reference regions are classified as active buffer pixels. These pixels are light sensitive but are not tested for defects and non-uniformities.

Image Acquisition

An electronic representation of an image is formed when incident photons falling on the sensor plane create electron-hole pairs within the individual silicon photodiodes. These photoelectrons are collected locally by the formation of potential wells at each photosite. Below photodiode saturation, the number of photoelectrons collected at each pixel is linearly dependent upon light level and exposure time and non-linearly dependent on wavelength. When the photodiodes charge capacity is reached, excess electrons are discharged into the substrate to prevent blooming

ESD Protection

Adherence to the power-up and power-down sequence is critical. Failure to follow the proper power-up and power-down sequences may cause damage to the sensor. See Power Up and Power Down Sequence section.

BAYER COLOR FILTER PATTERN

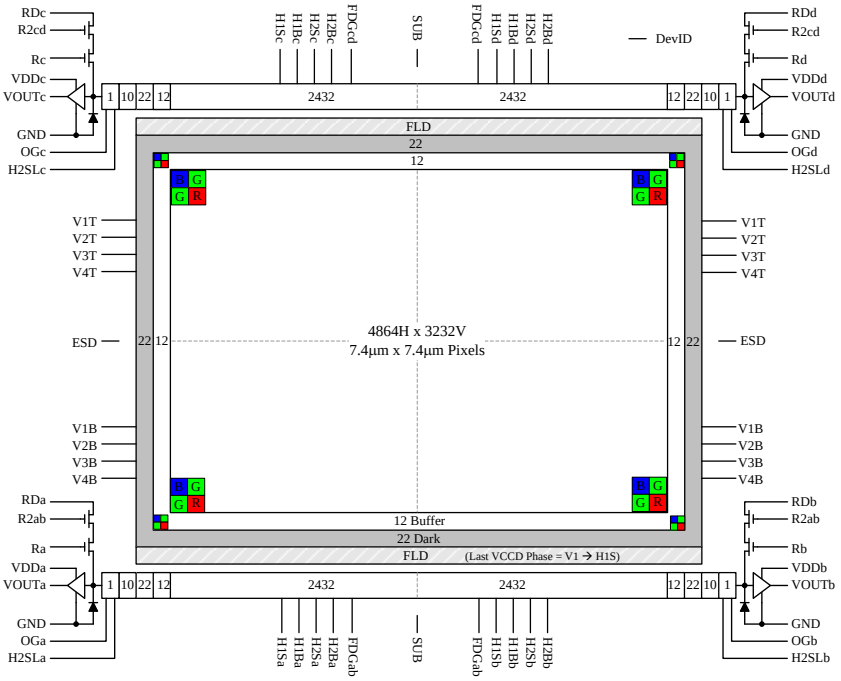


Figure 2: Bayer Color Filter Pattern

TRUESENSE SPARSE COLOR FILTER PATTERN

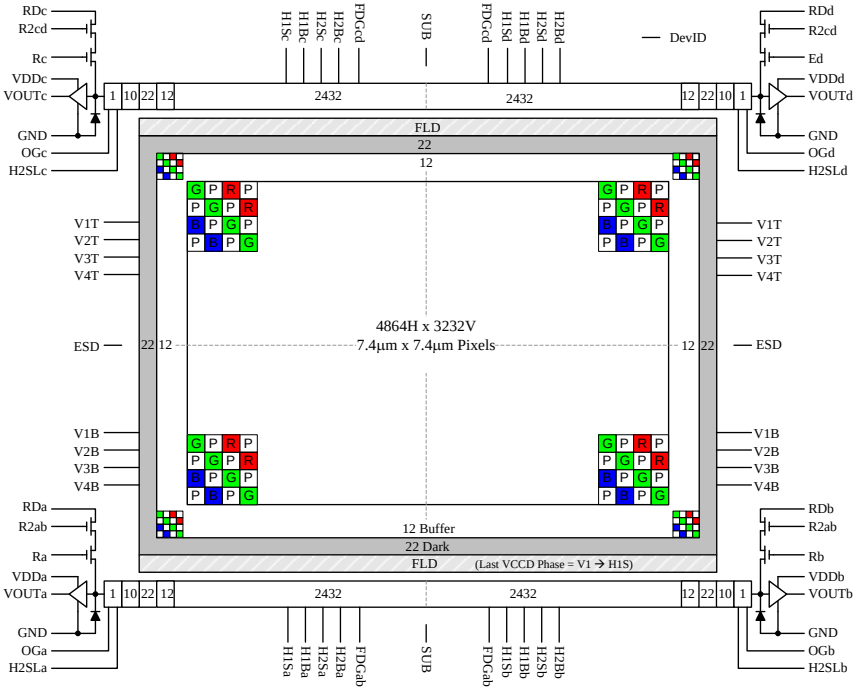


Figure 3: TRUESENSE Sparse Color Filter Pattern

PHYSICAL DESCRIPTION

Pin Description and Device Orientation

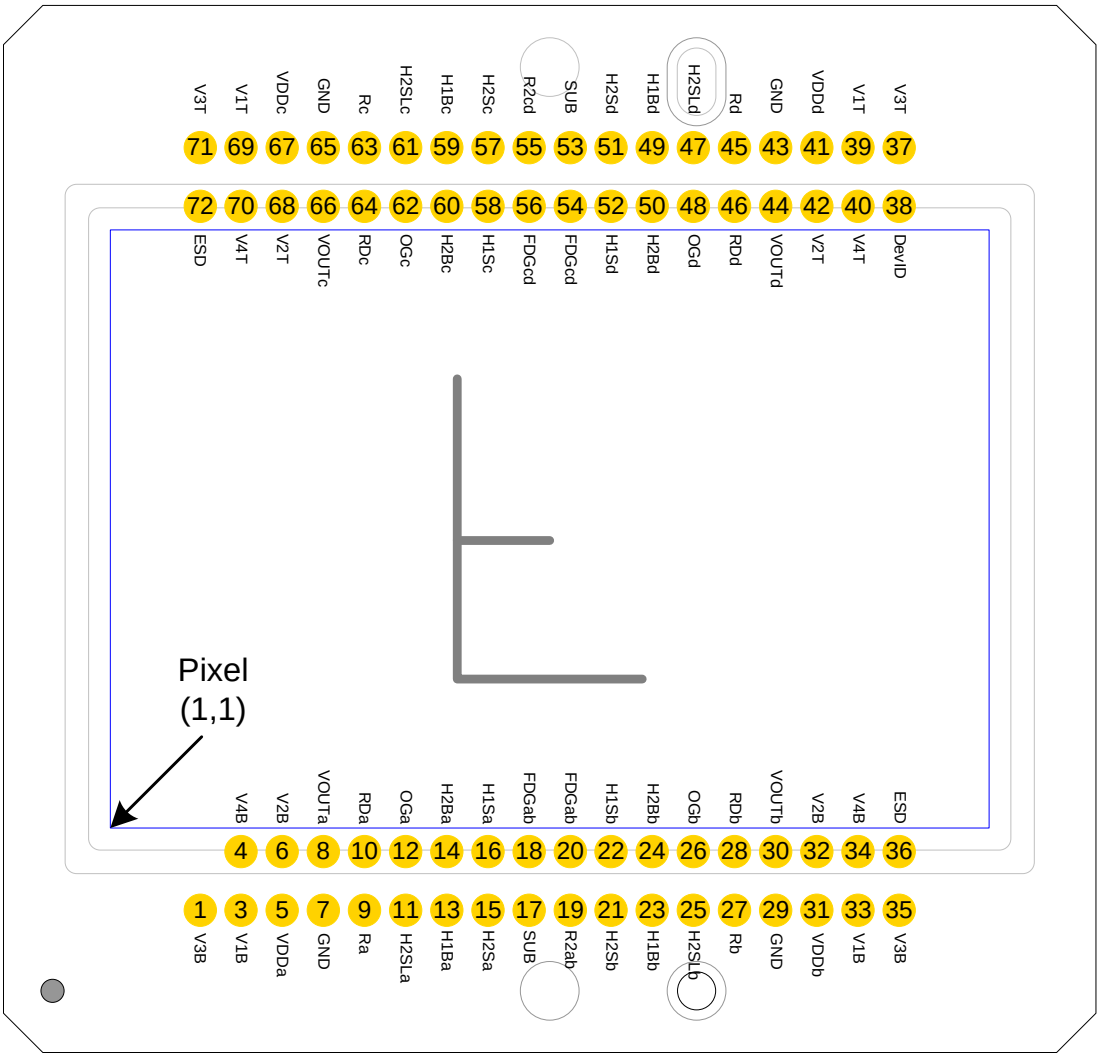


Figure 4: Package Pin Designations - Top View

Pin	Name	Description
1	V3B	Vertical CCD Clock, Phase 3, Bottom
[2]		[no pin – keyed]
3	V1B	Vertical CCD Clock, Phase 1, Bottom
4	V4B	Vertical CCD Clock, Phase 4, Bottom
5	VDDa	Output Amplifier Supply, Quadrant a
6	V2B	Vertical CCD Clock, Phase 2, Bottom
7	GND	Ground
8	VOUTa	Video Output, Quadrant a
9	Ra	Reset Gate, standard (high) gain, Quadrant a
10	RDa	Reset Drain, Quadrant a
11	H2SLa	Horizontal CCD Clock, Phase 2, Storage, Last Phase, Quadrant a
12	OGa	Output Gate, Quadrant a
13	H1Ba	Horizontal CCD Clock, Phase 1, Barrier, Quadrant a
14	H2Ba	Horizontal CCD Clock, Phase 2, Barrier, Quadrant a
15	H2Sa	Horizontal CCD Clock, Phase 2, Storage, Quadrant a
16	H1Sa	Horizontal CCD Clock, Phase 1, Storage, Quadrant a
17	SUB	Substrate
18	FDGab	Fast Line Dump Gate, Bottom
19	R2ab	Reset Gate, low gain, Quadrants a&b
20	FDGab	Fast Line Dump Gate, Bottom
21	H2Sb	Horizontal CCD Clock, Phase 2, Storage, Quadrant b
22	H1Sb	Horizontal CCD Clock, Phase 1, Storage, Quadrant b
23	H1Bb	Horizontal CCD Clock, Phase 1, Barrier, Quadrant b
24	H2Bb	Horizontal CCD Clock, Phase 2, Barrier, Quadrant b
25	H2SLb	Horizontal CCD Clock, Phase 2, Storage, Last Phase, Quadrant b
26	OGb	Output Gate, Quadrant b
27	Rb	Reset Gate, standard (high) gain, Quadrant b
28	RDb	Reset Drain, Quadrant b
29	GND	Ground
30	VOUTb	Video Output, Quadrant b
31	VDDb	Output Amplifier Supply, Quadrant b
32	V2B	Vertical CCD Clock, Phase 2, Bottom
33	V1B	Vertical CCD Clock, Phase 1, Bottom
34	V4B	Vertical CCD Clock, Phase 4, Bottom
35	V3B	Vertical CCD Clock, Phase 3, Bottom
36	ESD	ESD Protection Disable

Pin	Name	Description
72	ESD	ESD Protection Disable
71	V3T	Vertical CCD Clock, Phase 3, Top
70	V4T	Vertical CCD Clock, Phase 4, Top
69	V1T	Vertical CCD Clock, Phase 1, Top
68	V2T	Vertical CCD Clock, Phase 2, Top
67	VDDc	Output Amplifier Supply, Quadrant c
66	VOUTc	Video Output, Quadrant c
65	GND	Ground
64	RDc	Reset Drain, Quadrant c
63	Rc	Reset Gate, standard (high) gain, Quadrant c
62	OGc	Output Gate, Quadrant c
61	H2SLc	Horizontal CCD Clock, Phase 2, Storage, Last Phase, Quadrant c
60	H2Bc	Horizontal CCD Clock, Phase 2, Barrier, Quadrant c
59	H1Bc	Horizontal CCD Clock, Phase 1, Barrier, Quadrant c
58	H1Sc	Horizontal CCD Clock, Phase 1, Storage, Quadrant c
57	H2Sc	Horizontal CCD Clock, Phase 2, Storage, Quadrant c
56	FDGcd	Fast Line Dump Gate, Top
55	R2cd	Reset Gate, low gain, Quadrants c&d
54	FDGcd	Fast Line Dump Gate, Top
53	SUB	Substrate
52	H1Sd	Horizontal CCD Clock, Phase 1, Storage, Quadrant d
51	H2Sd	Horizontal CCD Clock, Phase 2, Storage, Quadrant d
50	H2Bd	Horizontal CCD Clock, Phase 2, Barrier, Quadrant d
49	H1Bd	Horizontal CCD Clock, Phase 1, Barrier, Quadrant d
48	OGd	Output Gate, Quadrant b
47	H2SLd	Horizontal CCD Clock, Phase 2, Storage, Last Phase, Quadrant d
46	RDd	Reset Drain, Quadrant d
45	Rd	Reset Gate, standard (high) gain, Quadrant d
44	VOUTd	Video Output, Quadrant d
43	GND	Ground
42	V2T	Vertical CCD Clock, Phase 2, Top
41	VDDd	Output Amplifier Supply, Quadrant d
40	V4T	Vertical CCD Clock, Phase 4, Top
39	V1T	Vertical CCD Clock, Phase 1, Top
38	DevID	Device Identification
37	V3T	Vertical CCD Clock, Phase 3, Top

Notes:

1. Liked named pins are internally connected and should have a common drive signal.

Imaging Performance

TYPICAL OPERATION CONDITIONS

Unless otherwise noted, the Imaging Performance Specifications are measured using the following conditions.

Description	Condition	Notes
Light Source	Continuous red, green and blue LED illumination	1
Operation	Nominal operating voltages and timing	

Notes:
1. For monochrome sensor, only green LED used.

SPECIFICATIONS

All Configurations

Description	Symbol	Min.	Nom.	Max.	Units	Sampling Plan	Temperature Tested At (°C)	Notes
Dark Field Global Non-Uniformity	DSNU	-	-	5	mVpp	Die	27, 40	
Bright Field Global Non-Uniformity		-	2	12	%rms	Die	27, 40	1
Bright Field Global Peak to Peak Non-Uniformity	PRNU	-	10	30	%pp	Die	27, 40	1
Bright Field Center Non-Uniformity		-	1	2	%rms	Die	27, 40	1
Maximum Photo-response Nonlinearity High Gain (4,000 to 20,000 electrons) High Gain (4,000 to 40,000 electrons) Low Gain (8,000 to 80,000 electrons)	 NL_HG1 NL_HG2 NL_LG1	 - - -	 2 3 6	 - - -	 % % %	Design		
Maximum Gain Difference Between Outputs	ΔG	-	10	-	%	Design		2
Horizontal CCD Charge Capacity	HNe	-	90	-	ke^-	Design		
Vertical CCD Charge Capacity	VNe	-	60	-	ke^-	Design		
Photodiode Charge Capacity	PNe	-	44	-	ke^-	Die	27, 40	3
Floating Diffusion Capacity – High Gain	FNe_HG	40	-	-	ke^-	Die	27, 40	
Floating Diffusion Capacity – Low Gain	FNe_LG	160	-	-	ke^-	Die	27, 40	
Linear Saturation Level – High Gain	Lsat_HG	-	40	-	ke^-	Design		
Linear Saturation Level – Low Gain	Lsat_LG	-	160	-	ke^-	Design		
Horizontal CCD Charge Transfer Efficiency	HCTE	0.999995	0.999999	-		Die		
Vertical CCD Charge Transfer Efficiency	VCTE	0.999995	0.999999	-		Die		
Photodiode Dark Current	Ipd	-	2	70	e/p/s	Die	40	
Vertical CCD Dark Current	Ivd	-	200	600	e/p/s	Die	40	
Image Lag	Lag	-	-	10	e^-	Design		
Antiblooming Factor	Xab	1000	-	-		Design		
Vertical Smear	Smr	-	-115	-	dB	Design		
Read Noise	n_{e-T}	-	12/45	-	e^-rms	Design	High gain/low gain	4
Dynamic Range, standard	DR	-	70.5	-	dB	Design		4, 5
Dynamic Range, extended linear Dynamic range mode (XLDR)	XLDR	-	82.5	-	dB	Design		4, 5
Output Amplifier DC Offset	V_{odc}	5	9.0	14	V	Die	27, 40	
Output Amplifier Bandwidth	f_{-3db}	-	250	-	MHz	Design		6
Output Amplifier Impedance	R_{OUT}	100	127	200	Ohms	Die	27, 40	
Output Amplifier Sensitivity High Gain Low Gain	ΔV/ΔN	 - -	 33 9.7	 - -	 $\mu V/e^-$	Design		

KAI-16070-AXA and KAI-16070-PXA Configurations

Description	Symbol	Min.	Nom.	Max.	Units	Sampling Plan	Temperature Tested At (°C)	Notes
Peak Quantum Efficiency	QE_{max}	-	48	-	%	Design		
Peak Quantum Efficiency Wavelength	λQE	-	500	-	nm	Design		

KAI-16070-CXA and KAI-16070-PXA Configurations

Description		Symbol	Min.	Nom.	Max.	Units	Sampling Plan	Temperature Tested At (°C)	Notes
Peak Quantum Efficiency	Blue Green Red	QE_{max}	-	39 41 32	-	%	Design		
Peak Quantum Efficiency Wavelength	Blue Green Red	λQE	-	470 540 620	-	nm	Design		

Notes:

1. Per color
2. Value is over the range of 10% to 90% of photodiode saturation.
3. The operating value of the substrate voltage, VAB, will be marked on the shipping container for each device. The value of VAB is set such that the photodiode charge capacity is 1450 mV. This value is determined while operating the device in the low gain mode. VAB level assigned is valid for both modes; high gain or low gain.
4. At 40 MHz.
5. Uses 20LOG(PNe/ n_{e-T})
6. Assumes 5pF load

