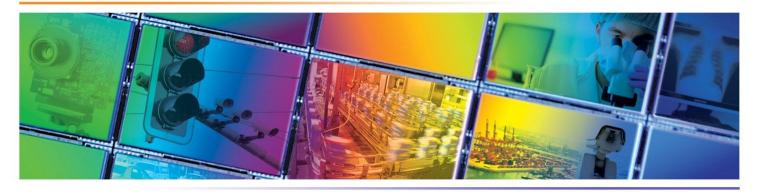


# KAC-12040 IMAGE SENSOR 4000 (H) X 3000 (V) CMOS IMAGE SENSOR



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## **Summary Specification**

## KAC-12040 Image Sensor

#### DESCRIPTION

The KAC-12040 Image Sensor is a high-speed 12 megapixel CMOS image sensor in a 4/3" optical format based on a 4.7 µm 5T CMOS platform. The image sensor features very fast frame rate, excellent NIR sensitivity, and flexible readout modes with multiple regions of interest (ROI). The readout architecture enables use of 8, 4, or 2 LVDS output banks for full resolution readout of 70 frames per second.

Each LVDS output bank consists of up to 8 differential pairs operating at 160 MHz DDR for a 320 Mbps data rate per pair. The pixel architecture allows rolling shutter operation for motion capture with optimized dynamic range or global shutter for precise still image capture.

The image sensor has a pre-configured QFHD (4 x 1080p, 16:9) video mode, fully programmable, multiple ROI for windowing, programmable sub-sampling, and reverse readout (flip and mirror). The two ADCs can be configured for 8-bit, 10-bit, 12-bit or 14-bit conversion and output.

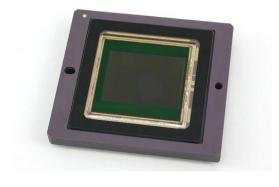
Additional features include interspersed video streams (dual-video), on-chip responsivity calibration, black clamping, overflow pixel for blooming reduction, blacksun correction (anti-eclipse), column and row noise correction, and integrated timing generation with SPI control, 4:1 and 9:1 averaging decimation modes.

#### **F**EATURES

- Global shutter and rolling shutter
- Very fast frame rate
- High NIR sensitivity
- Multiple regions of interest
- Interspersed video streams

#### **APPLICATIONS**

- Machine Vision
- Intelligent Transportation Systems
- Surveillance



Parameter	Typical Value
Architecture	5T Global Shutter CMOS
Resolution	12 megapixels
Aspect Ratio	4:3
Pixel Size	4.7 μm (H) x 4.7 μm (V)
Total Number of Pixels	4224 (H) x 3192 (V)
Number of Effective Pixels	4016 (H) x 3016 (V)
Number of Active Pixels	4000 (H) x 3000 (V)
Active Image Size	18.8 mm (H) x 14.1 mm (V) 23.5 mm (diag.), 4/3" optical format
Master Clock Input Speed	5 MHz to 50 MHz
Maximum Pixel Clock Speed	160 MHz DDR LVDS, 320 Mbps
Number of LVDS Outputs	64 differential pairs
Number of Output Banks	8, 4, ог 2
Frame Rate, 12 MP	1 - 70 fps 10 bits 1 - 75 fps 8 bits
Charge Capacity	16,000 electrons
Quantum Efficiency KAC-12040-CBA KAC-12040-ABA	40%, 47%, 45% (470, 540, 620 nm) 53%, 15%, 10% (500, 850, 900 nm)
Read Noise (at maximum LVDS clock)	3.7 e <sup>-</sup> rms, Rolling Shutter 25.5 e <sup>-</sup> rms, Global Shutter
Dynamic Range	73 dB, Rolling Shutter 56 dB, Global Shutter
Blooming Suppression	>10,000x
Image Lag	1.3 electron
Digital Core Supply	2.0 V
Analog Core Supply	1.8 V
Pixel Supply	2.8 V & 3.5 V
Power Consumption	1.5 W for 12 Mp @ 70 fps 10 bits
Package	267 pin ceramic micro-PGA
Cover Glass	AR Coated, 2-sides

All parameters are specified at T = 40 °C unless otherwise noted



# **Ordering Information**

## KAC-12040 IMAGE SENSOR

Catalog Number	Product Name	Description	Marking Code	
4H2227	KAC-12040-ABA-JD-AA Monochrome, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Standard Grade		KAC-12040-ABA	
4H2229 (1)         KAC-12040-ABA-JD-AE         Monochrome, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Engineering Grad		Monochrome, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Engineering Grade	Serial Number	
4H2231	KAC-12040-CBA-JD-AA       Bayer (RGB) Color Filter Pattern, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Standard Grade		КАС-12040-СВА	
4H2233 (1) KAC-12040-CBA-JD-AE		Bayer (RGB) Color Filter Pattern, micro-PGA Package, Sealed Clear Cover Glass with AR coating(both sides), Engineering Grade	Serial Number	

Notes:

1. Engineering Grade samples might not meet final production testing limits, especially for cosmetic defects such as clusters, but also possibly column and row artifacts. Overall performance is representative of final production parts.

## **EVALUATION SUPPORT**

Catalog Number	Product Name	Description
4H2290	KEK-4H2290-KAC-12040-CB	Evaluation Hardware for KAC-12040 Image Sensor (color). Includes Image Sensor.
4H2291	KEK-4H2291-KAC-12040-AB	Evaluation Hardware for KAC-12040 Image Sensor (monochrome). Includes Image Sensor.
4H2211	Lens Mount Kit	Lens Mount Kit that supports C, CS, and F mount lenses. Includes IR cut-filter for color imaging.

See Application Note Product Naming Convention for a full description of the naming convention used for image sensors. For reference documentation, including information on evaluation kits, please visit our web site at www.truesenseimaging.com.

Please address all inquiries and purchase orders to:

Truesense Imaging, Inc. 1964 Lake Avenue Rochester, New York 14615

Phone: (585) 784-5500 E-mail: <u>info@truesenseimaging.com</u>

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## **Device Description**

### ARCHITECTURE

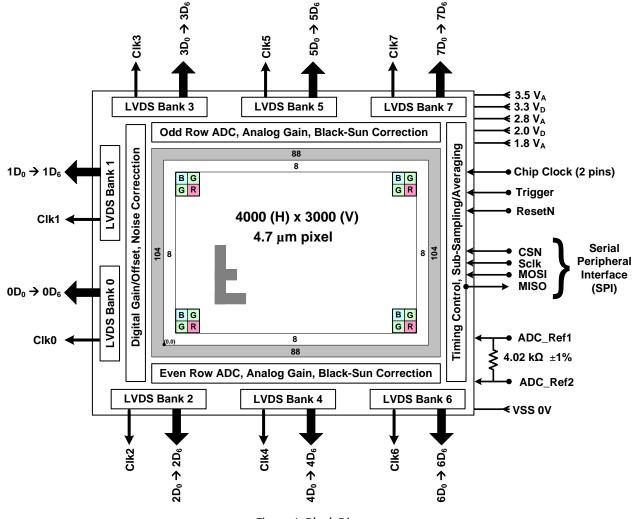


Figure 1: Block Diagram



#### **PHYSICAL ORIENTATION**

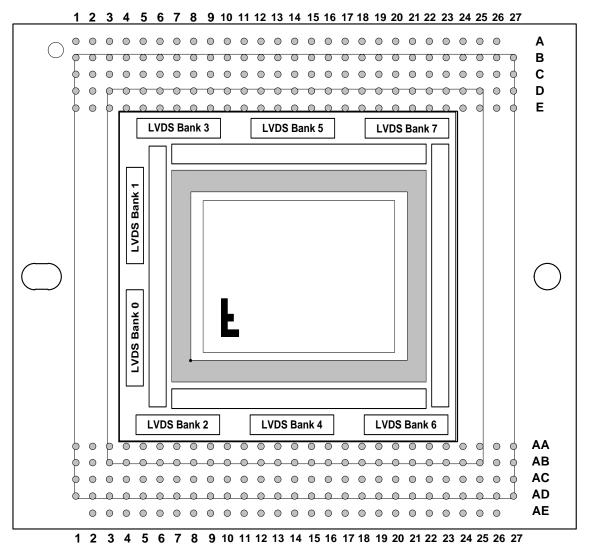


Figure 2: Package Pin Orientation – Top x-ray view

#### Notes:

- 1. The center of the pixel array is aligned to the physical package center.
- 2. The region under the sensor die is clear of pins enabling the use of a heat sink.
- 3. Non-symmetric mounting holes provide orientation and mounting precision.
- 4. Non-symmetric pins prevent incorrect placement in PCB.
- 5. Letter "F" indicator shows default readout direction relative to package pin 1



#### **PRIMARY PIN DESCRIPTION**

Pin	Name	Туре	Description
AB09	RESETN	DI	Sensor reset (0 V = Reset State)
E07	CLK_In1	DI	Sensor Input Clk_In1 (45 – 50 MHz)
D08	CLK_In2	DI	Sensor Input Clk_In2 (connect to Clk1)
AB08	TRIGGER	DI	Trigger input (optional)
AA05	SCLK	DI	SPI Master Clock
AA08	MOSI	DI	SPI Master Output Slave Input
AA07	MISO	DO	SPI Master Input, Slave Output
AA06	CSN	DI	SPI Chip Select (0 V = Selected)
AA14	ADC_Ref1	AO	4.02 k $\Omega$ ±1% resistor between Ref1 & Ref2
AA15	ADC_Ref2	AO	4.02 k $\Omega$ ±1% resistor between Ref1 & Ref2
AB07	MSO	DO	Mechanical Shutter output sync (optional)
AB06	FLO	DO	Flash output sync (optional)
E05	FEN	DO	Frame ENable reference output (optional)
E06	LEN	DO	Line ENable reference output (optional)

Notes:

- 1. DI = Digital Input, DO = Digital Output, AO = Analog Output
- 2. Tie unused DI pins to Ground, NC unused DO pins
- 3. By default Clk\_In2 should equal Clk\_In1 and should be the same source clock.
- 4. The RESETN pin has a 62 k $\Omega$  internal pull-up resistor, so if left floating the chip will not be in reset mode.
- 5. The TRIGGER pin has an internal 100 kΩ pull down resistor. If left floating (and at default polarity) then the sensor state will not be affected by this pin (ie defaults to 'not triggered' mode if floated)
- 6. All of the DI and DO pins nominally operate at  $0 V \rightarrow 2.0 V$  and are associated with the VDD\_DIG power supply.

Name	Voltage	Pins	Description
VDD_LVDS	3.3V D	C04, C05, C23, C24, D04, D24, E04, E24, AA04, AA24, AB04, AB24, AC04, AC05, AC23, AC24	LVDS output supply
VDD_DIG	2.0V D	C18, C19, D18, D19, E18, AA18, AB18, AB19, AC18, AC19, C20, C21, C22, D20, D21, D22, D23, E20, E21, E22, AA20, AA21, AA22, AB20, AB21, AB22, AB23, AC20, AC21, AC22, AB15, E08	Digital core supply
AVDD_HV	3.5V A	C11, D11, E11, AA11, AB11, AC11, C10, D10, E10, AA10, AB10, AC10	Pixel supply 1
Vref_P	2.8V A	C13, D13, E13, AA13, AB13, AC13	Pixel supply 2
AVDD_LV	1.8V A	C17, D16, D17, E17, AA17, AB16, AB17, AC17	Analog low voltage supply
Vpixel_low	0 V	E09	Pixel Supply 3. Combine with VSS for normal operation. Can be pulsed for Extended Dynamic Range Operation
VSS	0 V	C12, C14, D12, D14, E12, AA12, AB12, AB14, AC12, AC14, E15, D15, AA09, A02, A14, A26, B14, C03, C06, C25, D03, D25, E03, E19, E23, E25, AA03, AA19, AA23, AA25, AB03, AB25, AC03, AC06, AC25, AD14, AE02, AE14, AE26	Sensor ground reference
No Connect	NA	A01, AC09, E14, E16, C09, D09, D05, D06, D07, AA16, AB05	Unused and test-only pins. These pins must be floated.

## **Power Pin Description**



## LVDS PIN DESCRIPTION

Pin	Name	Descr	Pin	Name	Descr
E01	1DCLK+	Bank 1 LVDS	C07	3DCLK+	Bank 3 LVDS
E02	1DCLK-	Clock	C08	3DCLK-	Clock
D01	1DATA0+		A07	3DATA0+	
D02	1DATA0-		B07	3DATA0-	
C01	1DATA1+		A08	3DATA1+	
C02	1DATA1-		B08	3DATA1-	
B01	1DATA2+	Bank 1 LVDS	A09	3DATA2+	
B02	1DATA2-		B09	3DATA2-	
A03	1DATA3+		A10	3DATA3+	Bank 3 LVDS
B03	1DATA3-	Data	B10	3DATA3-	Data
A04	1DATA4+		A11	3DATA4+	
B04	1DATA4-		B11	3DATA4-	
A05	1DATA5+		A12	3DATA5+	
B05	1DATA5-		B12	3DATA5-	
A06	1DATA6+		A13	3DATA6+	
B06	1DATA6-		B13	3DATA6-	

Pin	Name	Descr
C15	5DCLK+	Bank 5 LVDS
C16	5DCLK-	Clock
A15	5DATA0+	
B15	5DATA0-	
A16	5DATA1+	
B16	5DATA1-	
A17	5DATA2+	
B17	5DATA2-	
A18	5DATA3+	Bank 5 LVDS
B18	5DATA3-	Data
A19	5DATA4+	
B19	5DATA4-	
A20	5DATA5+	
B20	5DATA5-	
A21	5DATA6+	
B21	5DATA6-	

Pin	Name	Descr
A22	7DCLK+	Bank 7 LVDS
B22	7DCLK-	Clock
A23	7DATA0+	
B23	7DATA0-	
A24	7DATA1+	
B24	7DATA1-	
A25	7DATA2+	
B25	7DATA2-	
B27	7DATA3+	Bank 7 LVDS
B26	7DATA3-	Data
C27	7DATA4+	
C26	7DATA4-	
D27	7DATA5+	
D26	7DATA5-	
E27	7DATA6+	
E26	7DATA6-	

Descr Bank 6 LVDS Clock

Bank 6 LVDS Data

Pin	Name	Descr		Pin	Name	Descr	Pin
AA01	0DCLK+	Bank 0 LVDS		AC07	2DCLK+	Bank 2 LVDS	AC15
AA02	0DCLK-	Clock		AC08	2DCLK-	Clock	AC16
AB01	0DATA0+			AE07	2DATA0+		AE15
AB02	0DATA0-			AD07	2DATA0-		AD15
AC01	0DATA1+			AE08	2DATA1+		AE16
AC02	0DATA1-	Bank 0 LVDS Data		AD08	2DATA1-		AD16
AD01	0DATA2+			AE09	2DATA2+		AE17
AD02	0DATA2-			AD09	2DATA2-		AD17
AE03	0DATA3+			AE10	2DATA3+	Bank 2 LVDS	AE18
AD03	0DATA3-			AD10	2DATA3-	Data	AD18
AE04	0DATA4+			AE11	2DATA4+		AE19
AD04	0DATA4-			AD11	2DATA4-		AD19
AE05	0DATA5+			AE12	2DATA5+		AE20
AD05	0DATA5-			AD12	2DATA5-		AD20
AE06	0DATA6+		AE13	2DATA6+		AE21	
AD06	0DATA6-			AD13	2DATA6-		AD21