Linear Signal Range

High Gain

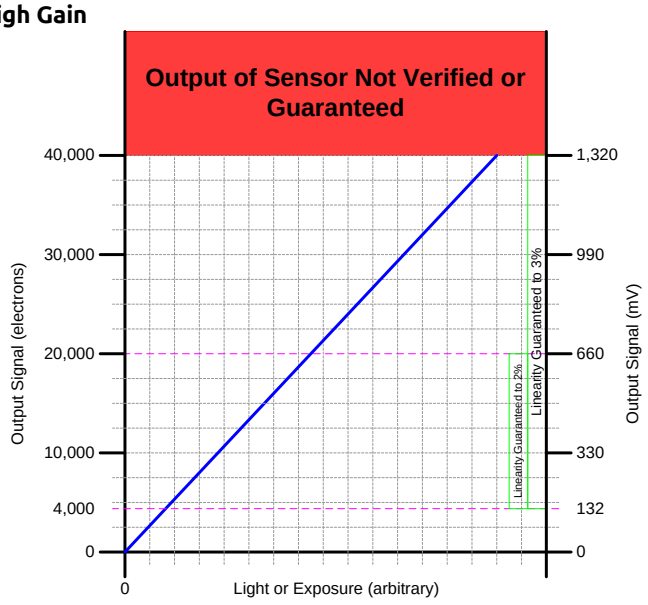


Figure 5: High Gain Linear Signal Range

Low Gain

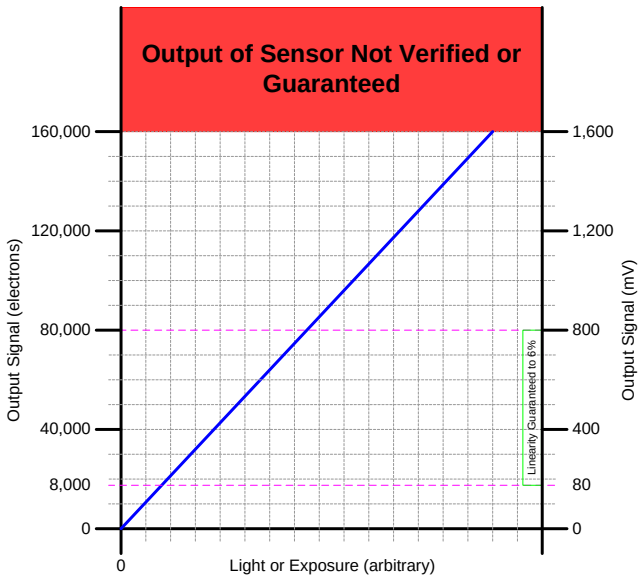


Figure 6: Low Gain Linear Signal Range

Typical Performance Curves

QUANTUM EFFICIENCY

Monochrome with Microlens

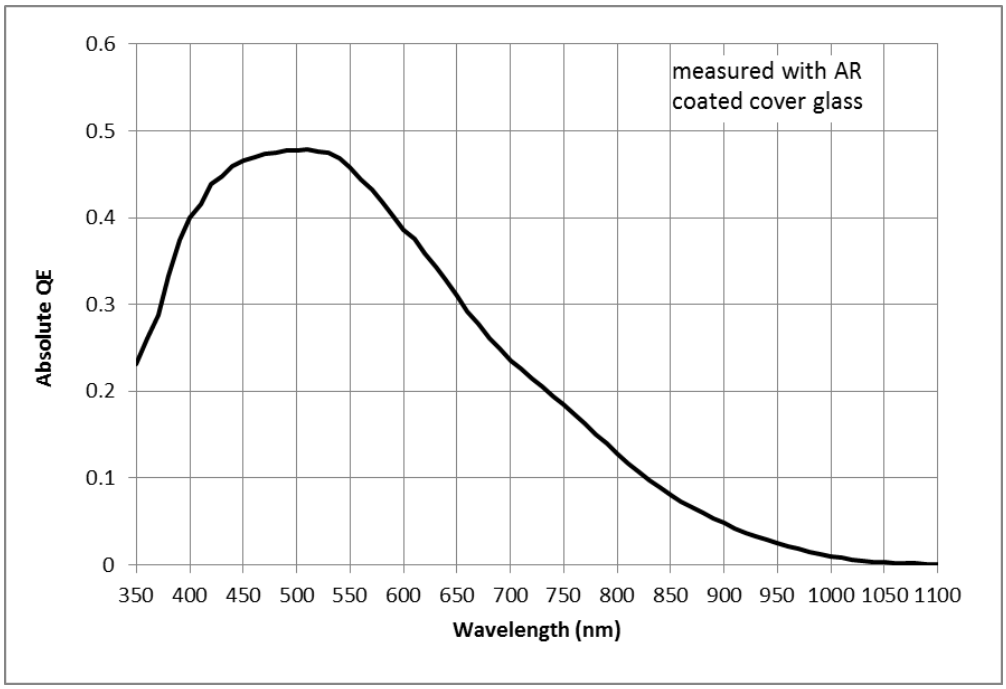


Figure 7: Monochrome with Microlens Quantum Efficiency

Color (Bayer RGB) with Microlens

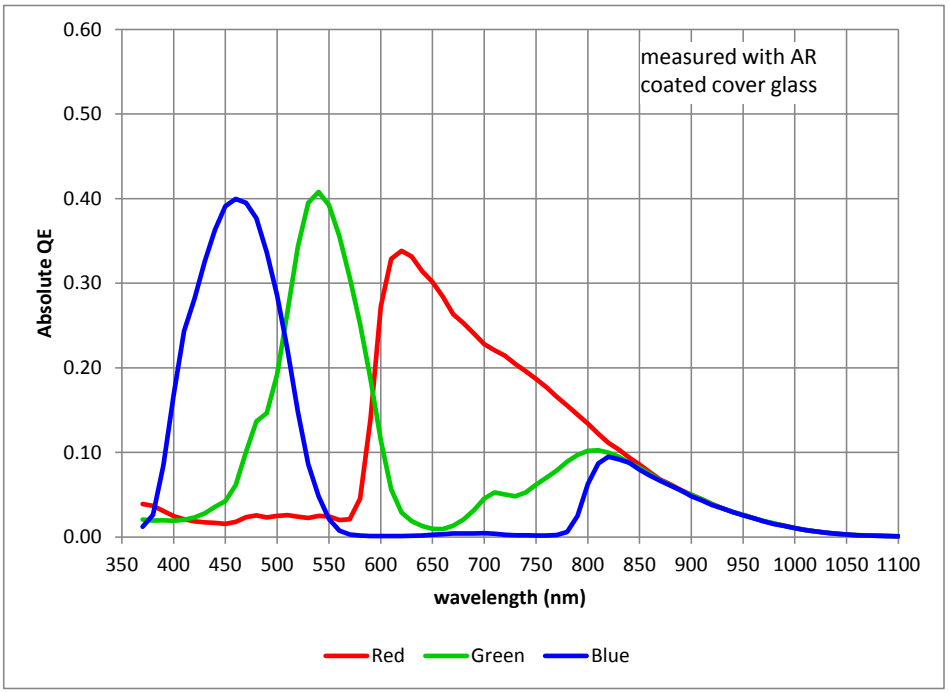


Figure 8: Color (Bayer) with Microlens Quantum Efficiency

Color (TRUESENSE Sparse CFA) with Microlens

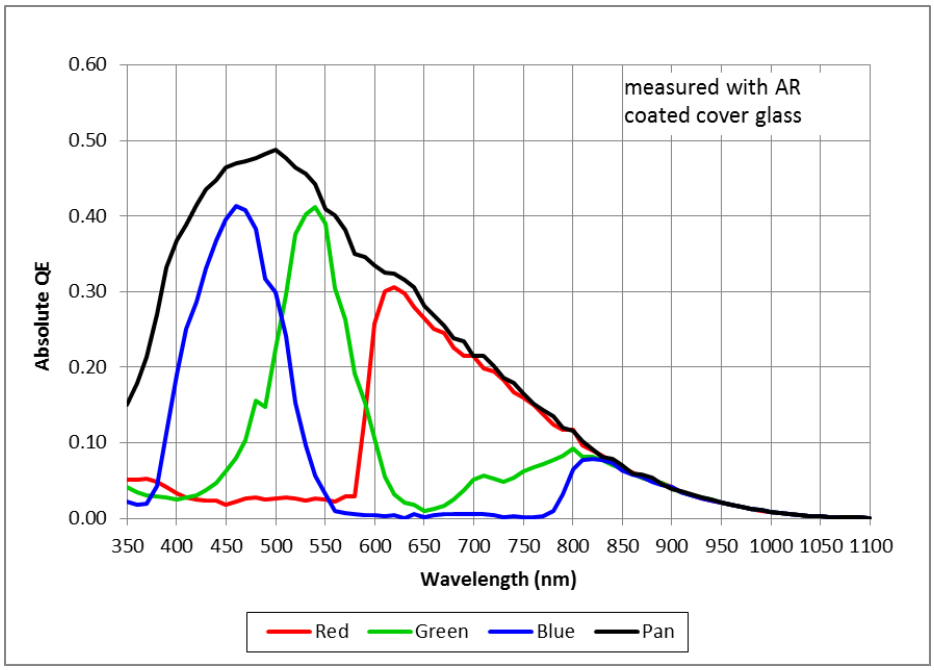


Figure 9: Color (TRUESENSE Sparse CFA) with Microlens Quantum Efficiency

ANGULAR QUANTUM EFFICIENCY

For the curves marked "Horizontal", the incident light angle is varied in a plane parallel to the HCCD. For the curves marked "Vertical", the incident light angle is varied in a plane parallel to the VCCD.

Monochrome with Microlens

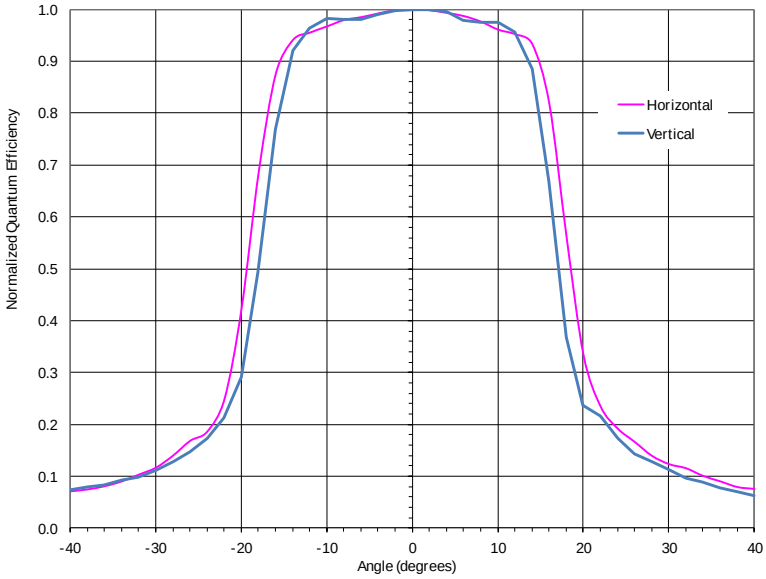


Figure 10: Monochrome with Microlens Angular Quantum Efficiency

Dark Current versus Temperature

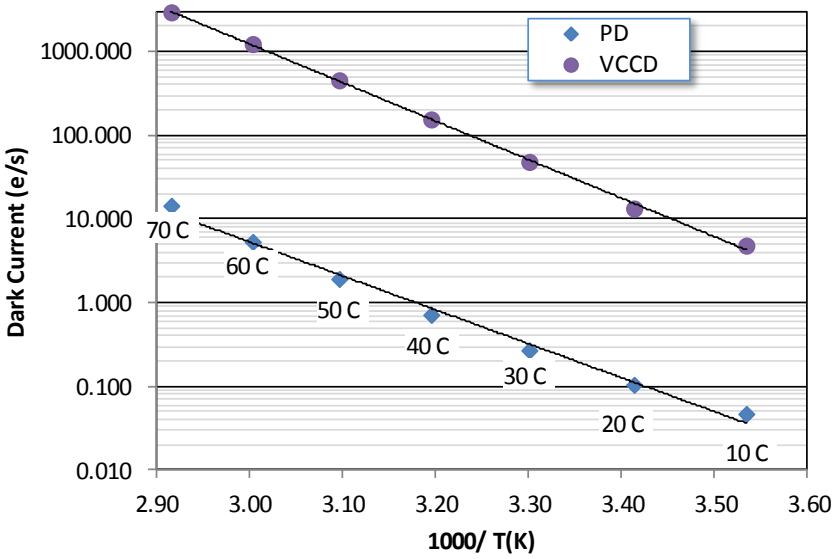
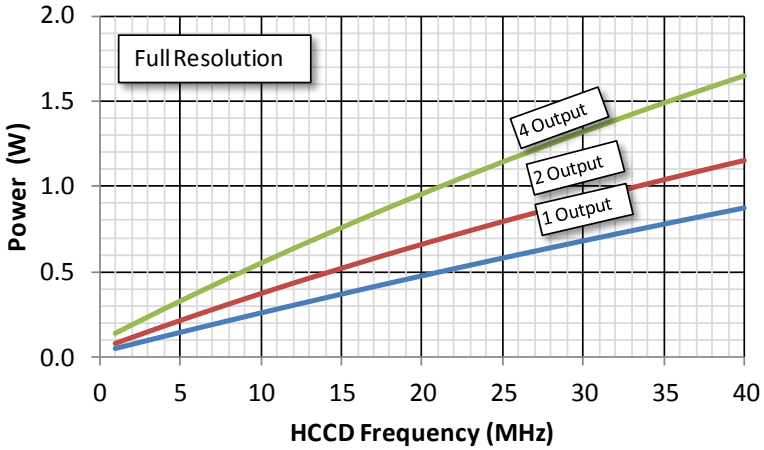
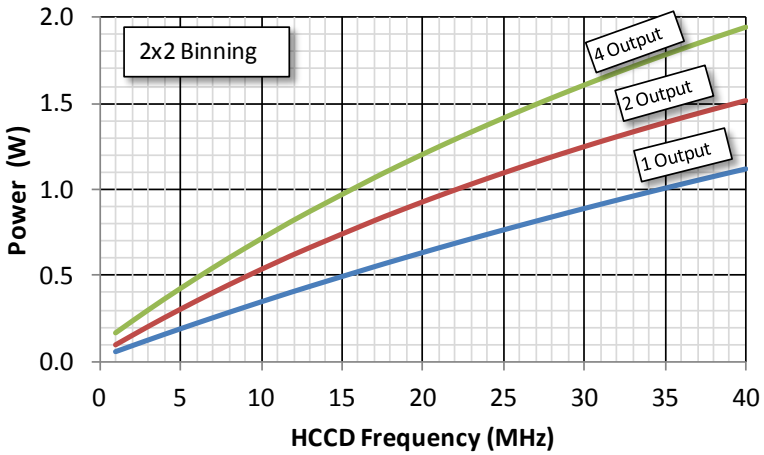


Figure 11: Dark Current versus Temperature

POWER – ESTIMATED





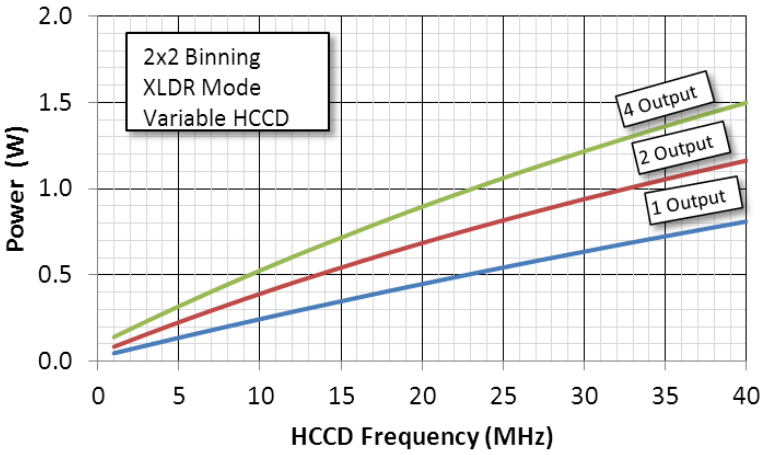
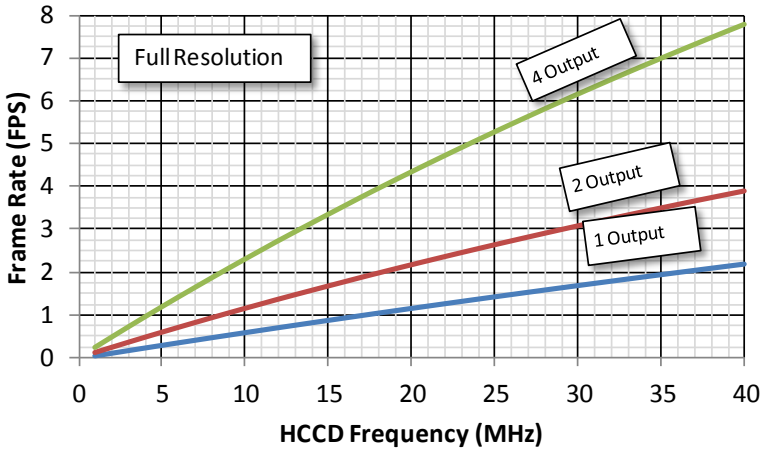
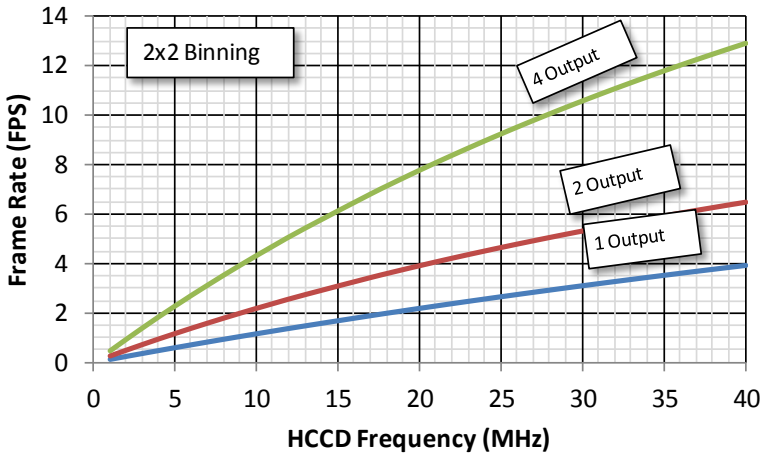


Figure 12: Power

FRAME RATES





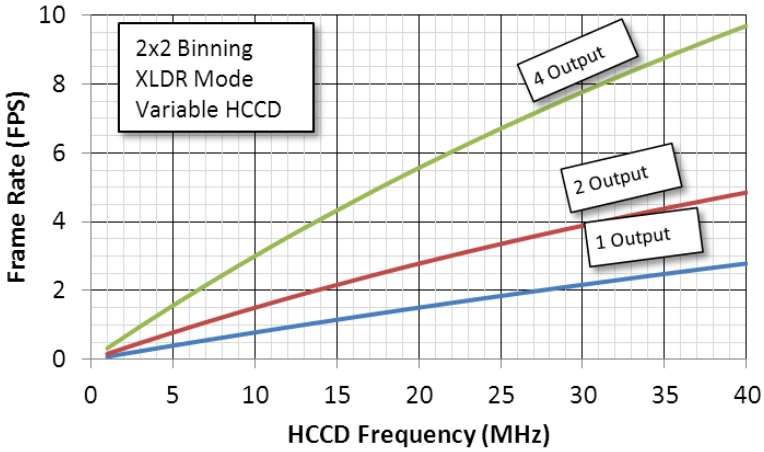


Figure 13: Frame Rates

Defect Definitions

Operation Conditions for Defect Testing at 40 °C

Description	Condition	Notes
Operational Mode	One output using VOUTa, continuous readout	
HCCD Clock Frequency	20 MHz	
Pixels Per Line	5000	1
Lines Per Frame	3354	2
Line Time	266 µsec	
Frame Time	894 msec	
Photodiode Integration Time	PD_Tint = Frame Time = 894 msec, no electronic shutter used	
Temperature	40 °C	
Light Source	Continuous red, green and blue LED illumination	3
Operation	Nominal operating voltages and timing	

Notes
1. Horizontal overclocking used
2. Vertical overclocking used
3. For monochrome sensor, only the green LED is used.

Defect Definitions for Testing at 40 °C

Description	Definition	Grade 1	Grade 2 Mono	Grade 2 Color	Notes
Major dark field defective bright pixel	PD_Tint = Frame Time Defect ≥ 325 mV	150	300	300	1
Major bright field defective dark pixel	Defect ≥ 15 %				
Minor dark field defective bright pixel	PD_Tint = Frame Time Defect ≥ 163 mV	1500	3000	3000	
Cluster Defect	A group of 2 to 19 contiguous major defective pixels, but no more than 4 adjacent defects horizontally.	30	30	30	2
Column defect	A group of more than 10 contiguous major defective pixels along a single column	0	4	15	2

Notes:
1. For the color devices (KAI-16070-CXA and KAI-16070-PXA), a bright field defective pixel deviates by 12% with respect to pixels of the same color.
2. Column and cluster defects are separated by no less than two (2) good pixels in any direction (excluding single pixel defects).
3. Tested at 40 °C with no electronic shutter used.

Operation Conditions for Defect Testing at 27 °C

Description	Condition	Notes
Operational Mode	Two outputs, using VOUTa and VOUTc, continuous readout	
HCCD Clock Frequency	20 MHz	
Pixels Per Line	5000	1
Lines Per Frame	3354	2
Line Time	266 µsec	
Frame Time	894 msec	
Photodiode Integration Time (PD_Tint)	PD_Tint = Frame Time = 894 msec, no electronic shutter used	
Temperature	27 °C	
Light Source	Continuous red, green and blue LED illumination	3
Operation	Nominal operating voltages and timing	

Notes:
1. Horizontal overclocking used
2. Vertical overclocking used
3. For monochrome sensor, only the green LED is used.

Defect Definitions for Testing at 27 °C

Description	Definition	Grade 1	Grade 2 Mono	Grade 2 Color	Notes
Major dark field defective bright pixel	PD_Tint = Frame Time -> Defect ≥ 100 mV	150	300	300	1
Major bright field defective dark pixel	Defect ≥ 15 %				
Minor dark field defective bright pixel	PD_Tint = Frame Time Defect ≥ 52 mV	1500	3000	3000	
Cluster Defect	A group of 2 to 19 contiguous major defective pixels, but no more than 4 adjacent defects horizontally.	30	30	30	2
Column defect	A group of more than 10 contiguous major defective pixels along a single column	0	4	15	2

Notes:
1. For the color devices (KAI-106070-CXA and KAI-16070), a bright field defective pixel deviates by 12% with respect to pixels of the same color.
2. Column and cluster defects are separated by no less than two (2) good pixels in any direction (excluding single pixel defects).
3. Tested at 27 °C with no electronic shutter used.

Defect Map

The defect map supplied with each sensor is based upon testing at an ambient (27 °C) temperature. Minor point defects are not included in the defect map. All defective pixels are reference to pixel 1,1 in the defect maps. See Figure 14: Regions of Interest for the location of pixel 1,1.

Test Definitions

TEST REGIONS OF INTEREST

Image Area ROI: Pixel (1, 1) to Pixel (4888, 3256)

Active Area ROI: Pixel (13, 13) to Pixel (4876, 3244)

Center ROI: Pixel (2345, 1527) to Pixel (2444, 1628)

Only the Active Area ROI pixels are used for performance and defect tests.

OVERCLOCKING

The test system timing is configured such that the sensor is overclocked in both the vertical and horizontal directions. See Figure 14 for a pictorial representation of the regions of interest.

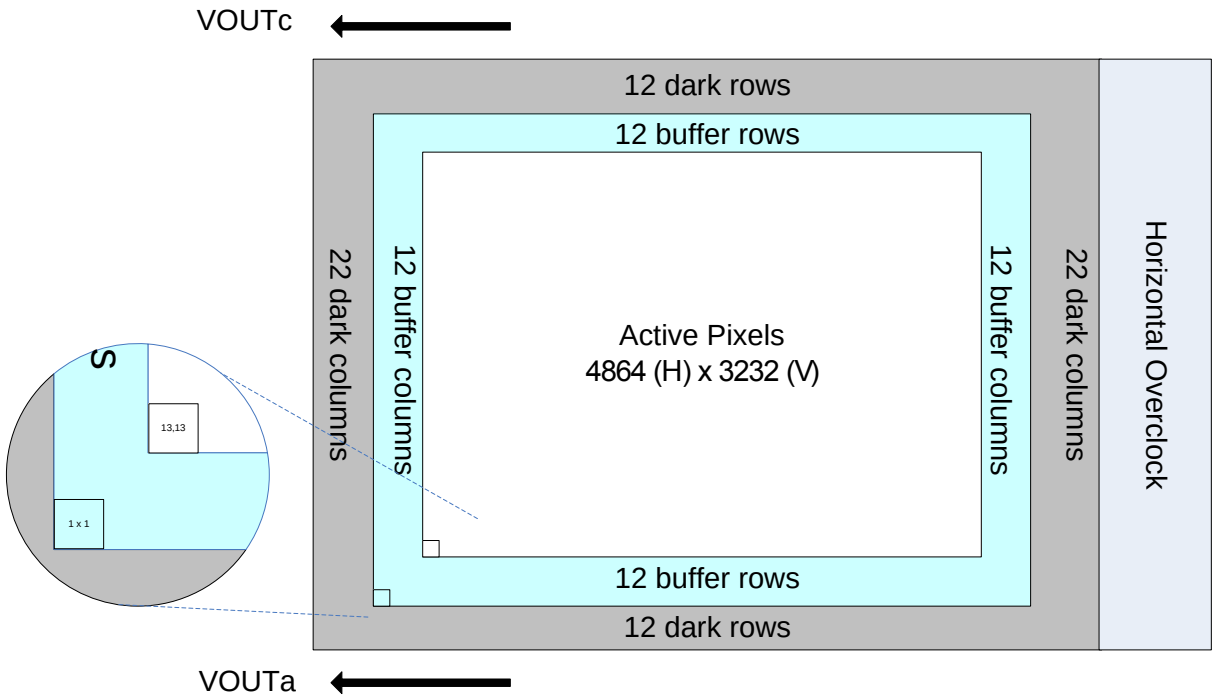


Figure 14: Regions of Interest

TESTS

Dark Field Global Non-Uniformity

This test is performed under dark field conditions. The sensor is partitioned into 1mm x 1mm sub regions, each of which is 135 by 135 pixels in size. The average signal level of each of the sub regions of interest is calculated. The signal level of each of the sub regions of interest is calculated using the following formula:

Signal of ROI[i] = (ROI Average in counts – Horizontal overclock average in counts) * mV per count

Where i = 1 to total # of sub regions. During this calculation on the sub regions of interest, the maximum and minimum signal levels are found. The dark field global uniformity is then calculated as the maximum signal found minus the minimum signal level found.

Units: mVpp (millivolts peak to peak)

Global Non-Uniformity

This test is performed with the imager illuminated to a level such that the output is at 70% of saturation (approximately 924 mV). Prior to this test being performed the substrate voltage has been set such that the charge capacity of the sensor is 1320 mV. Global non-uniformity is defined as

$$\text{Global Non-Uniformity} = 100 * \left(\frac{\text{Active Area Standard Deviation}}{\text{Active Area Signal}} \right)$$ Units: %rms

Active Area Signal = Active Area Average – Dark Column Average

Global Peak to Peak Non-Uniformity

This test is performed with the imager illuminated to a level such that the output is at 70% of saturation (approximately 924 mV). Prior to this test being performed the substrate voltage has been set such that the charge capacity of the sensor is 1320 mV. The sensor is partitioned into sub regions of interest, each of which is 135 by 135 pixels in size. The average signal level of each of the before mentioned sub regions of interest (ROI) is calculated. The signal level of each of the sub regions of interest is calculated using the following formula:

Signal of ROI[i] = (ROI Average in counts – Horizontal overclock average in counts) * mV per count

Where i = 1 to total # of sub regions. During this calculation on the sub regions of interest, the maximum and minimum signal levels are found. The global peak to peak uniformity is then calculated as:

$$\text{Global Uniformity} = 100 * \frac{\text{Maximum Signal} - \text{Minimum Signal}}{\text{Active Area Signal}}$$

Units: %pp

Center Non-Uniformity

This test is performed with the imager illuminated to a level such that the output is at 70% of saturation (approximately 924 mV). Prior to this test being performed the substrate voltage has been set such that the charge capacity of the sensor is 1320 mV. Defects are excluded for the calculation of this test. This test is performed on the center 100 by 100 pixels of the sensor. Center uniformity is defined as:

$$\text{Center ROI Uniformity} = 100 * \left(\frac{\text{Center ROI Standard Deviation}}{\text{Center ROI Signal}} \right)$$

Units: %rms. Center ROI Signal = Center ROI Average – Dark Column Average.

Dark field defect test

This test is performed under dark field conditions. The sensor is partitioned into 1mm x 1mm sub regions, each of which is 135 by 135 pixels in size. In each region of interest, the median value of all pixels is found. For each region of interest, a pixel is marked defective if it is greater than or equal to the median value of that region of interest plus the defect threshold specified in the "Defect Definitions" section.

Bright field defect test

This test is performed with the imager illuminated to a level such that the output is at approximately 924 mV. Prior to this test being performed the substrate voltage has been set such that the charge capacity of the sensor is 1320 mV. The average signal level of all active pixels is found. The bright and dark thresholds are set as:

Dark defect threshold = Active Area Signal * threshold

Bright defect threshold = Active Area Signal * threshold

The sensor is then partitioned into 1mm x 1mm sub regions sub regions of interest, each of which is 135 by 135 pixels in size. In each region of interest, the average value of all pixels is found. For each region of interest, a pixel is marked defective if it is greater than or equal to the median value of that region of interest plus the bright threshold specified or if it is less than or equal to the median value of that region of interest minus the dark threshold specified.

Example for major bright field defective pixels:

- Average value of all active pixels is found to be 924 mV
- Dark defect threshold: 924 mV * 15 % = 138 mV
- Bright defect threshold: 924 mV * 15 % = 138 mV
- Region of interest #1 selected. This region of interest is pixels 13, 13 to pixels 147, 147.
    - Median of this region of interest is found to be 918 mV.
    - Any pixel in this region of interest that is >= (918 + 138 mV) 1062 mV in intensity will be marked defective.
    - Any pixel in this region of interest that is <= (918 - 138 mV) 780 mV in intensity will be marked defective.
- All remaining sub regions of interest are analyzed for defective pixels in the same manner. Any remaining factor of pixels less than 135 pixels that are not covered by this moving ROI is placed over the remaining pixels at the active area boundry. A portion of pixels that were tested in the previous ROI will be retested to keep the test ROI at a full 135 by 135 pixels.

Operation

Absolute Maximum Ratings

Absolute maximum rating is defined as a level or condition that should not be exceeded at any time per the description. If the level or the condition is exceeded, the device will be degraded and may be damaged. Operation at these values will reduce Mean Time to Failure (MTTF).

Description	Symbol	Minimum	Maximum	Units	Notes
Operating Temperature	T_{OP}	-50	+70	°C	1
Humidity	RH	+5	+90	%	2
Output Bias Current	Iout	-	60	mA	3
Off-chip Load	CL	-	10	pF	

Notes:
1. Noise performance will degrade at higher temperatures.
2. T=25 °C. Excessive humidity will degrade MTTF.
3. Total for all outputs. Maximum current is -15 mA for each output. Avoid shorting output pins to ground or any low impedance source during operation. Amplifier bandwidth increases at higher current and lower load capacitance at the expense of reduced gain (sensitivity).

Absolute Maximum Voltage Ratings Between Pins and Ground

Description	Minimum	Maximum	Units	Notes
VDDα, VOUTα	-0.4	17.5	V	1
RDα	-0.4	15.5	V	1
V1B, V1T	ESD – 0.4	ESD + 24.0	V	
V2B, V2T, V3B, V3T, V4B, V4T	ESD – 0.4	ESD + 14.0	V	
FDGab, FDGcd	ESD – 0.4	ESD + 14.0		
H1Sα, H1Bα, H2Sα, H2Bα, H2SLα, Rα, OGα	ESD – 0.4	ESD + 14.0	V	1
ESD	-10.0	0.0	V	
SUB	-0.4	+40.0	V	2

Notes:
1. α denotes a, b, c or d
2. Refer to Application Note *Using Interline CCD Image Sensors in High Intensity Visible Lighting Conditions*

KAI-29050 COMPATIBILITY

The KAI-16070 is pin-for-pin compatible with a camera designed for the KAI-29050 image sensor with the following accommodations:

1. To operate in accordance with a system designed for KAI-29050, the target substrate voltage should be set to be 2.0V higher than the value recorded on the KAI-16070 shipping container. This setting will cause the charge capacity to be limited to 20 Ke^- (or 660mV).
2. On the KAI-16070, pins 19 (R2ab) and 55 (R2cd) should be left floating per the KAI-29050 Device Performance Specification.
3. The KAI-16070 will operate in only the high gain mode (33 µV/e).
4. All timing and voltages are taken from the KAI-29050 specification sheet.
5. The number of horizontal and vertical CCD clock cycles is reduced as appropriate.
6. In addition, if the intent is to operate the KAI-16070 image sensor in a camera designed for the KAI-29050 sensor that has been modified to accept and process the full 40,000 e^- (1,320 mV) output, the following changes to the following voltage bias must be made:

Voltage bias differences	KAI-29050	KAI-16070
pins 10, 28, 46, and 64	12.0V per the specification	Increase this value to 12.6V

Note: To make use of the low gain mode or dual gain mode the KAI-16070 voltages and timing specification must be used.

RESET PIN, LOW-GAIN (R2AB AND R2CD)

The R2ab and R2bc (pins 19 and 55) each have an internal circuit to bias the pins to 4.3 V. This feature assures the device is set to operate in the high gain mode when pins 19 and 55 are not connected in the application to a clock driver (for KAI-29050 compatibility). Typical capacitor coupled drivers will not drive this structure.

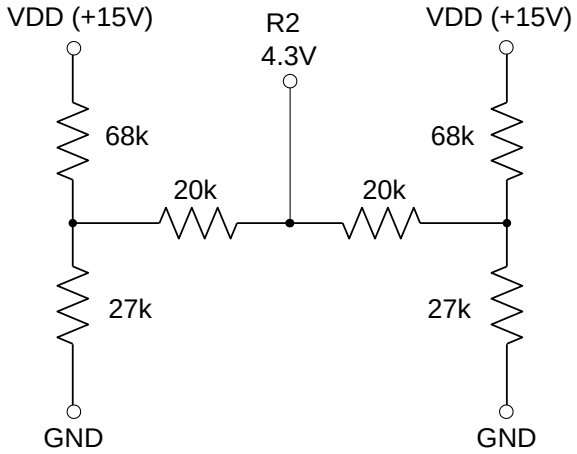


Figure 15: Equivalent Circuit for Reset Gate, low Gain (R2ab and R2cd)

Power Up and Power Down Sequence

Adherence to the power-up and power-down sequence is critical. Failure to follow the proper power-up and power-down sequences may cause damage to the sensor.

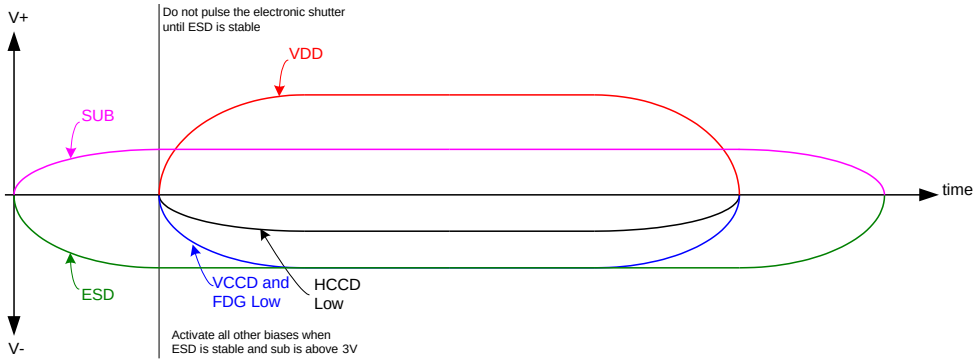
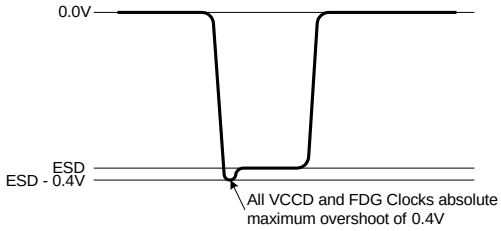


Figure 16: Power Up and Power Down Sequence

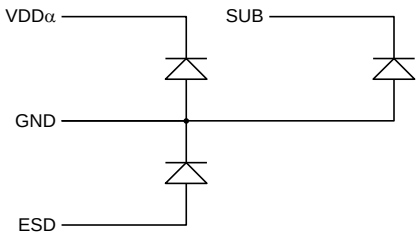
Notes:

1. Activate all other biases when ESD is stable and SUB is above 3V
2. Do not pulse the electronic shutter until ESD is stable
3. VDD cannot be +15V when SUB is 0V
4. The image sensor can be protected from an accidental improper ESD voltage by current limiting the SUB current to less than 10mA. SUB and VDD must always be greater than GND. ESD must always be less than GND. Placing diodes between SUB, VDD, ESD and ground will protect the sensor from accidental overshoots of SUB, VDD and ESD during power on and power off. See the figure below.

The VCCD clock waveform must not have a negative overshoot more than 0.4V below the ESD voltage.



Example of external diode protection for SUB, VDD and ESD. α denotes a, b, c or d.