Pin	Name	Descr	Pin	N
AC15	4DCLK+	Bank 4 LVDS	AE22	6
AC16	4DCLK-	Clock	AD22	6
AE15	4DATA0+		AE23	6[
AD15	4DATA0-		AD23	6
AE16	4DATA1+		AE24	6[
AD16	4DATA1-		AD24	6
AE17	4DATA2+		AE25	60
AD17	4DATA2-		AD25	6[
AE18	4DATA3+	Bank 4 LVDS	AD26	60
AD18	4DATA3-	Data	AD27	6[
AE19	4DATA4+		AC26	6D
AD19	4DATA4-		AC27	6D
AE20	4DATA5+		AB26	6D
AD20	4DATA5-		AB27	60
AE21	4DATA6+		AA26	6D
AD21	4DATA6-		AA27	6D

#### Notes:

- 1. All LVDS Data and Clock lines must be routed with 100  $\Omega$  differential transmission line traces.
- 2. All the traces for a single LVDS Bank should be the same physical length to minimize skew between the clock and data lines.
- 3. In 2 Bank mode, only LVDS banks 0 and 1 are active.
- 4. In 4 Bank mode, only LVDS bank 0, 1, 2, and 3 are active.
- 5. Float the pins of unused LVDS Banks to conserve power.
- 6. Unused pins in active banks (due to ADC bit depth <14) are automatically tri-stated to save power, but these can also be floated.



# Imaging Performance

## **TYPICAL OPERATION CONDITIONS**

Unless otherwise noted, the Imaging Performance Specifications are measured using the following conditions.

Description	Condition	Notes
Light Source	Continuous red, green and blue LED illumination	1
Temperature	Measured die temperature: 40 °C and 27 °C	
Integration Time	16.6 msec (1400d LL, register 0201h)	
Readout Mode	Dual-Scan, Global Shutter, 320 MHz PLL2	
Clamps	Column/Row Noise Correction active, Frame Black Level Clamp active	
ADC Bit Depth	10 bit	
Analog Gain	Unity gain or referred back to unity gain.	

Notes:

1. For monochrome sensor, only green LED used.

## **PERFORMANCE SPECIFICATIONS ALL CONFIGURATIONS**

Description	Symbol	Min.	Nom.	Max.	Units	Sampling Plan	Temperature Tested At (°C)	Test	Notes
Photodiode Charge Capacity	PNe	-	16	-	ke <sup>-</sup>	Die	27, 40	16	
Read Noise	ne-T	-	3.7 RS 25.5 GS	-	e <sup>-</sup> rms	Die	27	8	
Total Pixelized Noise		-	4.5 RS 28.3 GS	-	e <sup>-</sup> rms	Die	27	19	
Dynamic Range	DR	-	73 RS 56 GS	-	dB	Die	27		3
Column Noise	Cn	-	0.6 RS 3.0 GS	-	eīrms	Die	27	9	5
Row Noise	Rn	-	1.0 RS 5.0 GS	-	e <sup>-</sup> rms	Die	27	10	6
Dark Field Local Non-Uniformity Floor	DSNU_flr	-	3.0 RS 21 GS	-	e <sup>-</sup> rms	Die	27 & 40	1	4
Bright Field Global Photoresponse Non- Uniformity	PRNU_1	-	1.5	-	%rms	Die	27, 40	2	1
Bright Field Global Peak to Peak Photoresponse Non-Uniformity	PRNU_2	-	6.5	-	%рр	Die	27, 40	3	1
Maximum Photoresponse Nonlinearity	NL	-	6.3	-	%	Die	27, 40	11	2
Maximum Gain Difference Between Outputs	ΔG	-	0.3	-	%	Die	27, 40	12	7
Photodiode Dark Current	Ipd	-	4.6	70	e/p/s	Die	40	13	8
Storage Node Dark Current	Ivd	-	1200	5000	e/p/s	Die	40	14	4
Image Lag	Lag	-	1.3	10	e	Design	27, 40	15	
Black-Sun Anti-Blooming	Xab	-	12 >10,000	-	W/cm <sup>2</sup> xIlumSat	Design	27	7	13
Parasitic Light Sensitivity	PLS	-	730	-	-	Design	27	6	9
Dual-Video WDR		-	140 RS 120 GS	-	dB	Design	27		10, 11
Pulsed Pixel WDR (GS only)		-	100	-	dB	Design	27		12, 11

RS = Rolling Shutter operation mode, GS = Global Shutter operation mode



## KAC-12040-ABA Configuration (Monochrome)

Description		Symbol	Wavelength (nm)	Min.	Nom.	Max	Units	Sampling Plan	Temperature Tested At (°C)	Test
Peak Quantum Efficiency	Green NIR1 NIR2	QE <sub>max</sub>	550 850 900	-	53 15 10	-	%	Design	27	
Responsivity				-	84	-	ke <sup>-</sup> Lux * s	Design	27	20
Responsivity				-	7.0	-	$\frac{V}{Lux * s}$	Design	27	21

## KAC-12040-CBA Configuration (Bayer RGB)

Description		Symbol	Wavelength (nm)	Min.	Nom.	Max	Units	Sampling Plan	Temperature Tested At (°C)	Test
Peak Quantum Efficiency	Green NIR1 NIR2	QE <sub>max</sub>	470 540 620 850 900	-	40 47 45 15 10	-	%	Design	27	
Responsivity			Blue Green Red	-	17 35 38	-	$\frac{ke^-}{Lux * s}$	Design	27	20
Responsivity			Blue Green Red	-	1.4 2.9 3.2	-	$\frac{V}{Lux * s}$	Design	27	21

#### Notes:

- 1. Measured per color, worst of all colors reported
- 2. Value is over the range of 10% to 90% of photodiode saturation, Green response used.
- 3. Uses 20LOG(PNe/ ne-T)
- 4. Photodiode dark current made negligible
- 5. Column Noise Correction active
- 6. Row Noise Correction active
- 7. Measured at ~70% illumination
- 8. Storage node dark current made negligible
- 9. GSE (Global Shutter Efficiency) = 1-1/PLS
- 10. Min vs Max integration time at 30 fps
- 11. WDR measures expanded exposure latitude from linear mode DR
- 12. Min/Max responsivity in a 30 fps image
- 13. Saturation Illumination referenced to a 3 line time integration.



# **Typical Performance Curves**

## **QUANTUM EFFICIENCY**

#### Monochrome with Microlens

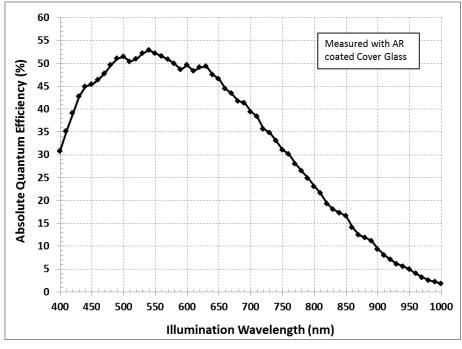


Figure 3: Monochrome QE (with Microlens)

## Color (Bayer RGB) with Microlens

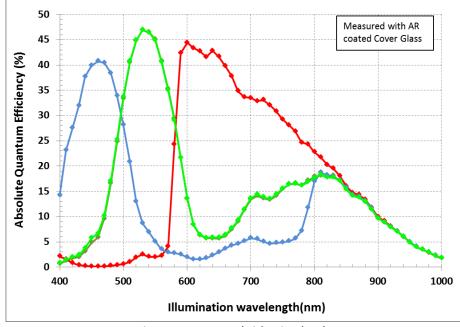


Figure 4: Bayer QE (with Microlens)



## **ANGULAR QUANTUM EFFICIENCY**

For the curves marked "Horizontal", the incident light angle is varied along the wider array dimension.

For the curves marked "Vertical", the incident light angle is varied along the shorter array dimension.