DC Bias Operating Conditions

Description	Pins	Symbol	Minimum	Nominal	Maximum	Units	Maximum DC Current	Notes
Reset Drain	RDα	RD	+12.4	+12.6	+12.8	V	10 µA	1,9
Output Gate	OGα	OG	-2.2	-2.0	-1.8	V	10 µA	1
Output Amplifier Supply	VDDα	VDD	+14.5	+15.0	+15.5	V	11.0 mA	1, 2
Ground	GND	GND	0.0	0.0	0.0	V	-1.0 mA	
Substrate	SUB	VSUB	+5.0	VAB	VDD	V	50 µA	3, 8
ESD Protection Disable	ESD	ESD	-9.5	-9.0	Vx_L	V	50 µA	6, 7,10
Output Bias Current	VOUTα	Iout	-3.0	-5.0	-10.0	mA	——	1, 4, 5

Notes:

1. α denotes a, b, c or d
2. The maximum DC current is for one output. Idd = Iout + Iss. See Figure 17.
3. The operating value of the substrate voltage, VAB, will be marked on the shipping container for each device. The value of VAB is set such that the photodiode charge capacity is the nominal PNe (see Specifications).
4. An output load sink must be applied to each VOUT pin to activate each output amplifier.
5. Nominal value required for 40MHz operation per output. May be reduced for slower data rates and lower noise.
6. Adherence to the power-up and power-down sequence is critical.
7. ESD maximum value must be less than or equal to V1_L+0.4V and V2_L+0.4V
8. Refer to Application Note Using Interline CCD Image Sensors in High Intensity Visible Lighting Conditions
9. 12.0 V may be used if the total output signal desired is 20,000 e- or less.
10. Where Vx_L is the level set for V1_L, V2_L, V3_L, or V4_L in the application.

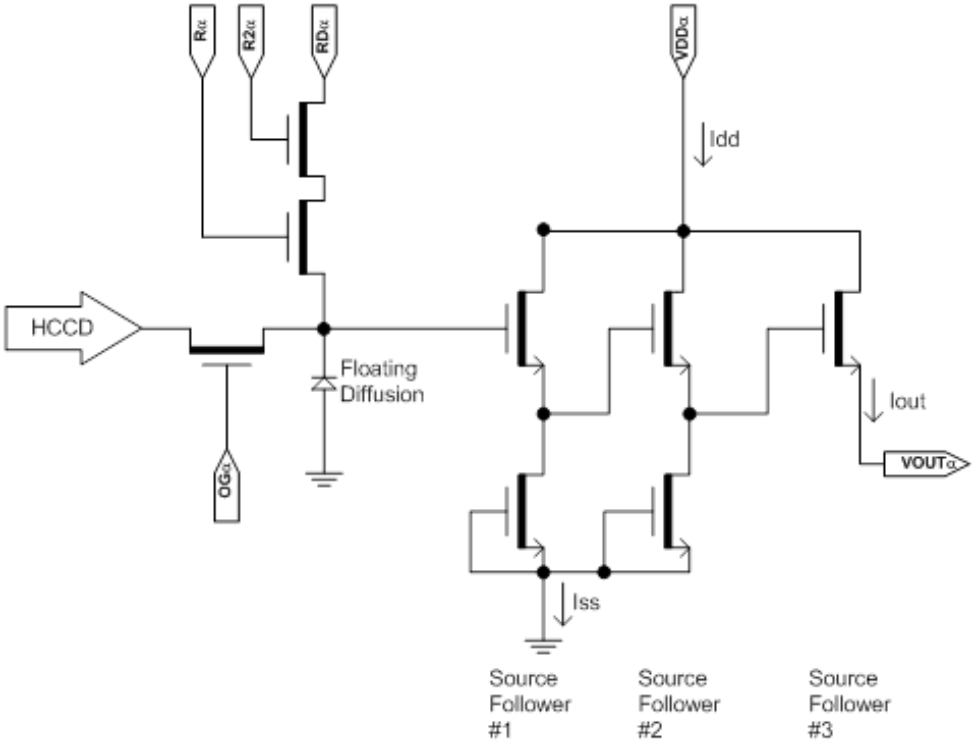


Figure 17: Output Amplifier – showing dual reset pins

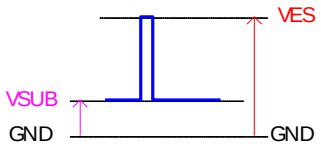
AC Operating Conditions

Clock Levels

Description	Pins	Symbol	Level	Minimum	Nominal	Maximum	Units
Vertical CCD Clock, Phase 1	V1B, V1T[1]	V1_L	Low	-8.2	-8.0	-7.8	V
		V1_M	Mid	-0.2	+0.0	+0.2	
		V1_H	High	+12.8	+13.0	+14.0	
Vertical CCD Clock, Phase 2	V2B, V2T[1]	V2_L	Low	-8.2	-8.0	-7.8	V
		V2_H	High	-0.2	+0.0	+0.2	
Vertical CCD Clock, Phase 3	V3B, V3T[1]	V3_L	Low	-8.2	-8.0	-7.8	V
		V3_H	High	-0.2	+0.0	+0.2	
Vertical CCD Clock, Phase 4	V4B, V4T[1]	V4_L	Low	-8.2	-8.0	-7.8	V
		V4_H	High	-0.2	+0.0	+0.2	
Horizontal CCD Clock, Phase 1 Storage	H1Sα[1]	H1S_L	Low	-5.0 (5)	-4.4	-4.2	V
		H1S_A	Amplitude	+4.2	+4.4	+5.0 (5)	
Horizontal CCD Clock, Phase 1 Barrier	H1Bα[1]	H1B_L	Low	-5.0 (5)	-4.4	-4.2	V
		H1B_A	Amplitude	+4.2	+4.4	+5.0 (5)	
Horizontal CCD Clock, Phase 2 Storage	H2Sα[1]	H2S_L	Low	-5.0 (5)	-4.4	-4.2	V
		H2S_A	Amplitude	+4.2	+4.4	+5.0 (5)	
Horizontal CCD Clock, Phase 2 Barrier	H2Bα[1]	H2B_L	Low	-5.0 (5)	-4.4	-4.2	V
		H2B_A	Amplitude	+4.2	+4.4	+5.0 (5)	
Horizontal CCD Clock, Last Phase[2]	H2SLα[1]	H2SL_L	Low	-5.2	-5.0	-4.8	V
		H2SL_A	Amplitude	+4.8	+5.0	+5.2	
Reset Gate 1	Rα[1]	R_L[3]	Low	-3.2	-3.0	-2.8	V
		R_A	Amplitude	+6.0	——	+6.4	
Reset Gate 2	R2αb, R2cd	R_L[3]	Low	-2.0	-1.8	-1.6	V
		R_A	Amplitude	+6.0	——	+6.4	
Electronic Shutter[4]	SUB	VES	High	+29.0	+30.0	+40.0	V
Fast Line Dump Gate	FDGα[1]	FDG_L	Low	-8.2	-8.0	-7.8	V
		FDG_H	High	+4.5	+5.0	+5.5	

Notes:
1. α denotes a, b, c or d
2. Use separate clock driver for improved speed performance.
3. Reset low should be set to -3 volts for signal levels greater than 40,000 electrons.
4. Refer to Application Note *Using Interline CCD Image Sensors in High Intensity Visible Lighting Conditions*
5. If the minimum horizontal clock low level is used (-5.0V), then the maximum horizontal clock amplitude should be used (5V amplitude) to create a -5.0V to 0.0V clock

The figure below shows the DC bias (VSUB) and AC clock (VES) applied to the SUB pin. Both the DC bias and AC clock are referenced to ground.



Capacitance

	V1B	V2B	V3B	V4B	V1T	V2T	V3T	V4T	GND	All Pins	Units
V1B		17	11	14	6	5	6	4	24	88	nF
V2B			21	10	5	3	4	3	7	74	nF
V3B				19	6	5	6	4	8	83	nF
V4B					5	4	5	3	23	76	nF
V1T						14	11	17	24	86	nF
V2T							16	6	22	75	nF
V3T								19	11	84	nF
V4T									5	73	nF
FDGT	0.6	0.5	0.5	0.4	16	3.1	1.0	1.1	94	117	pF
FDGB	0.6	0.5	0.5	0.4	16	3.1	1.0	1.1	94	117	pF
VSub	2	2	2	2	2	2	2	2	11	11	nF

	H2S	H1B	H2B	GND	All Pins	Units
H1S	45	75	44	196	360	pF
H2S		47	41	281	368	pF
H1B			12	313	324	pF
H2B				293	293	pF

Notes:
1. Table shows typical cross capacitance between pins of the device.
2. Capacitance is total for all like named pins.

DEVICE IDENTIFICATION

The device identification pin (DevID) may be used to identify different members of the Truesense Imaging 5.5-micron and 7.4-micron Interline Transfer CCD Platforms.

Description	Pins	Symbol	Min	Nominal	Max	Units	Max DC Current	Notes
Device Identification	DevID	DevID	32,000	40,000	48,000	Ohms	50 µA	1, 2, 3

Notes:
1. Nominal value subject to verification and/or change during release of preliminary specifications.
2. If the Device Identification is not used, it may be left disconnected.
3. After Device Identification resistance has been read during camera initialization, it is recommended that the circuit be disabled to prevent localized heating of the sensor due to current flow through the R_DeviceID resistor.

Recommended Circuit

Note that V1 must be a different value than V2.

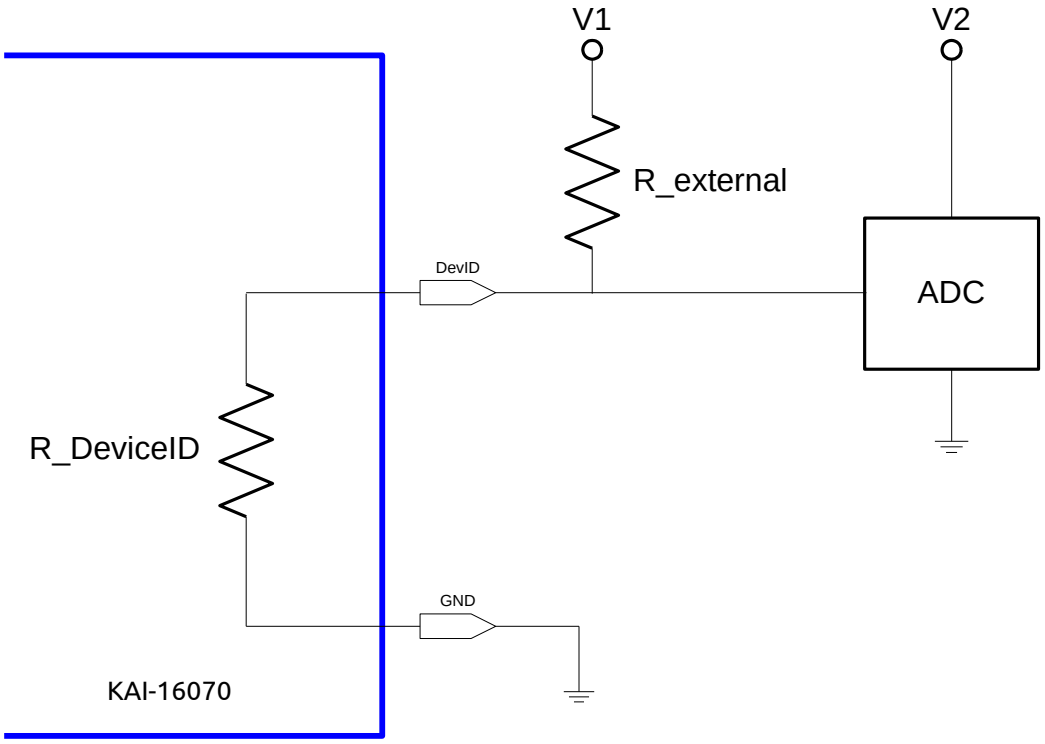


Figure 18: Device Identification Recommended Circuit

Timing

REQUIREMENTS AND CHARACTERISTICS

Description	Symbol	Minimum	Nominal	Maximum	Units	Notes
Photodiode Transfer	t_{pd}	6	-	-	µs	
VCCD Leading Pedestal	t_{3p}	16	-	-	µs	
VCCD Trailing Pedestal	t_{3d}	16	-	-	µs	
VCCD Transfer Delay	t_d	2	-	-	µs	
VCCD Transfer	t_v	4	-	-	µs	
VCCD Rise, Fall Times	t_{VR}, t_{VF}	5	-	10	%	1, 2
FDG Delay	t_{fdg}	2	-	-	µs	
HCCD Delay	t_{hs}	2	-	-	µs	
HCCD Transfer	t_e	25.0	-	-	ns	
Shutter Transfer	t_{sub}	2	-	-	µs	
Shutter Delay	t_{hd}	2	-	-	µs	
Reset Pulse	t_r	2.5	-	-	ns	
Reset – Video Delay	t_{rv}	-	2.2	-	ns	
H2SL – Video Delay	t_{hv}	-	2.2	-	ns	
Line Time	t_{line}	77.9	-	-	µs	Dual HCCD Readout
		140	-	-		Single HCCD Readout
Frame Time	t_{frame}	129	-	-	ms	Quad HCCD Readout
		257	-	-		Dual HCCD Readout
		461	-	-		Single HCCD Readout
Line Time (XLDR Bin 2x2)	t_{line}	124.9	-	-	µs	Dual HCCD Readout
		217.4	-	-		Single HCCD Readout
Frame Time (XLDR Bin 2x2) Constant HCCD timing3	t_{frame}	133	-	-	ms	Quad HCCD Readout
		267	-	-		Dual HCCD Readout
		466	-	-		Single HCCD Readout
Frame Time (XLDR Bin 2x2) Variable HCCD Timing	t_{frame}	103	-	-	ms	Quad HCCD Readout
		206	-	-		Dual HCCD Readout
		359	-	-		Single HCCD Readout

Notes:
1. Refer to Figure 36: VCCD Clock Rise Time and Fall Time .
2. Relative to the pulse width, t_V.

TIMING FLOW CHARTS

In the timing flow charts the number of HCCD clock cycles per row, NH, and the number of VCCD clock cycles per frame, NV, is given by the following table.

	Full Resolution		1/4 Resolution		XLDR	
	NV	NH	NV	NH	NV	NH
Quad	1650	2477	825	1238	825	1238
Dual VOUTa, VOUTc	1650	4943	825	2471	825	2471
Dual VOUTa, VOUTb	3278	2477	1639	1238	1639	1238
Single VOUTa	3278	4943	1639	2471	1639	2471

Table 1: Values for NH and NV when operating the sensor in the various modes of resolution.

The time to read out one line TL = Line timing (See Table 3) + NH / (pixel frequency).

The time to read out one frame TF = NV * TL + frame timing (See Table 2).

No Electronic Shutter

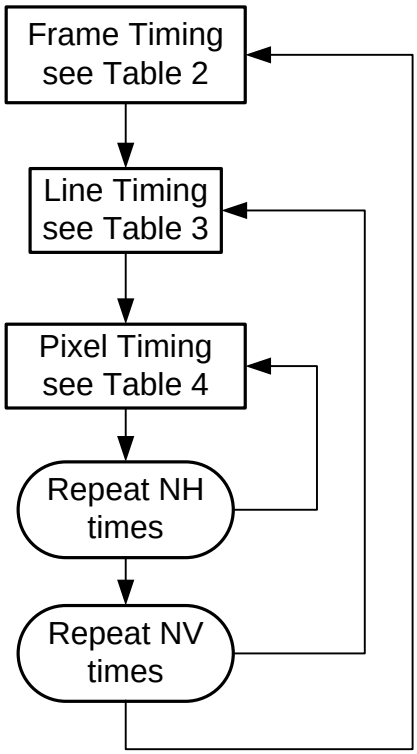


Figure 19: The timing flow when the electronic shutter is not used.

In this case the photodiode exposure time is equal to the time to read out an image. This flow chart applies to both full resolution and ¼ resolution modes.

Using the Electronic Shutter

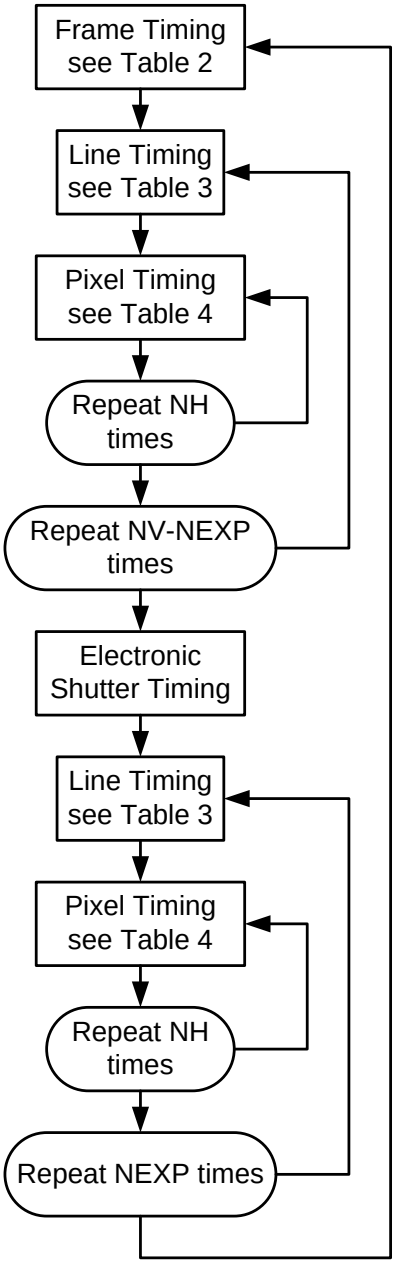


Figure 20: The timing flow chart using the electronic shutter for exposure control.

This flow chart applies to both full resolution and ¼ resolution modes. The exposure time begins on the falling edge of the shutter pulse on the SUB pin. The exposure time ends on the falling edge of the +13 V to 0 V transition of the V1T and V1B pins. NEXP is varied to change the exposure time in increments of the line time. The electronic shutter timing is obtained from Figure 28.