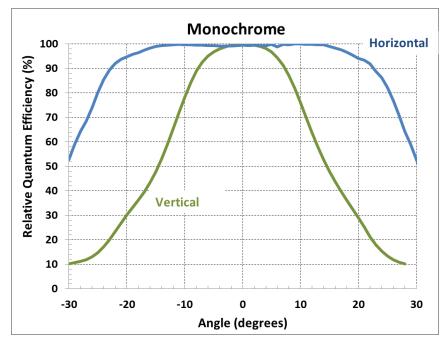


Figure 5: Monochrome Relative Angular QE (with Microlens)

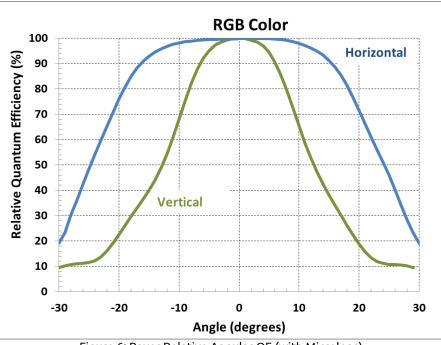


Figure 6: Bayer Relative Angular QE (with Microlens)



## **DARK CURRENT VERSUS TEMPERATURE**

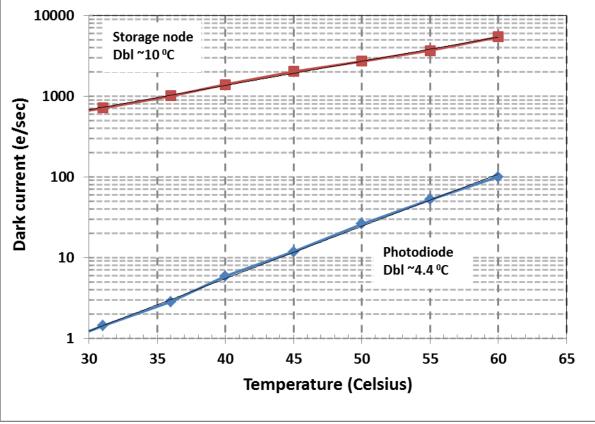


Figure 7: Dark Current vs Temperature

Note: "Dbl" denotes an approximate doubling temperature for the dark current for the displayed temperature range.



## **POWER VS FRAME RATE**

The most effective method to use the maximum PLL2 speed (313 -> 320 MHz) and control frame rate with minimum Power and maximum image quality is to adjust Vertical Blanking. (register 01F1h). Unnecessary chip operations are suspended during Vertical Blanking conserving significant power consumption and also minimizing the image storage time on the storage node when in Global Shutter Operation.

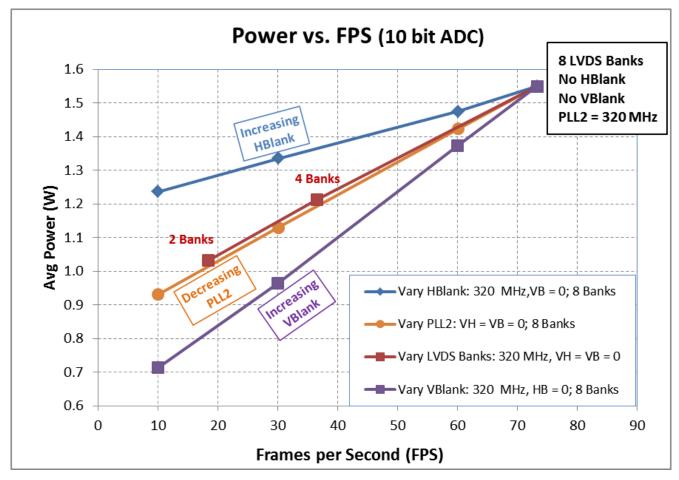


Figure 8: Power vs Frame Rate, 10 bit mode

#### Notes:

1. The LVDS clock is ½ the PLL2 clock speed



#### Power and Frame Rate vs ADC Bit depth

Increasing the ADC bit depth impacts the frame rate by changing the ADC conversion time. The following figure shows the power and Frame rate range for several typical cases.

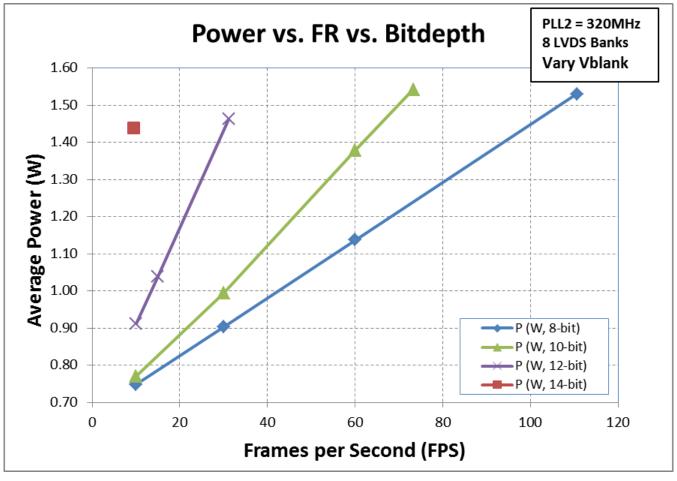


Figure 9: ADC Bit Depth impact on Frame Rate and Power



## **Defect Definitions**

## **OPERATION CONDITIONS FOR DEFECT TESTING**

Description	Condition	Notes
Operational Mode	10 bit ADC, 8 LVDS outputs, Global Shutter and Rolling Shutter modes, Dual-Scan, Black Level Clamp on, Column/Row Noise Correction on, 1x Analog Gain, 1x Digital Gain	
Pixels Per Line	4000	
Lines Per Frame	3000	
Line Time	8.7 µsec	
Frame Time	13.9 msec	
Photodiode Integration Time	33 msec	
Storage Readout Time	13.9 msec	
Temperature	40 °C and 29 °C	
Light Source	Continuous red, green and blue LED illumination	1
Operation	Nominal operating voltages and timing, PLL1 = 320 MHz, PLL2 = 320 MHz, Wafer Test	

Notes:

1. For monochrome sensor, only the green LED is used.

## **DEFECT DEFINITIONS FOR TESTING**

Description	Definition	Limit	Test	Notes	
Dark Field Defective Pixel	30 °C RS: Defect ≥ 20 dn GS: Defect ≥ 180 dn	120	4	3, 4	
Bright Field Defective Pixel	Defect ≥ ±12% from local mean			5	1, 4
Cluster Defect	A group of 2 to 10 contiguous de adjacent defects horizontally	fective pixels, but no more than 3	22		2
Column/Row Major Defect	A group of more than 10 contigu column or row	ous defective pixels along a single	0		
Dark Field Faint Column/Row Defect	RS: 3 dn threshold GS: 10 dn threshold	0	17		
Bright Field Faint Column/Row Defect	RS: 12 dn threshold GS: 18 dn threshold	0	18		

Notes:

RS = Rolling Shutter, GS = Global Shutter

- 1. For the color devices, all bright defects are defined within a single color plane, each color plane is tested
- 2. Cluster defects are separated by no less than two good pixels in any direction.
- 3. Rolling Shutter Dark Field points are dominated by photodiode integration time, Global Shutter Dark Field defects are dominated by the readout time.
- 4. The net sum of all bright and dark field pixel defects in rolling and global shutter are combined and then compared to the test limit

## **DEFECT MAP**

The defect map supplied with each sensor is based upon testing at an ambient (29 °C) temperature. All defective pixels are reference to pixel (0, 0) in the defect maps. See Figure 10 for the location of pixel (0, 0).

#### KAC-12040 Image Sensor

### **ON Semiconductor®**



## **Test Definitions**

### **TEST REGIONS OF INTEREST**

Image Area ROI:	Pixel (0, 0) to Pixel (4015, 3015)
-----------------	------------------------------------

Active Area ROI: Pixel (8, 8) to Pixel (3999, 2999)

Center ROI: Pixel (1958, 1458) to Pixel (2057, 1557)

Only the Active Area ROI pixels are used for performance and defect tests.

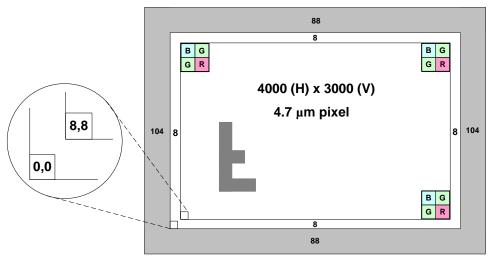


Figure 10: Regions of Interest

## **TEST DESCRIPTIONS**

### 1) Dark Field Local Non-Uniformity Floor (DSNU\_flr)

This test is performed under dark field conditions. A 4 frame average image is collected. This image is partitioned into 300 sub-regions of interest, each of which is 200 by 200 pixels in size. For each sub-region the standard deviation of all its pixels is calculated. The dark field local non-uniformity is the largest standard deviation found from all the sub regions of interest. Units: e<sup>-</sup> rms (electrons rms)

### 2) Bright Field Global Photoresponse Non-Uniformity (PRNU\_1)

The sensor illuminated to 70% of saturation (~700 dn). In this condition a 4 frame average image is collected. From this 4 frame average image a 4 frame average dark image is subtracted. The Active Area Standard Deviation is the standard deviation of the resultant image and the Active Area Signal is the average of the resultant image.

$$PRNU_1 = 100 * \left( \frac{Active Area Standard Deviation}{Active Area Signal} \right)$$
 Units: %rms

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## 3) Bright Field Global Peak to Peak Non-Uniformity (PRNU\_2)

This test is performed with the sensor uniformly illuminated to 70% of saturation (~700 dn), a 4 frame average image is collected and a 4 frame averaged dark image is subtracted. The resultant image is partitioned into 300 sub regions of interest, each of which is 200 by 200 pixels in size. The average signal level of each sub regions of interest (sub-ROI) is calculated.

The highest sub-ROI average (Maximum Signal) and the lowest sub-ROI average (Minimum Signal) are then used in the following formula to calculate PRNU\_2.

 $PRNU_2 = 100 * \frac{Maximum Signal - Minimum Signal}{Active Area Signal} \qquad Units: %pp$ 

## 4) Dark Field Defect Test

This test is performed under dark field conditions. The sensor is partitioned into 300 sub regions of interest, each of which is 128 by 128 pixels in size. In each region of interest, the median value of all pixels is found. For each region of interest, a pixel is marked defective if it is greater than or equal to the median value of that region of interest plus the defect threshold specified in the Defect Definition Table section.

## 5) Bright Field Defect Test

This test is performed with the imager illuminated to a level such that the output is at approximately 700 dn. The average signal level of all active pixels is found. The bright and dark thresholds are set as:

Dark defect threshold = Active Area Signal \* threshold

Bright defect threshold = Active Area Signal \* threshold

The sensor is then partitioned into 300 sub regions of interest, each of which is 128 by 128 pixels in size. In each region of interest, the average value of all pixels is found. For each region of interest, a pixel is marked defective if it is greater than or equal to the median value of that region of interest plus the bright threshold specified or if it is less than or equal to the median value of that region of interest minus the dark threshold specified.

Example for bright field defective pixels:

- Average value of all active pixels is found to be 700 dn
- Lower defect threshold: 700 dn \* 12 % = 84 dn
- A specific 128 x 128 ROI is selected:
  - Median of this region of interest is found to be 690 dn.
  - Any pixel in this region of interest that is  $\leq$  (690 84 dn) in intensity will be marked defective.
  - Any pixel in this region of interest that is  $\geq$  (690 84 dn) in intensity will be marked defective
- All remaining 299 sub regions of interest are analyzed for defective pixels in the same manner.



#### 6) Parasitic Light Sensitivity (PLS)

Parasitic Light Sensitivity is the ratio of the light sensitivity of the photodiode to the light sensitivity of the storage node in Global Shutter. There is no equivalent distortion in Rolling Shutter. A low PLS value can provide distortion of the image on the storage node by the scene during readout.

 $PLS = \frac{Photodiode \ Responsivity}{Storage \ Node \ Responsivity} \qquad (unitless \ ratio)$ 

GSE (Global Shutter Efficiency) is a related unit.  $GSE = \left(1 - \frac{1}{PLS}\right) \%$ 

Detailed method: Photodiode Responsivity:

The sensor is set in global shutter serial mode (integration time not overlapping readout) and the FLO signal is used to control a 550 nm normal incident (or large f# focused) illumination source so that the sensor is illuminated only during photodiode integration time (not illuminated during readout time). The integration time is not critical but should be large enough to create a measurable mean during this time. A 16 frame-average illuminated photodiode image is recorded. A 16 frame-average dark frame using the same sensor settings is captured and is subtracted from the illuminated image.

Detailed method: Storage Node Responsivity:

The sensor is set to a special characterization mode where the PD signal is discarded and does not impact the storage node. A long total frame time (storage node exposure time) is used to increase the storage node signal. A 16 frame-average dark frame is captured. The sensor is illuminated by the same 550nm incident light source used for the photodiode responsivity. A 16 frame-average illuminated photodiode image is recorded; the dark frame image is subtracted from this. The integration time is not critical but should be set such that a significant response is detected, typically several orders of magnitude greater than the photodiode integration time.

#### 7) Black-Sun Anti-blooming

A typical CMOS image sensor has a light response profile that goes from 0 dn to saturation (1023 dn for KAC-12040 in 10 bit ADC mode) and, with enough light, back to 0 dn. The sensor reaching 0 dn at very bright illumination is often called the "Black-sun" artifact and is undesirable. Black-sun artifact is typically the dominant form of anti-blooming image distortion. For the KAC-12040 the Black-sun artifact threshold is measured at the onset of saturation distortion, not at the point where the output goes to 0 dn. To first order the onset of black-sun artifact for the KAC-12040 is not proportional to the integration time or readout time.

The sensor is placed in the dark at unity gain and illuminated with a 532 nm laser with the intensity of about 26 W/cm<sup>2</sup> at the center of the sensor. The laser is strong enough to make the center of the laser spot below 1020 dn without any ND filters. ND filters are added to adjust the laser intensity until the signal in the region at the center of the spot increases to >1020 dn.

This illumination intensity at this ND filter is recorded (W/cm<sup>2</sup>) as the Black-Sun Anti-blooming. The 'xIlumSat' unit is calculated using and integration time of 100 µsec.

Exposing the sensor to very strong illumination for extended periods of time will permanently alter the sensor performance in that localized region.



#### 8) Read Noise

This test is performed with no illumination and one line of integration time. The read noise is defined as one standard deviation of the frequency histogram containing the values of all pixels after the excessively deviant pixels (± three standard deviations) are removed.

#### 9) Column Noise

After all rows are averaged together. Shading (low frequency change wrt column address) is removed. A frequency histogram is constructed of the resulting column values. The column noise is the standard deviation of the frequency histogram of the column values.

#### 10) Row Noise

All columns are averaged together. Shading (low frequency change wrt row address) is removed. A frequency histogram is constructed of the resulting row values. The row noise is the standard deviation of the frequency histogram of the row values.

#### 11) Maximum Photoresponse Non-Linearity

The photoresponse nonlinearity is defined as the deviation from the best fit of the sensor response using 70% of saturation and zero signal as the reference points. The different signal levels are determined by varying the integration time. The sensor saturation level is (1023-dark offset). The dark offset is subtracted from the image for the following Mavg and Lavg.