Window Readout Using the Fast Dump

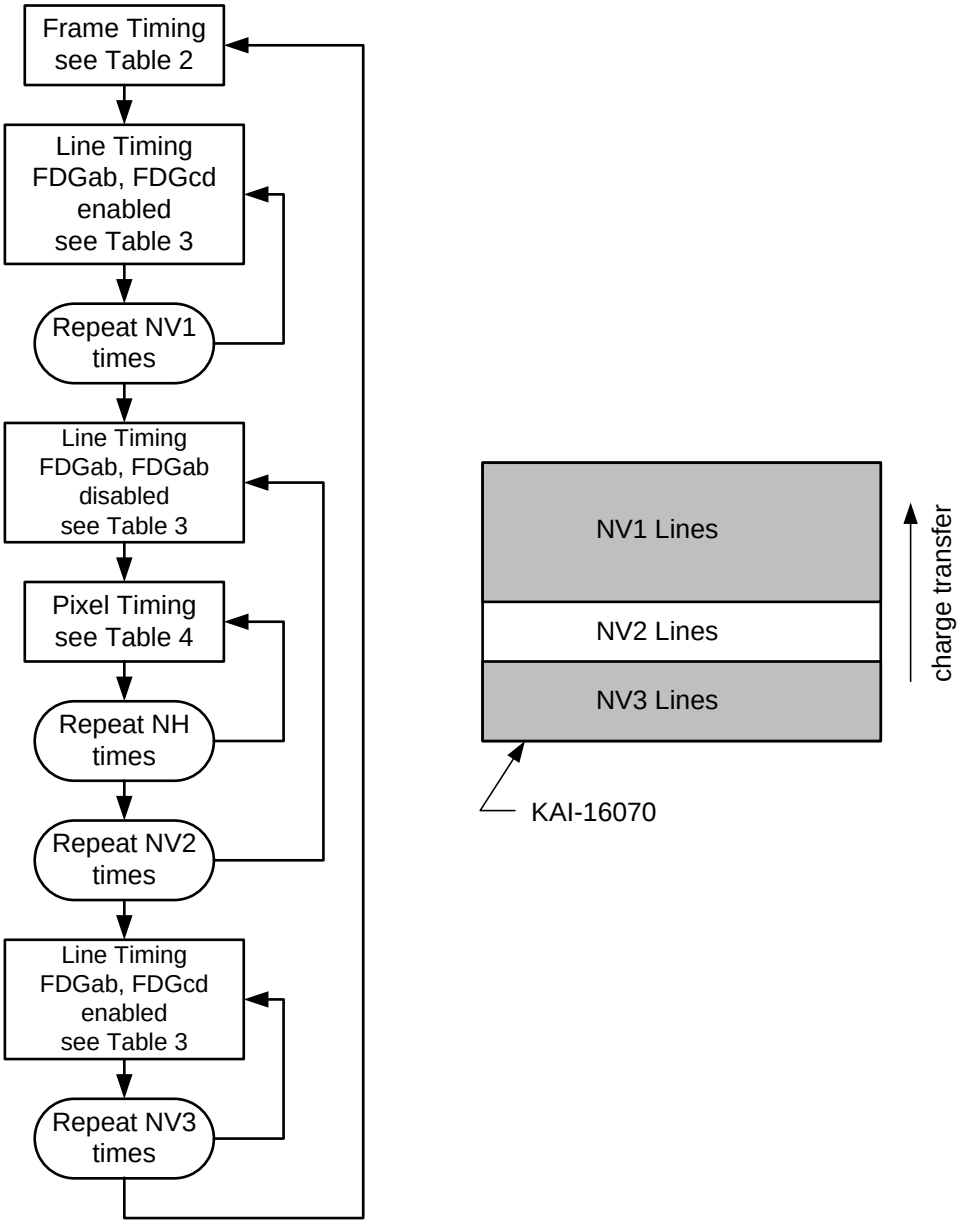


Figure 21: Sub window timing flow chart.

This timing quickly dumps NV1 lines, then reads out NV2 lines, and then quickly dumps another NV3 lines. NV1 + NV2 + NV3 must be greater than or equal to NV. Note when operating in quad or dual VOUTa+VOUTc modes the NV2 valid image lines must be in the center of the pixel array or contained entirely within the bottom half or top half of the pixel array. This is due to the top and bottom middle split of the VCCD. In the single output or dual VOUTa+VOUTb modes the NV2 valid image lines may be located anywhere within the pixel array.

The line timing with the FDGab and FDGcd pins disabled means those pins are held at a constant -9 V. When they are enabled, they are held at +5 V during a line transfer.

Line Sampling Readout Using the Fast Dump

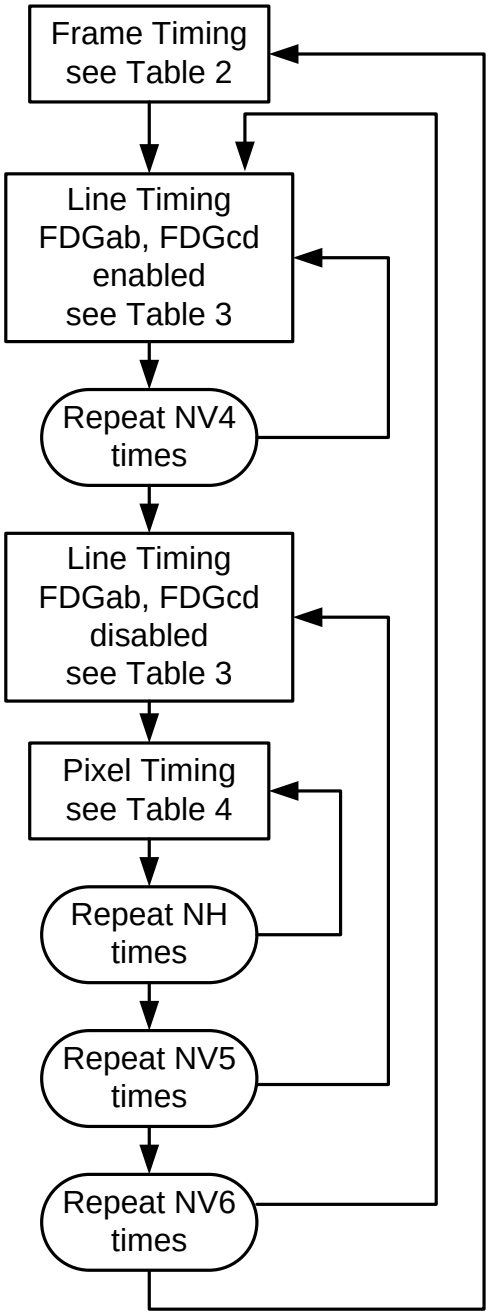


Figure 22: The timing flow chart to alternately skip and read rows for subsampling.

This timing repeats the process of dumping NV4 lines and reading NV5 lines. The total NV6 x (NV4 + NV5) must be greater than or equal to NV. This timing can be used for alternately skipping and reading lines. For example, if NV4 = 2 and NV5 = 1 then every third line will be read out (skip 2 read 1).

TIMING TABLES

Frame Timing

This timing table is for transferring charge from the photodiodes to the VCCD.

	Full Resolution, high gain OR low gain				1/4 Resolution, high gain OR low gain				1/4 Resolution XLDR			
Device Pin	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa
V1T	F1T		F1B		F1T		F1B		F1T		F1B	
V2T	F2T		F4B		F2T		F4B		F2T		F4B	
V3T	F3T		F3B		F3T		F3B		F3T		F3B	
V4T	F4T		F2B		F4T		F2B		F4T		F2B	
V1B	F1B				F1B				F1B			
V2B	F2B				F2B				F2B			
V3B	F3B				F3B				F3B			
V4B	F4B				F4B				F4B			
H1Sa	P1				P1Q				P1XL			
H1Ba	P1				P1Q				P1XL			
H2Sa	P2				P2Q				P2XL			
H2Ba	P2				P2Q				P2XL			
Ra	RHG/RLG				RHGQ/RLGQ				RXL			
H1Sb	P1				P1Q				P1XL			
H1Bb	P1	P2	P1	P2	P1Q	P2Q	P1Q	P2Q	P1XL	P2XL	P1XL	P2XL
H2Sb	P2				P2Q				P2XL			
H2Bb	P2	P1	P2	P1	P2Q	P1Q	P2Q	P1Q	P2XL	P1XL	P2XL	P1XL
Rb	RHG/RLG	Note 1	RHG/RLG	Note 1	RHGQ/RLGQ	Note 1	RHGQ/RLGQ	Note 1	RXL	Note 1	RXL	Note 1
R2ab	R2HG/R2LG				R2HGQ/R2LGQ				R2XL			
FDGab	-9 V				-9 V				-9 V			
H1Sc	P1		Note 1		P1Q		Note 1		P1XL		Note 1	
H1Bc	P1		Note 1		P1Q		Note 1		P1XL		Note 1	
H2Sc	P2		Note 1		P2Q		Note 1		P2XL		Note 1	
H2Bc	P2		Note 1		P2Q		Note 1		P2XL		Note 1	
Rc	RHG/RLG		Note 1		RHGQ/RLGQ		Note 1		RXL		Note 1	
H1Sd	P1		Note 1		P1Q		Note 1		P1XL		Note 1	
H1Bd	P1	P2	Note 1		P1Q	P2Q	Note 1		P1XL	P2XL	Note 1	
H2Sd	P2		Note 1		P2Q		Note 1		P2XL		Note 1	
H2Bd	P2	P1	Note 1		P2Q	P1Q	Note 1		P2XL	P1XL	Note 1	
Rd	RHG/RLG	Note 1			RHGQ/RLGQ	Note 1			RXL	Note 1		
R2cd	R2HG/R2LG		Note 1		R2HGQ/R2LGQ		Note 1		R2XL		Note 1	
FDGcd	-9 V				-9 V				-9 V			
SHP	SHP1				SHPQ				Note 4			
SHD	SHD1				SHDQ				Note 5			

Table 2: Frame timing table

Notes:

1. This clock should be held at its high level voltage (0V) or held at +5.0V for compatibility with TRUESENSE 5.5 micron Interline Transfer CCD family of products.
2. SHP and SHD are the sample clocks for the analog front end signal processor.
3. This note left intentionally empty.
4. Use SHPLG for the AFE processing the low gain signal. Use SHPHG for the AFE processing the high gain signal.
5. Use SHDLG for the AFE processing the low gain signal. Use SHDHG for the AFE processing the high gain signal.

Line Timing

This timing is for transferring one line of charge from the VCCD to the HCCD

	Full Resolution, high gain OR low gain				1/4 Resolution, high gain OR low gain				1/4 Resolution XLDR			
Device Pin	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa
V1T	L1T		L1B		2x L1T		2x L1B		2x L1T		2x L1B	
V2T	L2T		L4B		2x L2T		2x L4B		2x L2T		2x L4B	
V3T	L3T		L3B		2x L3T		2x L3B		2x L3T		2x L3B	
V4T	L4T		L2B		2x L4T		2x L2B		2x L4T		2x L2B	
V1B	L1B				2x L1B				2x L1B			
V2B	L2B				2x L2B				2x L2B			
V3B	L3B				2x L3B				2x L3B			
V4B	L4B				2x L4B				2x L4B			
H1Sa	P1L				P1LQ				P3XL			
H1Ba	P1L				P1LQ				P3XL			
H2Sa	P2L				P2LQ				P4XL			
H2Ba	P2L				P2LQ				P4XL			
Ra	RHG/RLG				RHGQ/RLGQ				RXL			
H1Sb	P1L				P1LQ				P3XL			
H1Bb	P1L	P2L	P1L	P2L	P1LQ	P2LQ	P1LQ	P2LQ	P3XL	P4XL	P3XL	P4XL
H2Sb	P2L				P2LQ				P4XL			
H2Bb	P2L	P1L	P2L	P1L	P2LQ	P1LQ	P2LQ	P1LQ	P4XL	P3XL	P4XL	P3XL
Rb	RHG/RLG	Note 1	RHG/RLG	Note 1	RHGQ/RLGQ	Note 1	RHGQ/RLGQ	Note 1	RXL	Note 1	RXL	Note 1
R2ab	R2HG/R2LG				R2HGQ/R2LGQ				R2XL			
FDGab	-9 V				-9 V				-9 V			
H1Sc	P1L		Note 1		P1LQ		Note 1		P3XL		Note 1	
H1Bc	P1L		Note 1		P1LQ		Note 1		P3XL		Note 1	
H2Sc	P2L		Note 1		P2LQ		Note 1		P4XL		Note 1	
H2Bc	P2L		Note 1		P2LQ		Note 1		P4XL		Note 1	
Rc	RHG/RLG		Note 1		RHGQ/RLGQ		Note 1		RXL		Note 1	
H1Sd	P1L		Note 1		P1LQ		Note 1		P3XL		Note 1	
H1Bd	P1L	P2L	Note 1		P1LQ	P2LQ	Note 1		P3XL	P4XL	Note 1	
H2Sd	P2L		Note 1		P2LQ		Note 1		P4XL		Note 1	
H2Bd	P2L	P1L	Note 1		P2LQ	P1LQ	Note 1		P4XL	P3XL	Note 1	
Rd	RHG/RLG	Note 1			RHGQ/RLGQ	Note 1			RXL	Note 1		
R2cd	R2HG/R2LG		Note 1		R2HGQ/R2LGQ		Note 1		R2XL		Note 1	
FDGcd	-9 V				-9 V				-9 V			
SHP	SHP1				SHPQ				Note 4			
SHD	SHD1				SHDQ				Note 5			

Table 3: Line timing table

Notes:

1. This clock should be held at its high level voltage (0V) or held at +5.0V for compatibility with TRUESENSE 5.5 micron Interline Transfer CCD family of products.
2. SHP and SHD are the sample clocks for the analog front end signal processor.
3. The notation 2x L1B means repeat the L1B timing twice for every line, this sums two rows into the HCCD.
4. Use SHPLG for the AFE processing the low gain signal. Use SHPHG for the AFE processing the high gain signal.
5. Use SHDLG for the AFE processing the low gain signal. Use SHDHG for the AFE processing the high gain signal.

Pixel Timing

This timing is for transferring one pixel from the HCCD to the output amplifier.

	Full Resolution, high gain OR low gain				1/4 Resolution, high gain OR low gain				1/4 Resolution XLDR			
Device Pin	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa	Quad	Dual VOUTa VOUTc	Dual VOUTa VOUTb	Single VOUTa
V1T	-9 V				-9 V				-9 V			
V2T	-9 V				-9 V				-9 V			
V3T	0 V				0 V				0 V			
V4T	0 V				0 V				0 V			
V1B	-9 V				-9 V				-9 V			
V2B	0 V				0 V				0 V			
V3B	0 V				0 V				0 V			
V4B	-9 V				-9 V				-9 V			
H1Sa	P1				P1Q				P1XL			
H1Ba	P1				P1Q				P1XL			
H2Sa	P2				P2Q				P2XL			
H2Ba	P2				P2Q				P2XL			
Ra	RHG/RLG				RHGQ/RLGQ				RXL			
H1Sb	P1				P1Q				P1XL			
H1Bb	P1	P2	P1	P2	P1Q	P2Q	P1Q	P2Q	P1XL	P2XL	P1XL	P2XL
H2Sb	P2				P2Q				P2XL			
H2Bb	P2	P1	P2	P1	P2Q	P1Q	P2Q	P1Q	P2XL	P1XL	P2XL	P1XL
Rb	RHG/RLG	Note 1	RHG/RLG	Note 1	RHGQ/RLGQ	Note 1	RHGQ/RLGQ	Note 1	RXL	Note 1	RXL	Note 1
R2ab	R2HG/R2LG				R2HGQ/R2LGQ				R2XL			
FDGab	-9 V				-9 V				-9 V			
H1Sc	P1		Note 1		P1Q		Note 1		P1XL		Note 1	
H1Bc	P1		Note 1		P1Q		Note 1		P1XL		Note 1	
H2Sc	P2		Note 1		P2Q		Note 1		P2XL		Note 1	
H2Bc	P2		Note 1		P2Q		Note 1		P2XL		Note 1	
Rc	RHG/RLG		Note 1		RHGQ/RLGQ		Note 1		RXL		Note 1	
H1Sd	P1		Note 1		P1Q		Note 1		P1XL		Note 1	
H1Bd	P1	P2	Note 1		P1Q	P2Q	Note 1		P1XL	P2XL	Note 1	
H2Sd	P2		Note 1		P2Q		Note 1		P2XL		Note 1	
H2Bd	P2	P1	Note 1		P2Q	P1Q	Note 1		P2XL	P1XL	Note 1	
Rd	RHG/RLG	Note 1			RHGQ/RLGQ	Note 1			RXL	Note 1		
R2cd	R2HG/R2LG		Note 1		R2HGQ/R2LGQ		Note 1		R2XL		Note 1	
FDGcd	-9 V				-9 V				-9 V			
SHP	SHP1				SHPQ				Note 4			
SHD	SHD1				SHDQ				Note 5			

Table 4: Pixel timing table

Notes:

1. This clock should be held at its high level voltage (0V) or held at +5.0V for compatibility with TRUESENSE 5.5 micron Interline Transfer CCD family of products.
2. SHP and SHD are the sample clocks for the analog front end signal processor.
3. This note left intentionally empty.
4. Use SHPLG for the AFE processing the low gain signal. Use SHPHG for the AFE processing the high gain signal.
5. Use SHDLG for the AFE processing the low gain signal. Use SHDHG for the AFE processing the high gain signal.