- The integration time is varied until the integration time required to reach the 70% saturation is determined. Mavg = the active array mean at the 70% saturation integration time.
- The integration is set to 1/14 (5% exposure point). Lavg = meant at the 5% exposure point
- PRNL (@ 5% saturation) =( (Lavg/Mavg) \* (14/1) -1) \* 100

#### 12) Maximum Gain Difference Between Outputs

The sensor contains two ADC and four channels of analog data in its highest frame rate configuration. The sensor is factory calibrated to reduce the gain differences between the channels. The gain variations are manifest as a row oriented pattern where every other row uses a different ADC. Using triple scan read out mode, an additional two analog channels are introduced resulting in a four row pattern. With one channel ('Top Ping') used as the reference, the residual gain difference is defined as:

( (Bottom Ping Row Average/Top Ping Row Average) -1 ) \* 100

((Top Pong Row Average/Top Ping Row Average) -1) \* 100

( (Bottom Pong Row Average/Top Ping Row Average) -1 ) \* 100



#### 13) Photodiode Dark Current

The photodiode dark current is measured in rolling shutter read out mode using 105 msec integration time and an analog gain = 8. The value is converted to electrons/pix/sec using the formula:

Photodiode dark current = average signal (DN) \* el-per-DN (gain=8) / .105 seconds

where 'average signal (DN)' is the average of all pixels in the sensor array, and 'el-per-DN(gain=8)' is measured on each sensor using the photon transfer method.

#### 14) Storage Node Dark Current

The storage node dark current is measured in global shutter read out mode using a special timing mode to prevent the photodiode dark current from being transferred to the storage node. In global shutter mode, the integration time of the storage node is the time it takes to read out a frame. The sensor analog gain is set to 2:

Storage node dark current = average signal (DN) \* el-per-DN (gain=2) / .0138 seconds

where 'average signal (DN)' is the average of all pixels in the sensor array and 'el-per-DN(gain=2)' is measured on each sensor using the photon transfer method.

#### 15) Lag

Lag is measured as the number of electrons left in the photodiode after readout when the sensor is illuminated at 70% of Photodiode Charge Capacity.

Analog gain is set to 8. With no illumination a 64 average dark image is recorded (Dark\_ref). The 'el-per-DN' is measured using the photon transfer method.

Illumination is adjusted blink every other frame such that the mean image output is 70% of the Photodiode Charge Capacity for even frames, and with no illumination for odd frames. A 64 frame average of Odd Dark Frames is recorded as Dark\_Lag.

Lag = (Dark\_Lag - Dark\_Ref) \* el-per-DN' Units: electrons rms

#### 16) Photodiode Charge Capacity

The sensor analog gain is reduced to <1 to prevent ADC clipping at 1023 dn. The 'el-per-DN' is measured using the photon transfer method. The sensor is illuminated at a light level ~1.5x the illumination at which the pixel output no longer linearly changes with illumination level. The Photodiode Charge Capacity is equal to the average signal (DN) \* el-per-DN. Units: electrons rms.

#### 17) Dark Field Faint Column/Row Defect

A 4 frame average, no illumination image is acquired at one line time of integration. Major defective pixels are removed (> 5 Sigma). All columns or rows are averaged together. The average of the local ROI of 128 columns or rows about the column/row being tested is determined. Any columns/rows greater than the local average by more than the threshold are identified.



#### 18) Bright Field Faint Column/Row Defect

A 4 frame average, 70% illumination image is acquired at one line time of integration. Major defective pixels are removed (> 5 Sigma). All columns or rows are averaged together. The average of the local ROI of 128 columns or rows about the column/row being tested is determined. Any columns/rows greater than the local average by more than the threshold are identified.

#### 19) Total Pixelized Noise

This test is performed with no illumination and one line of integration time. A single image is captured including both Temporal and Fixed Pattern Noise (FPN). A spatial low pass filter is applied to remove shading and deviant pixels (± three standard deviations) are removed. The Total Pixelized Noise is defined as one standard deviation of the frequency histogram.

#### 20) Responsivity ke/lux-sec

This number is calculated by integrating the multiplication of the sensor QE by the human photopic response assuming a 3200K light source with a QT100 IR filter. This is a sharp 650nm cutoff filter. If the IR filter is removed a higher response value will result.

#### 21) Responsivity V/lux-sec

Voltage levels are not output from the sensor. This metric uses the pixel output in volts at the ADC input for 1x Analog Gain.



## Operation

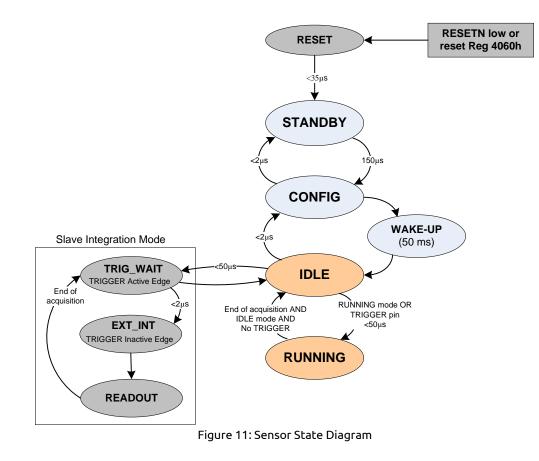
This section is a brief discussion of the most common features and functions assuming default conditions. See the *KAC-12040 User Guide* for a full explanation of the sensor operation modes, options, and registers.

## **REGISTER ADDRESSES**

The last bit of any register address is a Read/Write bit. Most references in this document refer to the Write address. All SPI reads are to an even address, all SPI writes are to an odd address.

## **SENSOR STATES**

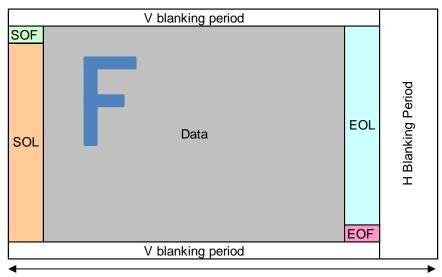
Figure 11 shows the sensor states, see the *KAC-12040 User Guide* for detailed explanation of the States.





## **ENCODED SYNCS**

To facilitate system acquisition synchronization the KAC-12040 places synchronization words (SW) at the beginning and at the end of each output row as indicated in the following Figure 12. This is performed for each of the 8 LVDS output banks providing frame, line, and output synchronization. See the *KAC-12040 User Guide* for additional detail on LVDS and Encoded Sync output.



Line length (LL) Figure 12: Encoded Frame Syncs



## LINE TIME

This Datasheet presumes the recommended startup script that is defined in the *KAC-12040 User Guide* has been applied. The KAC-12040 defaults to Dual-Scan mode. In this mode the LVDS data readout overlaps the pixel readout and ADC conversion time. The Pixel and ADC conversion time are fixed (for 10 bit operation) and total ~ 8.66 µsec. The LVDS time will be dependent on the PLL2 frequency selected. If the PLL2 < 313 MHz, then the LVDS data readout will dominate the row time. For PLL2 > 313 MHz, the Pixel + ADC will set the minimum Line Time. The Line Time is not impacted by the selection of Rolling Shutter or Global Shutter mode.

The KAC-12040 architecture always outputs two rows at once, one row from the top ADC, and one from the bottom ADC. Each ADC then divides up the pixel into  $1 \rightarrow 4$  parallel pixel output LVDS Banks. The default is 4 output banks per ADC for a total of 8 parallel pixel outputs to minimize the LVDS data output time. Since the sensor always outputs 2 rows at a time the timing and registers are based on a Line Time (LT) or Line Length (LL) where one LT = the time to readout 2 rows in parallel (one even row and one odd row).

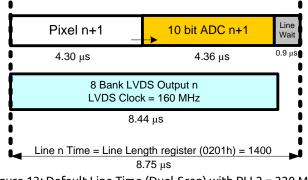


Figure 13: Default Line Time (Dual-Scan) with PLL2 = 320 MHz



## **FRAME TIME**

The frame time is defined in units of Line Time. 1 Line Time unit = 2 output rows. To first-order the frame rate is not directly impacted by selection of Global Shutter, Rolling Shutter, Dual-Scan, or Tri-Scan.