TIMING DIAGRAMS

Frame Timing Diagrams

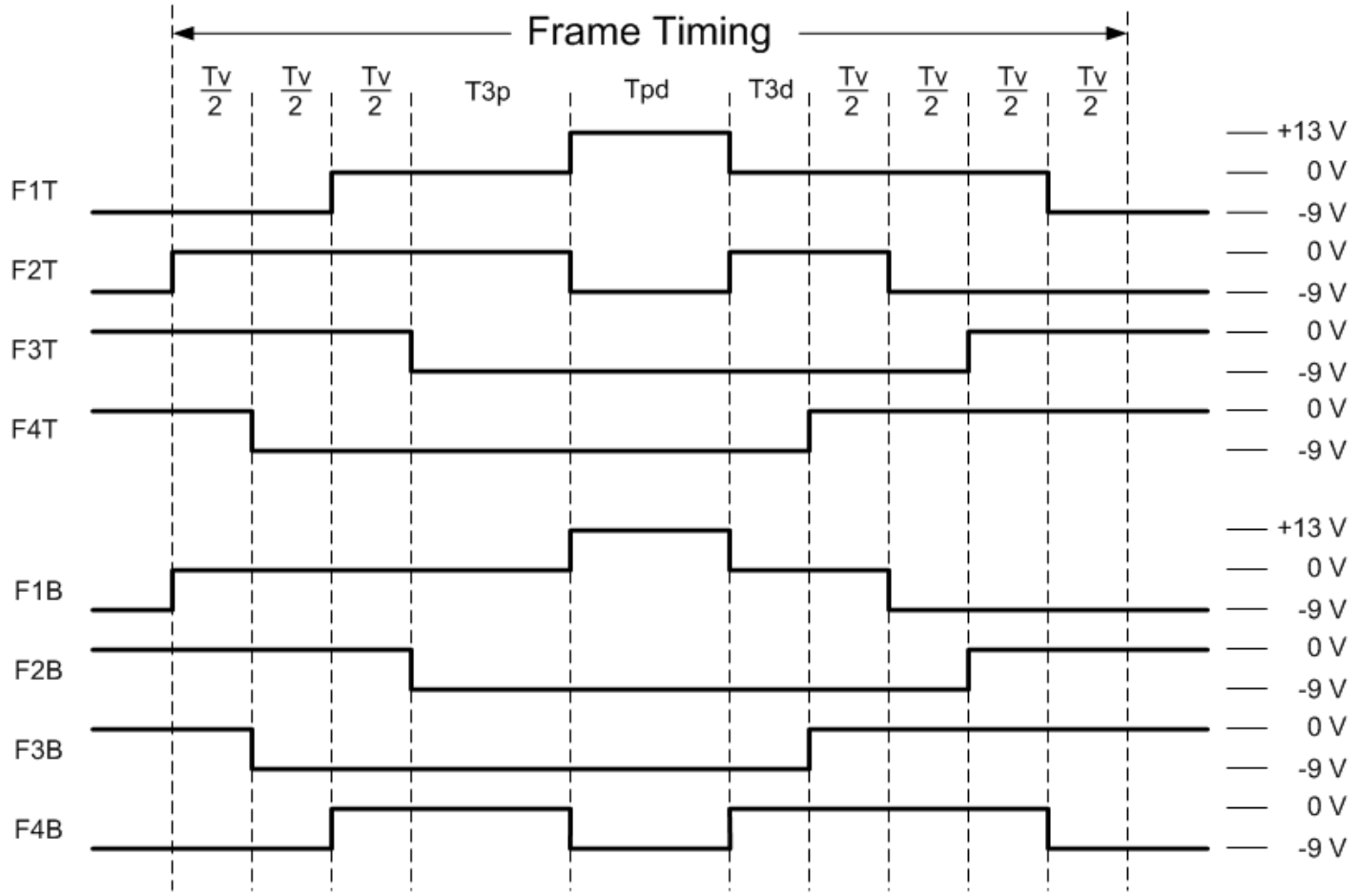


Figure 23: The frame timing diagram. See Table 2 for pin assignments

The charge in the photodiodes begins its transfer to the VCCD on the rising edge of the +13 V pulse and is completed by the falling edge of the +13 V pulse on F1T and F1B. During the time period when F1T and F1B are at +13V antiblooming protection is disabled. The photodiode integration time ends on the falling edge of the +13 V pulse.

Line Timing Diagrams

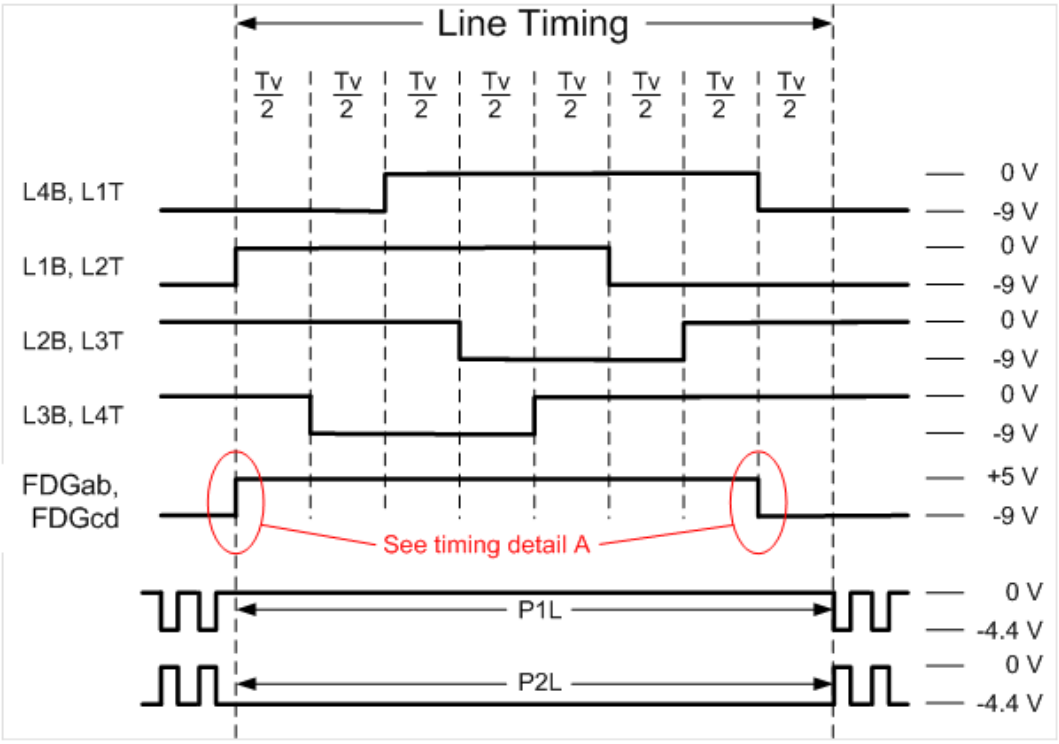


Figure 24: The line timing diagram. See Table 3 for device pin assignments

If the line is to be dumped then clock the FDGab and FDGcd pins as shown. This dumping process eliminates a line of charge and the HCCD does not have to be clocked. To transfer a line from the VCCD to the HCCD without dumping the charge, hold the FDGab and FDGcd pins at a constant -9 V.

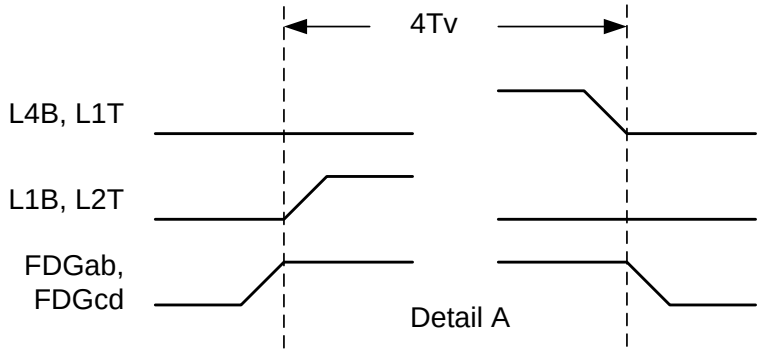


Figure 25: Fast dump gate timing detail A. See Table 3 for device pin assignments

When the VCCD is clocked while the FDGab and FDGcd pins are at +5 V, charge is diverted to a drain instead of transferring to the HCCD. The FDG pins must be at +5 V before the first VCCD timing edge begins its transition. The FDG pin must not begin its transition from +5 V back to -9 V until the last VCCD timing edge has completed its transition.

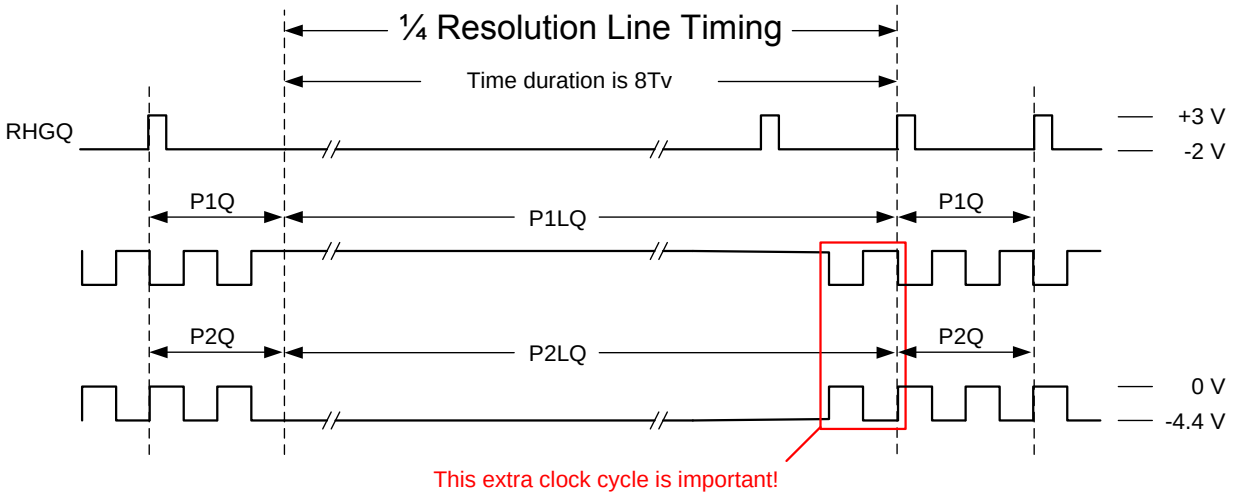


Figure 26: 1/4 resolution line timing diagram. See Table 3 center columns for pin assignments.

The HCCD ¼ resolution timing has one HCCD clock cycle added. This does a one pixel shift of the HCCD before the 2-pixel charge summing starts on the output amplifier. The one pixel shift is necessary because of the odd number (11 pixels) of dummy pixels at the start of the HCCD. Without the one pixel shift the last dark reference columns would be summed with the first photoactive column instead of adding together the first two photoactive columns.

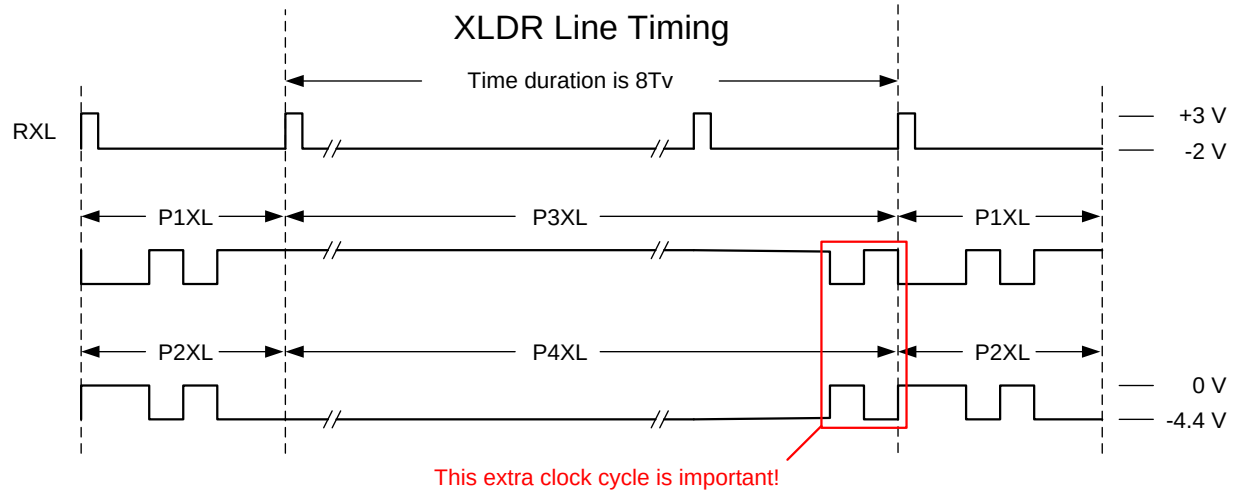


Figure 27: The XDLR line timing diagram. See Table 3 right columns for pin assignments.

Like the ¼ resolution mode, the XLDR timing also sums two pixels on the output amplifier sense node. Therefore it also requires one HCCD clock cycle within the line timing.

Electronic Shutter Timing Diagram

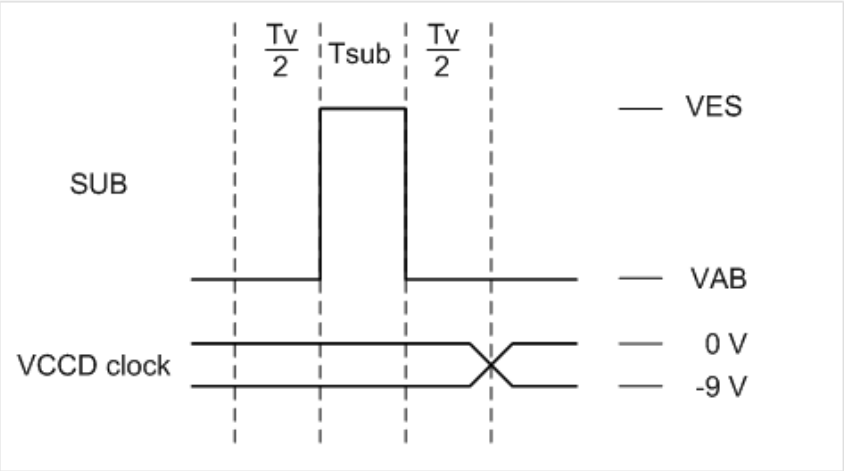


Figure 28: Electronic shutter timing diagram.

The electronic shutter pulse can be inserted at the end of any line of the CCD timing. The HCCD should be empty when pulsing the electronic shutter. A good place for the electronic shutter is just after the last pixel is read out of a line. The VCCD clocks should not resume until at least Tv/2 µs after the electronic shutter pulse has finished. The HCCD clocks can run during the electronics shutter pulse as long as the HCCD does not contain valid image data.

For short exposures less than one line time, the electronic shutter pulse can appear inside the frame timing diagram of Figure 23. Any electronic shutter pulse transition should be Tv/2 away from any VCCD clock transition.

Pixel Timing Diagrams

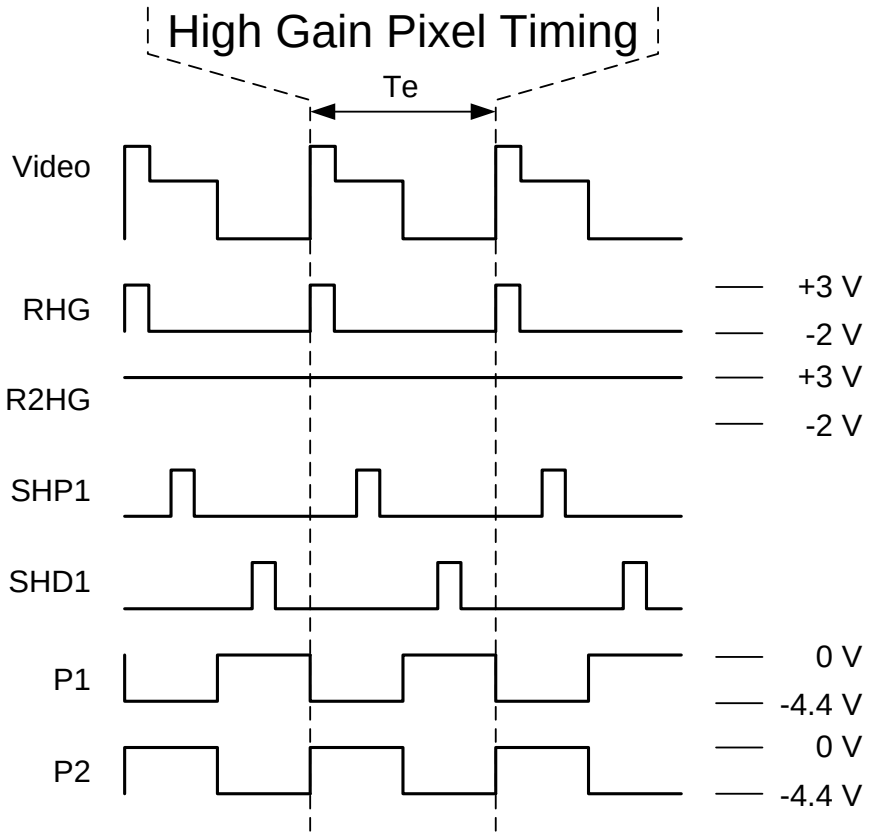


Figure 29: High gain pixel timing. See Table 4 left columns for pin assignments.

Use this pixel timing to read out every pixel at high gain. If the sensor is to be permanently operated at high gain, the R2ab and R2cd pins can be left floating or set to any DC voltage between +3 V and +5 V. They are internally biased to +4.3 V. The SHP1 and SHD1 pulses indicate where the camera electronics should sample the video waveform. The SHP1 and SHD1 pulses are not applied to the image sensor.

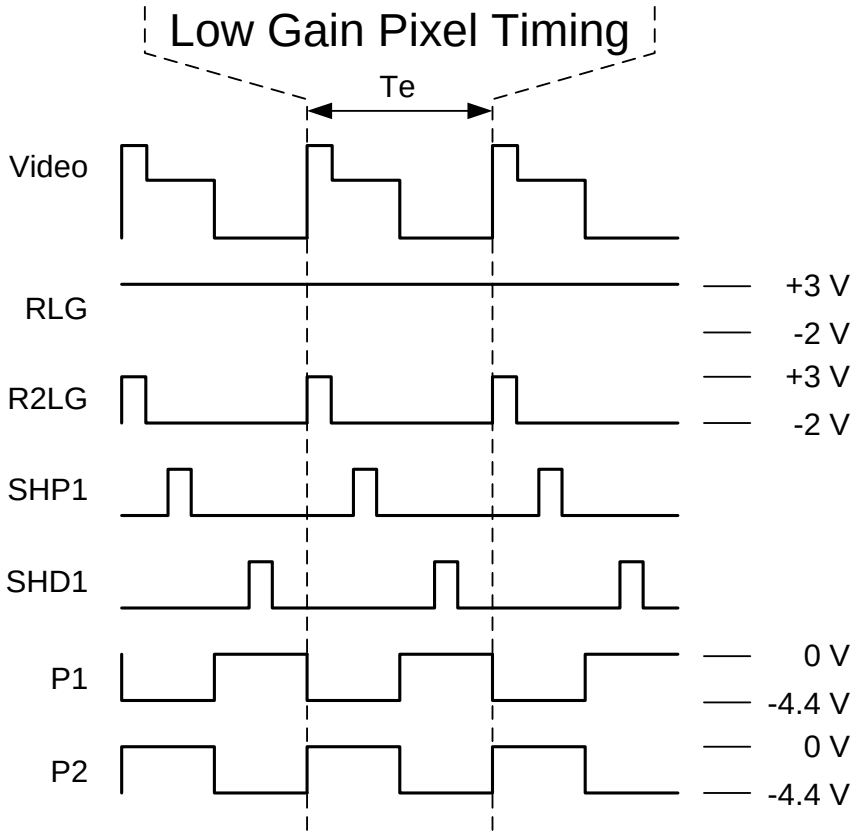


Figure 30: Low gain pixel timing. See Table 4 left columns for pin assignments.

Use this timing to read out every pixel at low gain. If the sensor is to be permanently operated at low gain, the Ra, Rb, Rc, and Rd pins can be set to any DC voltage between +3 V and +5 V. The SHP1 and SHD1 pulses indicate where the camera electronics should sample the video waveform. The SHP1 and SHD1 pulses are not applied to the image sensor.

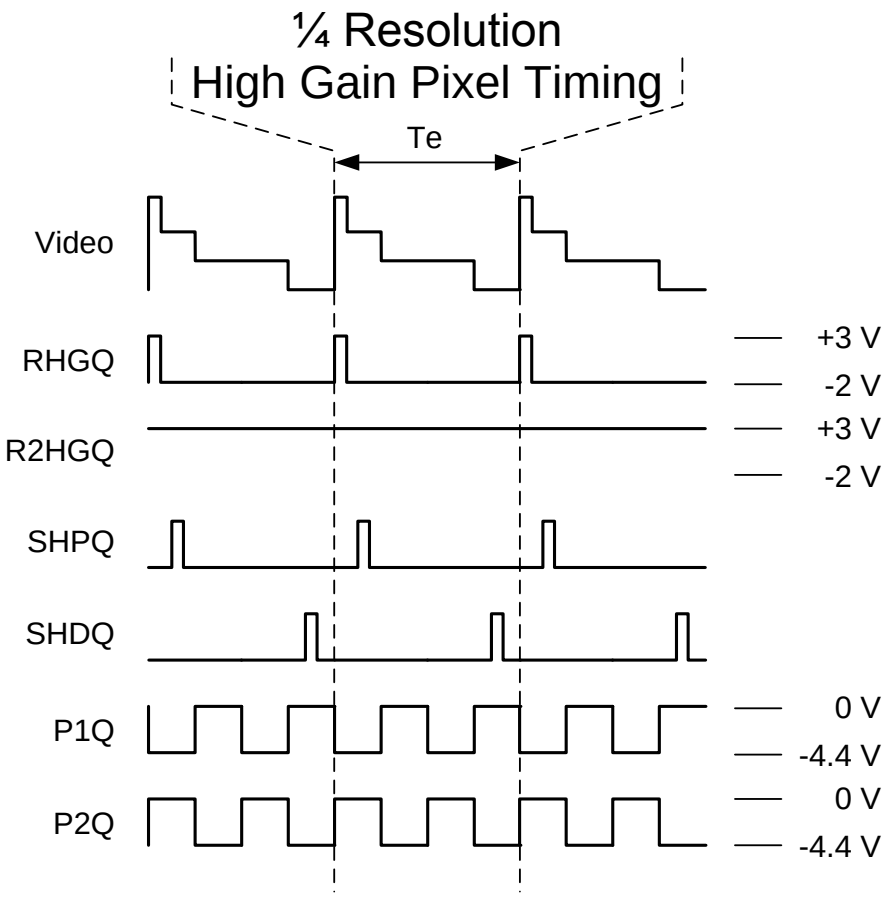


Figure 31: 1/4 resolution high gain pixel timing. See Table 4 center columns for pin assignments.

Use this pixel timing to read out every pixel at high gain. If the sensor is to be permanently operated at high gain, the R2ab and R2cd pins can be left floating or set to any DC voltage between +3 V and +5 V. They are internally biased to +4.3 V. The SHPQ and SHDQ pulses indicate where the camera electronics should sample the video waveform. The SHPQ and SHDQ pulses are not applied to the image sensor.

The Ra, Rb, Rc, and Rd pins are pulsed at half the frequency of the HCCD clocks. This causes two pixels to be summed on the output amplifier sense node. The SHPQ and SHDQ clocks are also half the frequency of the HCCD clocks.

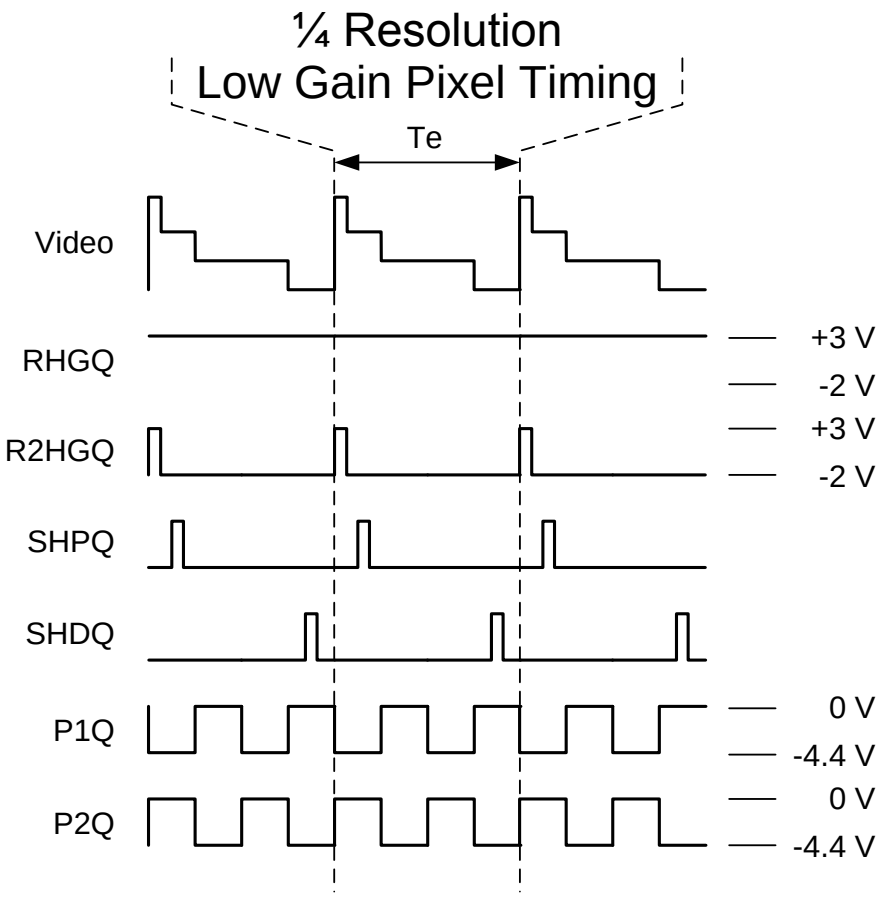


Figure 32: 1/4 resolution low gain pixel timing. See Table 4 center columns for pin assignments.

Use this timing to read out every pixel at low gain. If the sensor is to be permanently operated at low gain, the Ra, Rb, Rc, and Rd pins can be set to any DC voltage between +3 V and +5 V. The SHPQ and SHDQ pulses indicate where the camera electronics should sample the video waveform. The SHPQ and SHDQ pulses are not applied to the image sensor.

The R2ab, and R2cd pins are pulsed at half the frequency of the HCCD clocks. This causes two pixels to be summed on the output amplifier sense node. The SHPQ and SHDQ clocks are also half the frequency of the HCCD clocks.

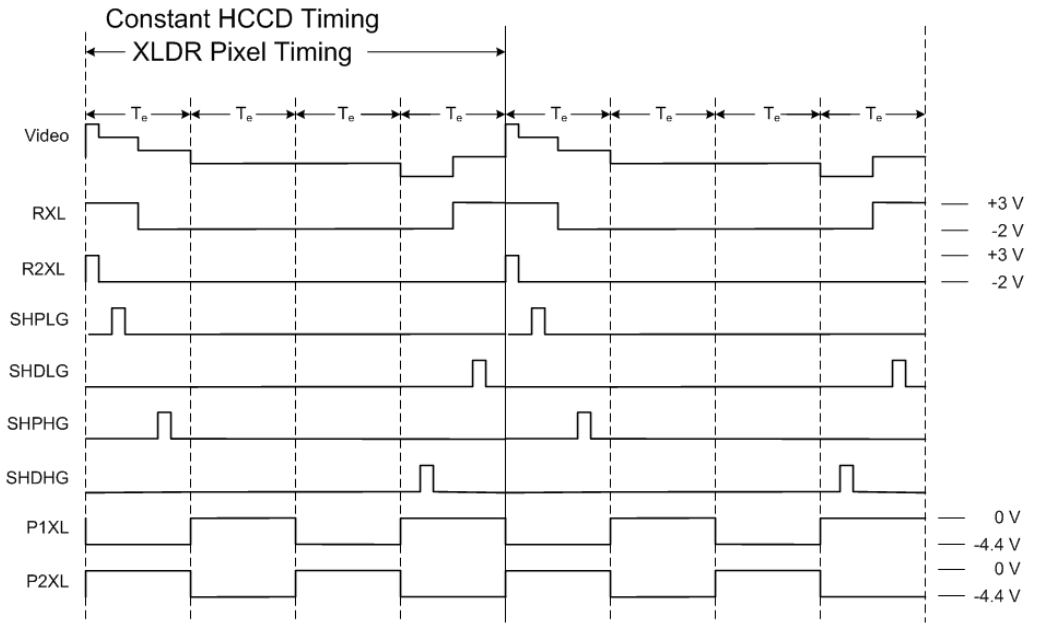


Figure 33: XDLR Timing with Constant HCCD. Operating at 20MHz. See Table 4 right columns for pin assignments.

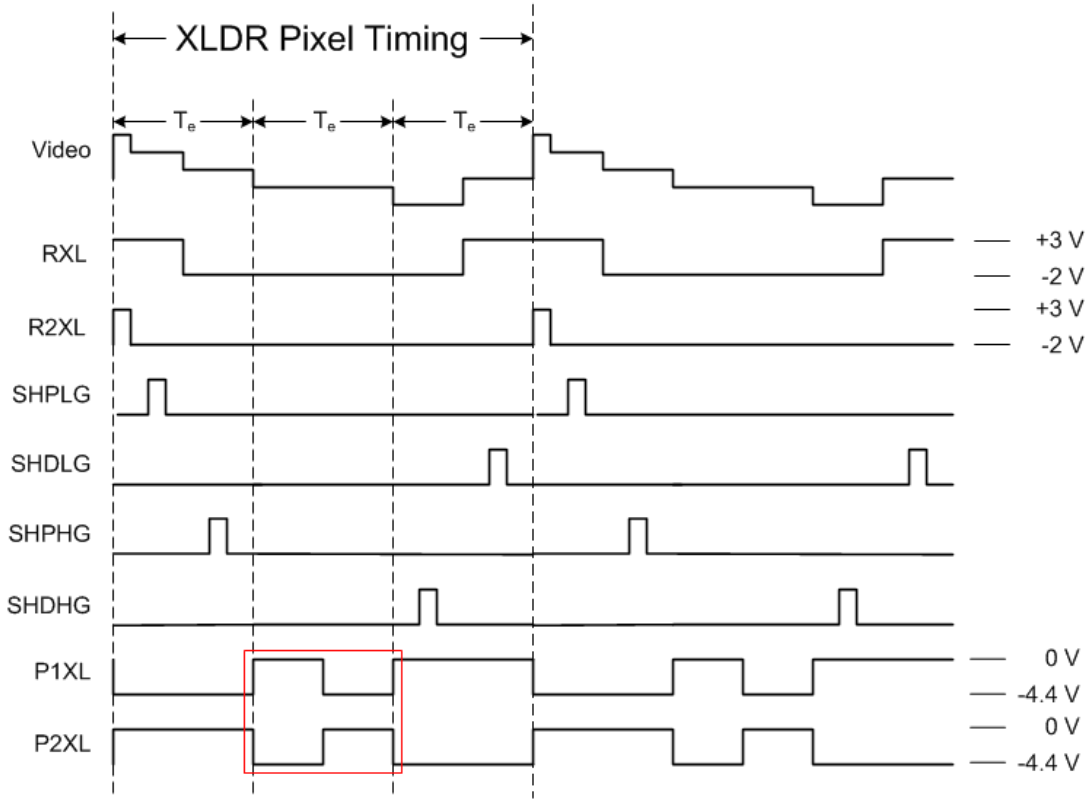


Figure 34: XDLR Timing with Variable HCCD clocking. See Table 4 right columns for pin assignments.

Use this pixel timing to operate the image sensor in the extended linear dynamic range mode (XLDR). This mode requires two sets of analog front end (AFE) signal processing electronics for each output. As shown in Figure 34, one AFE samples the pixel at low gain (SHPLG and SHDLG) and the other AFE samples the pixel at high gain (SHPHG and SHDHG).

Two HCCD pixels are summed on the output amplifier to obtain enough charge to fully use the 82 db dynamic range of the XLDR timing. Combined with two-line VCCD summing, a total of 160,000 electrons of signal (4x 40,000) can be sampled with 12 electrons or less noise. 82 db linear dynamic range is very large. Make certain the camera optics is capable of focusing an 82 db dynamic range image on the sensor. Lens flare caused by inexpensive optics or even dust on the lens will limit the dynamic range.

This timing shows the HCCD in Figure 34, not being clocked at a constant frequency. If this is a problem for the HCCD timing generator, then the HCCD may be clocked at a constant frequency at the expense of about 33% slower frame rate.

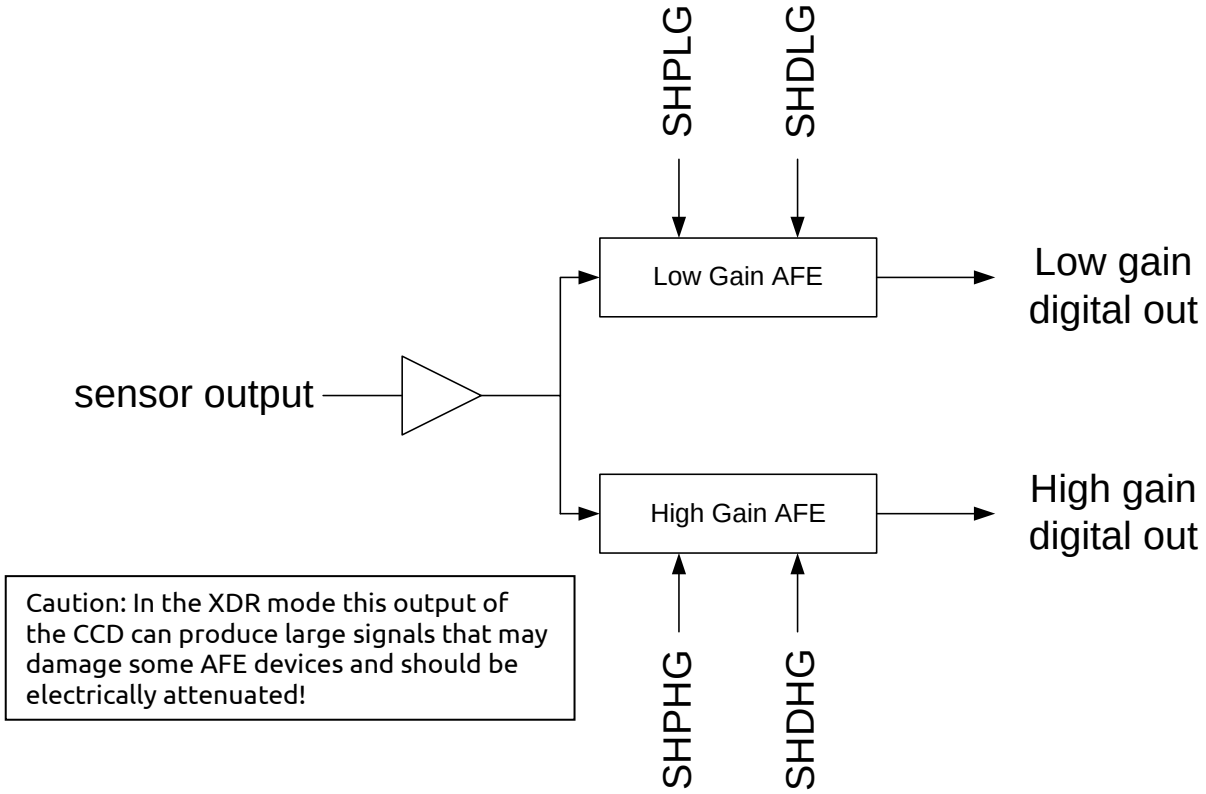


Figure 35: A block diagram showing the AFE connections for XLDR timing.

VCCD Clock Rise and Fall Time

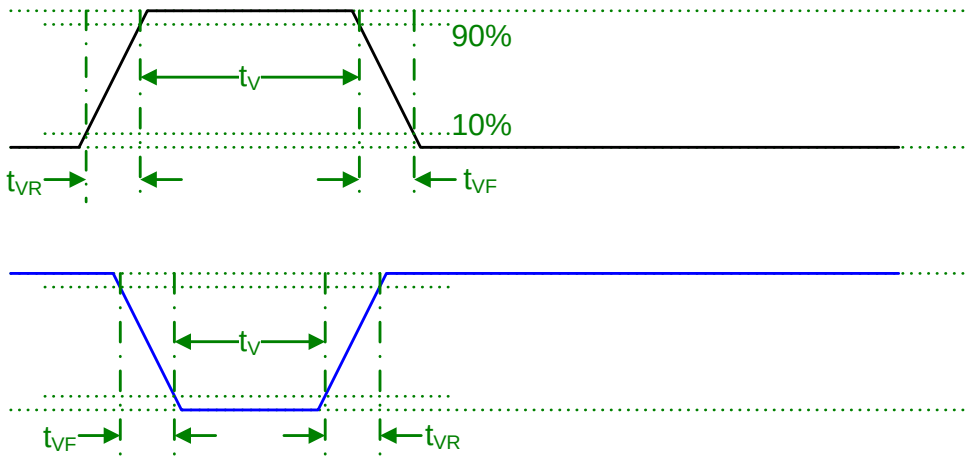


Figure 36: VCCD Clock Rise Time and Fall Time

Storage and Handling

Storage Conditions

Description	Symbol	Minimum	Maximum	Units	Notes
Storage Temperature	T_{ST}	-55	+80	°C	1
Humidity	RH	5	90	%	2

Notes:
1. Long-term storage toward the maximum temperature will accelerate color filter degradation.
2. T=25 °C. Excessive humidity will degrade MTTF.

ESD

1. This device contains limited protection against Electrostatic Discharge (ESD). ESD events may cause irreparable damage to a CCD image sensor either immediately or well after the ESD event occurred. Failure to protect the sensor from electrostatic discharge may affect device performance and reliability.
2. Devices should be handled in accordance with strict ESD procedures for Class 0 (<250V per JESD22 Human Body Model test), or Class A (<200V JESD22 Machine Model test) devices. Devices are shipped in static-safe containers and should only be handled at static-safe workstations.
3. See Application Note *Image Sensor Handling Best Practices* for proper handling and grounding procedures. This application note also contains workplace recommendations to minimize electrostatic discharge.
4. Store devices in containers made of electro-conductive materials.