The Frame Time is made up of three phases:

- 1. Integration Phase
- 2. Readout Phase
- 3. Frame Wait Phase (Vertical Blanking, Vblank)

By default the Integration Phase overlaps the Readout and Frame Wait Phases. If the Integration Phase is larger than the Readout + Frame Wait time, then the Integration Phase will determine the video frame rate. Otherwise the frame rate will be set by the Readout + Frame Wait time. In other words, if the programmed integration time is larger than the minimum readout time (and vertical blanking) then extra vertical blanking will be added and the frame rate will slow to accommodate the requested integration time.

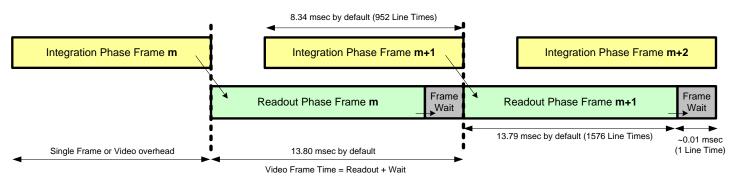


Figure 14: Default frame time configuration (Frame A)

If the Integration Phase is less than the Readout Phase then the start of integration is automatically delayed to minimize the storage time and dark current.

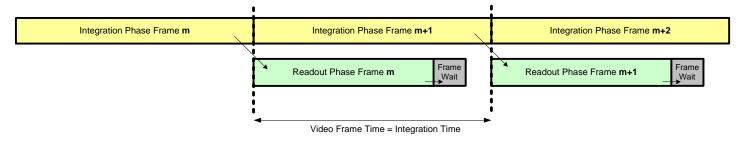


Figure 15: Frame time with extended integration time.

If the Readout Phase(+ Vblanking) is less than the Integration Phase, then the readout occurs as soon the integration is complete to minimize the storage time and dark current.

See the *KAC-12040 User Guide* for detailed calculation of the Integration Phase, Readout Phase, and Frame Wait.

To first-order the Readout Phase is equal to the number of rows \* row\_time.



## **GLOBAL SHUTTER READOUT**

Global Shutter readout provides the maximum precision for freezing scene motion. Any motion artifacts will be 100% defined by an ideal integration time edge. Every pixel in the array starts and stops integration at the same time.

Figure 16 illustrates a Global Shutter Frame readout assuming the recommended Start-up Script defined in the *KAC-12040 User Guide* (8 LVDS banks, Dual-Scan, 8.75 µSec line time). The Frame Wait Phase is not shown due to its small default size (1 LL) and for clarity.

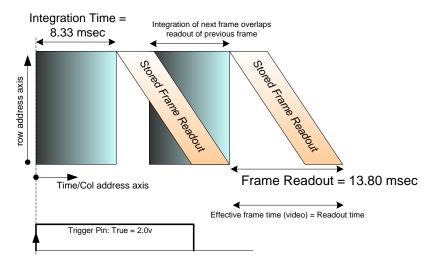


Figure 16: Illustration of frame time for Global Shutter readout

Global Shutter readout mode is selected using Bits [1:0] of Register 01D1h.

Images can be initiated by setting and holding the TRIGGER input pin or by placing the sensor into RUNNING mode by writing 03d to register 4019h. If the TRIGGER input pin is true when at the start of the integration time for the next frame then the sensor will complete an additional frame integration and readout. In the case shown in Figure 16 two frames will be output.



## **ROLLING SHUTTER READOUT**

The KAC-12040 high speed Rolling Shutter readout provides the maximum dynamic range while still providing excellent motion capture. In Rolling Shutter the readout more closely matches a film camera shutter. Each row of the image receives the same integration time, but each row starts and ends at a different time as the shutter travels from the top of the array to the bottom. In the Figure 17 frame time illustration this 'moving shutter' displays as a sloped edge for the blue pixel array region, just as the readout edge is sloped.

The Figure 17 illustration shows a 2 frame output sequence using the external TRIGGER pin.

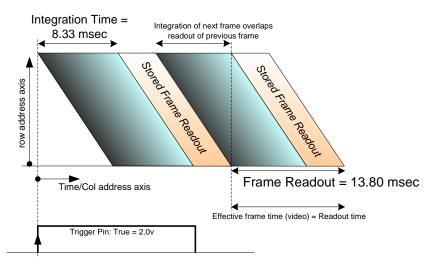


Figure 17: Illustration of Frame time for Rolling Shutter readout

Rolling Readout mode can be selected using Bits [1:0] of Register 01D1h.

Images can be initiated by setting and holding the TRIGGER input pin or by placing the sensor into RUNNING mode by writing 03d to register 4019h. If the TRIGGER input pin is True when at the start of the integration time for the next frame then the sensor will complete an additional frame integration and readout.



## 8 Bank LVDS Data Readout

## **LVDS BANKS**

The KAC-12040 provides 8 parallel pixel banks, each consisting of 8 LVDS differential pairs (7 data pairs + 1clock pair). This allows the output of 8 pixels per LVDS clock period. All 7 data pairs, of each bank, are used only in 14 bit operation mode. By default only 5 data pairs are used for 10 bit mode (D4 -> D0). The unused pairs are held in low-power high impedance mode.

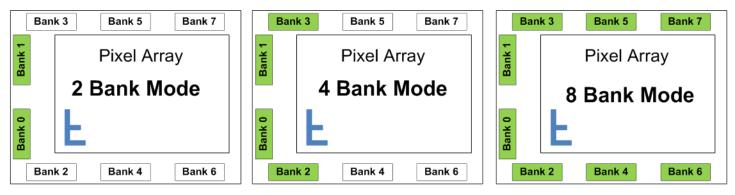


Figure 18: LVDS Bank labeling

The number of output banks used is independent of the ADC bit depth chosen. By default the KAC-12040 uses all 8 output banks for maximum frame rate. If technical restrictions prevent the use of 8 LVDS banks, the sensor can be programmed to use 4 or 2 banks, however this can result in reduced frame rate and reduction of image quality. It is recommended that 8 banks be used when possible. Only the 8 bank option is discussed in detail in this specification, see the *KAC-12040 User Guide* for additional detail on 4 and 2 bank mode.

In order to minimize the LVDS clock rate (and power) for a given data rate the pixels are output in DDR (Double Data Rate) where the MSB is always sent first (on rising edge) and the LSB second (falling edge) This is not programmable.

### PORTS PER LVDS BANK

The MSB comes out first on the falling edge, followed by the LSB on the net rising edge.

Bit Depth	Edge of DATA_CLK	Data0	Data1	Data2	Data3	Data4	Data5	Data6
14 bits	Falling (MSB nibble)	D7	D8	D9	D10	D11	D12	D13
	Rising (LSB nibble)	D0	D1	D2	D3	D4	D5	D6
12 bits	Falling (MSB nibble)	D6	D7	D8	D9	D10	D11	HiZ
12 DILS	Rising (LSB nibble)	D0	D1	D2	D3	D4	D5	HiZ
10 bits	Falling (MSB nibble)	D5	D6	D7	D8	D9	HiZ	HiZ
TODICS	Rising (LSB nibble)	D0	D1	D2	D3	D4	HiZ	HiZ
8 bits	Falling (MSB nibble)	D4	D5	D6	D7	HiZ	HiZ	HiZ
o DILS	Rising (LSB nibble)	D0	D1	D2	D3	HiZ	HiZ	HiZ

Figure 19: Number of LVDS pairs (ports) used vs. bit depth



## **8 BANK PIXEL ORDER**

The KAC-12040 always processes two rows at a time. Even row decodes are sent to the bottom ADC and LVDS output banks (0, 2, 4, 6). Odd rows are sent to the top ADC and LVDS banks (1, 3, 5, 7). The ROI must be (and is internally forced to) an even size and always starting on an even row decode.

The rows are read out progressively left to right (small column address to large). Eight pixels are sent out of the chip at once, one pixel per LVDS bank per LVDS clock cycle.