Cover Glass Care and Cleanliness

1. The cover glass is highly susceptible to particles and other contamination. Perform all assembly operations in a clean environment.
2. Touching the cover glass must be avoided.
3. Improper cleaning of the cover glass may damage these devices. Refer to Application Note *Image Sensor Handling Best Practices*.

Environmental Exposure

1. Extremely bright light can potentially harm CCD image sensors. Do not expose to strong sunlight for long periods of time, as the color filters and/or microlenses may become discolored. In addition, long time exposures to a static high contrast scene should be avoided. Localized changes in response may occur from color filter/microlens aging. For Interline devices, refer to Application Note *Using Interline CCD Image Sensors in High Intensity Visible lighting Conditions*.
2. Exposure to temperatures exceeding maximum specified levels should be avoided for storage and operation, as device performance and reliability may be affected.
3. Avoid sudden temperature changes.
4. Exposure to excessive humidity may affect device characteristics and may alter device performance and reliability, and therefore should be avoided.
5. Avoid storage of the product in the presence of dust or corrosive agents or gases, as deterioration of lead solderability may occur. It is advised that the solderability of the device leads be assessed after an extended period of storage, over one year.

Soldering Recommendations

1. The soldering iron tip temperature is not to exceed 370 °C. Higher temperatures may alter device performance and reliability.
2. Flow soldering method is not recommended. Solder dipping can cause damage to the glass and harm the imaging capability of the device. Recommended method is by partial heating using a grounded 30 W soldering iron. Heat each pin for less than 2 seconds duration.

Mechanical Information

COMPLETED ASSEMBLY

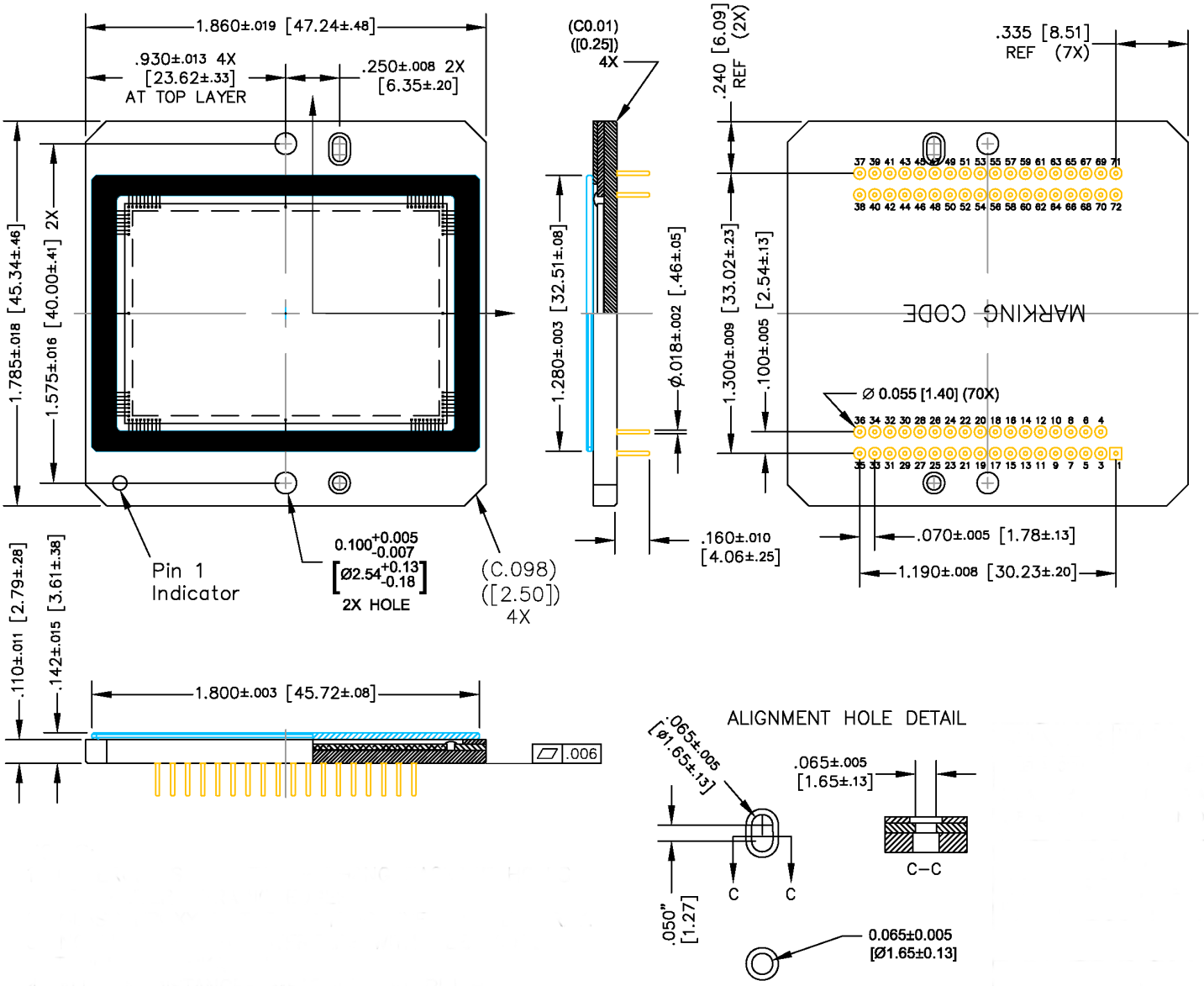


Figure 37: Completed Assembly (1 of 2)

Notes:

1. See Ordering Information for marking code.
2. Cover glass not to overhang package holes or outer ceramic edges
3. Glass epoxy not to extend over image array
4. No materials to interfere with clearance through package holes.
5. Units: IN [MM]

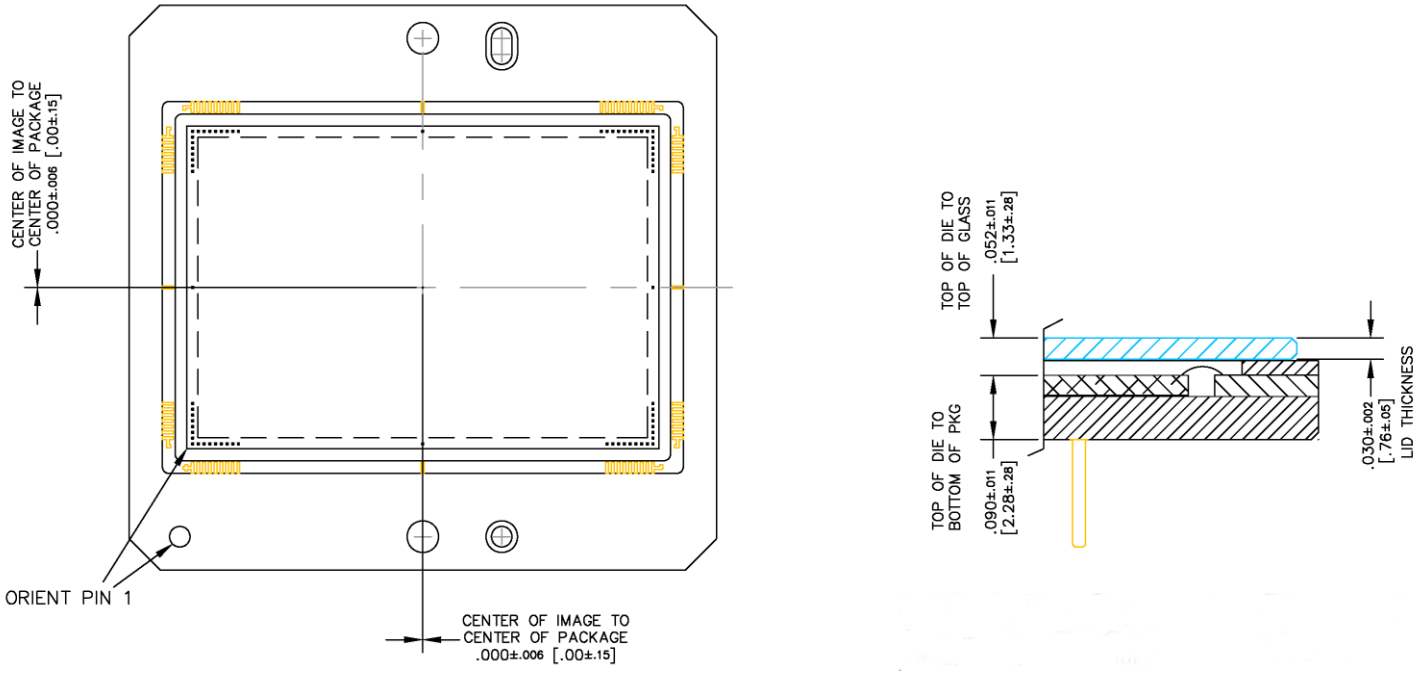


Figure 38: Completed Assembly (2 of 2)

Notes:

1. Units IN [MM]

COVER GLASS

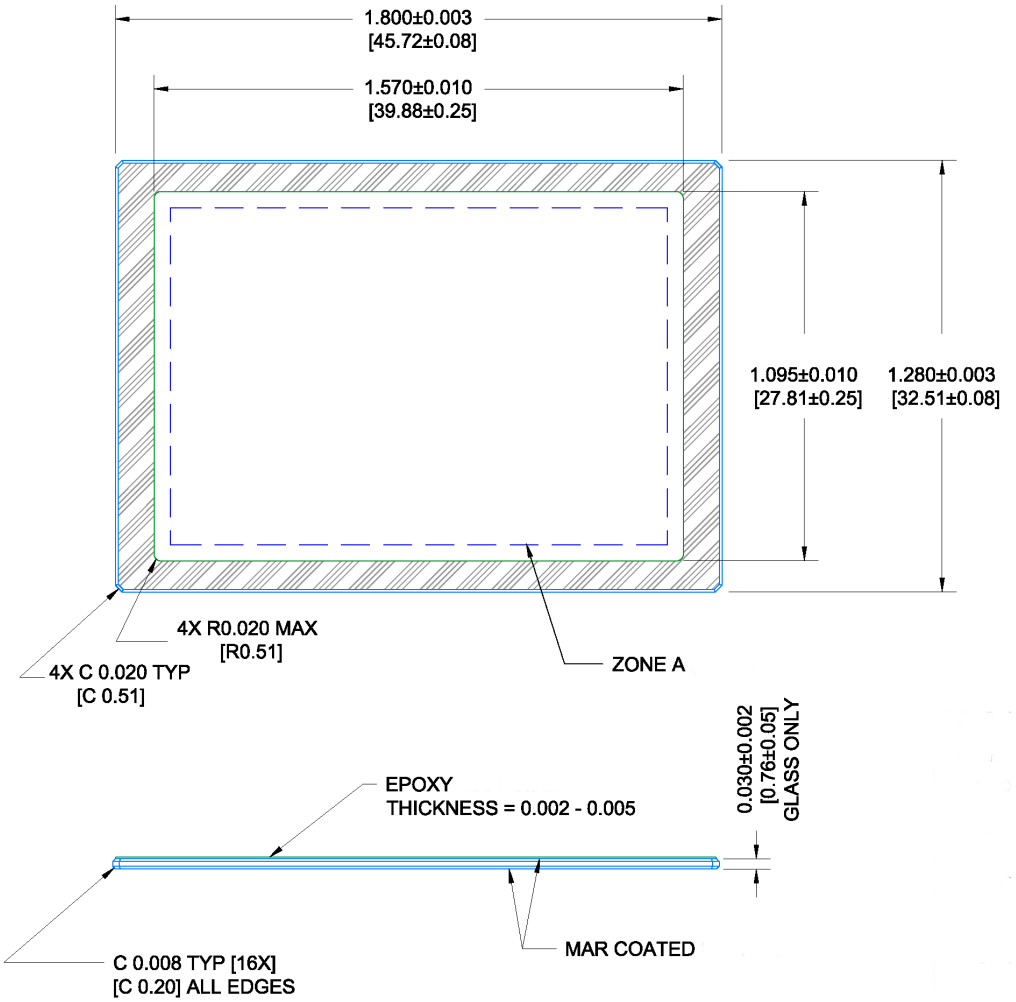


Figure 39: Cover Glass

Notes:

1. Substrate = Schott D263T eco
2. Dust, Scratch, Inclusion Specifocation:
    a. 20 µm Max size in Zone A
    b. Zone A = 1.474 x 1.000 [16.43 x 10.08] Centered
3. MAR coated both sides
4. Spectral Transmission
    a. 350 – 365 nm: T ≥ 88%
    b. 365 – 405 nm: T ≥ 94%
    c. 405 – 450 nm: T ≥ 98%
    d. 450 – 650 nm: T ≥ 99%
    e. 650 – 690 nm: T ≥ 98%
    f. 690 – 770 nm: T ≥ 94%
    g. 770 – 870 nm: T ≥ 88%
5. Units: IN [MM]

COVER GLASS TRANSMISSION

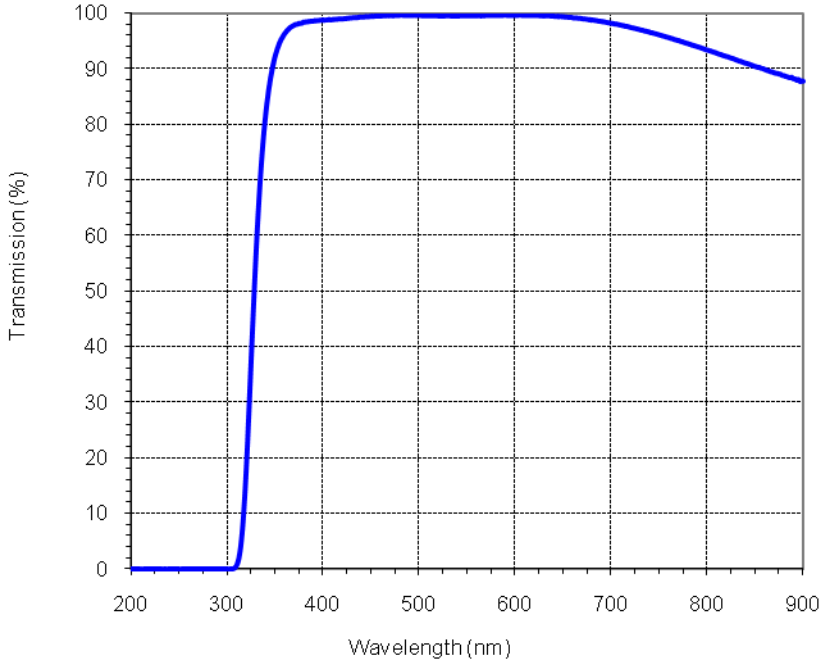


Figure 40: Cover Glass Transmission

Quality Assurance and Reliability

Quality and Reliability

All image sensors conform to the specifications stated in this document. This is accomplished through a combination of statistical process control and visual inspection and electrical testing at key points of the manufacturing process, using industry standard methods. Information concerning the quality assurance and reliability testing procedures and results are available from ON Semiconductor upon request. For further information refer to Application Note *Quality and Reliability*.

Replacement

All devices are warranted against failure in accordance with the *Terms of Sale*. Devices that fail due to mechanical and electrical damage caused by the customer will not be replaced.

Liability of the Supplier

A reject is defined as an image sensor that does not meet all of the specifications in this document upon receipt by the customer. Product liability is limited to the cost of the defective item, as defined in the *Terms of Sale*.

Liability of the Customer

Damage from mishandling (scratches or breakage), electrostatic discharge (ESD), or other electrical misuse of the device beyond the stated operating or storage limits, which occurred after receipt of the sensor by the customer, shall be the responsibility of the customer.

Test Data Retention

Image sensors shall have an identifying number traceable to a test data file. Test data shall be kept for a period of 2 years after date of delivery.

Mechanical

The device assembly drawing is provided as a reference.

ON Semiconductor reserves the right to change any information contained herein without notice. All information furnished by ON Semiconductor is believed to be accurate.

Life Support Applications Policy

ON Semiconductor image sensors are not authorized for and should not be used within Life Support Systems without the specific written consent of ON Semiconductor.

Revision Changes

MTD/PS-1248

Revision Number	Description of Changes
1.0	• Initial Release

PS-0010

Revision Number	Description of Changes
1.0	• Initial release with new document number, updated branding and document template • Updated *Storage and Handling* and *Quality Assurance and Reliability* sections
2.0	• Updated figure "*Output Amplifier – showing dual reset pins*" by adding the second reset pin added to the artwork. • Adjusted description in the *KAI-29050 Compatibility* section for clarity. • Added two figures to show the linear range of the low-gain mode and the high-gain mode. • Made changes to the *Image Performance* section, specifically, to the *Specifications* table. Added expected charge capacity for floating diffusion node, and photodiode non-linearity. • Revised operating levels for RG and RG2. New levels specified are R_L (to designate low level) and R_A (to designate amplitude). The specified low level for RG changes from -2V to -3V. The specified low level for RG changes from -2V to –1.8V. Both signals now specify a 6V swing with respect to the R_L level instead of the implied 5V swing in the previous revisions. • Updated the Vx_L and FDG levels from the current values of -9.0V +/- 0.5V to a new requirement of -8.0V +/- 0.2V. • Updated the VESD level from the current values of -9.0V +/- 0.5V to a new requirement of Vx_L max (-8.2V) to -9.5V min. • Reduced the RD maximum allowed value from 17.5V to 15.5V. • Changed specification for vertical rise time, t_{vr}, and vertical fall time, t_{vf}, to be specified at 5% min and a value of 10% max of the pulse width rather than 1us max. • Update the monochrome QE curve with new measured value. Restate the monochrome QEmax typical performance value from the current 50% value to a new value of 48%. • Update the RGB QE curves with new measured values. Restate the RGB QEmax typical performance values from the current 31, 42%, and 41% values to new values of 32%, 41%, 39% respectively. • Corrected artwork "*Electronic Shutter Timing Diagram*". "VAB +VES" label changed to "VAB". • Provided clarification on note stating "This clock should be held at its high level voltage." For the unused H register for alternate operation. New note states: "This clock should be held at its high level voltage (0V) or held at +5.0V for compatibility with TRUESENSE 5.5 micron Interline Transfer CCD family of products." • In the "*Line Timing Diagram*", changed reference signal label from FDB, FDT to FDGab , FDGcd for consistency. • Changed typographical error in the typical value for "*Maximum Gain Difference Between Outputs*", ΔG, from a typical value of 1% to 10%. • Changed the symbol found in note 2 of the *Timing Requirements and Characteristics* from tV to t_V.
2.1	• Updated branding

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