Pixel Readout order:

- 1. Two rows are selected, the even row is sent to the bottom ADC and the odd row to the top ADC.
- 2. Each ADC converts its row of pixel data at once and stores the result in a line buffer.
- 3. At default settings there are 4 output LVDS banks for each ADC.
- 4. Each LVDS Bank outputs one pixel per clock cycle, so 4 pixels of each row are output each full LVDS clock cycle, two rows in parallel for 8 pixels per clock cycle total.
- 5. The pixels are sent out from left to right (low column number to high column number). So the first 4 pixels are sent out on clock cycle 1, and the next 4 pixels to the right are sent out on clock cycle 2.
- 6. To conserve the number of wires per port, the 10 bits per pixel are sent out DDR (Dual Data Rate) over 5 ports. On the falling edge the upper 5 MSB bits are sent out, and on the rising edge the lower 5 bits LSB are sent out. Completing one full LVDS clock cycle and one set of eight pixels.

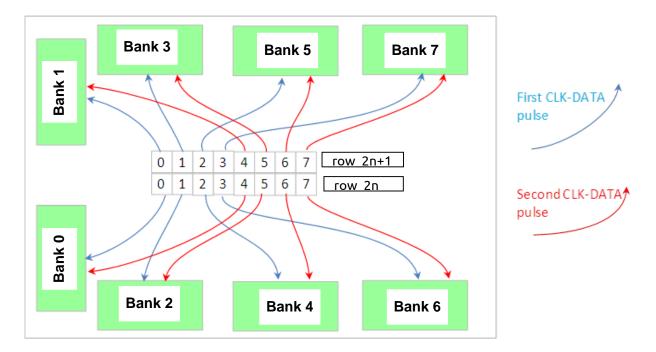


Figure 20: Pixel readout order diagram



LVDS Bank	Row	Pixel number										
Bank 0	2 n (even)	0	4	8	12	16						
Bank 2	2 n (even)	1	5	9	13	17						
Bank 4	2 n (even)	2	6	10	14	18						
Bank 6	2 n (even)	3	7	11	15	19						
Bank 1	2 n+1 (odd)	0	4	8	12	16						
Bank 3	2 n+1 (odd)	1	5	9	13	17						
Bank 5	2 n+1 (odd)	2	6	10	14	18						
Bank 7	2 n+1 (odd)	3	7	11	15	19						
	LVDS Clock Cycle	1	2	3	4	5						

Figure 21: Pixel readout order table

## **De-Serializer Settings**

Figure 22 shows the data stream of one LVDS bank for 10 bit resolution.

Data serialization is fixed at 2 cycle DDR for all bit depths. Data output order is MSB first on the falling edge, and LSB following on the rising edge.

Four pixel values per synchronization word are embedded into the video stream per LVDS bank.

The SOL/SOF synchronization words are sent out of each LVDS bank before the first valid pixel data from that bank. Each bank outputs all 4 syncs of the SOF or SOL.

And each of the active LVDS banks each output all 4 sync codes for the EOL/EOF.

Dc1k0	Ţ		Ţ		Ţ			ſ	Ţ		Ţ	ſ	Ţ	ſ	Ţ		Ļ		Ţ		Ţ		Ţ		Ţ		Ţ		Ţ	
Data0	D5	D0	D5	D0	D5	<b>D</b> 0	D5	D0	D5	<b>D0</b>	D5	DO	D5	DO	D5	D0			D5	<b>D</b> 0	D5	DO	D5	<b>D</b> 0	D5	D0	D5	D0	D5	DO
Data1	D6	D1	D6	D1	D6	D1	D6	D1	D6	D1	D6	D1	D6	D1	D6	D1			D6	D1	D6	D1	D6	D1	D6	D1	D6	D1	D6	D1
Data2	D7	D2	D7	D2	D7	D2	D7	D2	D7	D2	D7	D2	D7	D2	D7	D2			D7	D2	D7	D2	D7	D2	D7	D2	D7	D2	D7	D2
Data3	D8	D3	D8	D3	D8	D3	D8	D3	D8	D3	D8	D3	D8	D3	D8	D3			D8	D3	D8	D3	D8	D3	D8	D3	D8	D3	D8	D3
Data4	D9	D4	D9	D4	D9	D4	D9	D4	D9	D4	D9	D4	D9	D4	D9	D4			D9	D4	D9	D4	D9	D4	D9	D4	D9	D4	D9	D4
	MSB	LSB	MSB	LSB	MSB	LSB	MSB	LSB	MSB	LSB	MSB	LSB	MSB	LSB	MSE	B LSB			MSB	LSB	MSB	LSB	MSB	LSB	MSB	LSB	MSB	S LSB	MSB	LSB
	SW1 SW2				1 SW2 SW3 SW4 P0 P1					F	P2 P3 PN-1 PN					PN SW1 SW2 SW3 SW4				W4										
													_	~	-					$\nearrow$				_						
	Synchronisation word on 10 bits												Dat	a'on 1	10 bits							Syn	chror	nisatio	n w	ord or	n 10 t	bits		

Figure 22: Data stream of one LVDS Bank for 10bits ADC resolution



# **Register Definition**

16 bit Write addr (Hex)	Default Value Hex/Dec	SPI State	Group	Register Name	16 bit Write addr (Hex)	Default Value Hex/Dec	SPI State	Description
0001	420d			Frame A ROI y1	01D1	CC11h	CONFIG only	Config1
0009	2176d			Frame A ROI h1	01D9	0000h	CONFIG or IDLE	Config2
0011	80d			Frame A ROI x1	01E1	000Ah	CONFIG or IDLE	Analog/Digital Power Mode
0019	3856d			Frame A ROI w1	01E9	0000h	CONFIG or IDLE	Dual-Video Repetition
0021	0d			Frame A sub-ROI y2	01F1	b0	CONFIG or IDLE	Vertical Blanking
0029	0d			Frame A sub-ROI h2	01F9	1938d	CONFIG or IDLE	Fixed Frame Period
0031	0d			Frame A sub-ROI x2	0201	1376d	CONFIG or IDLE	Line Length (LL)
0039	0d			Frame A sub-ROI w2	0209	0028h	CONFIG or IDLE	ADC Bit Depth
0041	0d		по	Frame A sub-ROI y3	0211	0000h	CONFIG or IDLE	FLO Edge
0049	0d		Definition	Frame A sub-ROI h3	0219	0000h	CONFIG or IDLE	MSO Edge
0051	0d	Any	Defi	Frame A sub-ROI x3	0709	0000h	Any	CFA Feedback
0059	b0	A	<	Frame A sub-ROI w3	0711	0000h	Any	Temperature Sensor FB
0061	0d		Frame	Frame A sub-ROI y4	0719	0000h	Any	General Feedback
0069	b0		Fr	Frame A sub-ROI h4	2059	0300h	CONFIG only	Output Bank Select 1
0071	0d			Frame A sub-ROI x4	2099	2877h	CONFIG only	PLL1 Setting
0079	b0			Frame A sub-ROI w4	20A1	0861h	CONFIG only	PLL2 Setting
0081	0033h			Frame A Decimation	2449	0432h	CONFIG only	Sub-LVDS Enable
0089	b0			Frame A Video Blanking	2479	10ABh	Any	Column Clamp Threshold A
0091	800d			Frame A Integration Lines	2481	20C7h	Any	Column Clamp Threshold B
0099	0d			Frame A Integration Clocks	2499	0011h	CONFIG or IDLE	Test Pattern Control 1
00A1	10d			Frame A Black Level	24A1	0220h	CONFIG or IDLE	Test Pattern Control 2
00A9	001Fh			Frame A Gain	24B9	202d	STANDBY only	Slope 1 Length
00E9	0d			Frame B ROI y1	24C1	101d	STANDBY only	Slope 2 Length
00F1	3016d			Frame B ROI h1	24C9	101d	STANDBY only	Slope 3 Length
00F9	0d			Frame B ROI x1	24D1	101d	STANDBY only	Slope 4 Length
0101	4016d			Frame B ROI w1	24D9	101d	STANDBY only	Slope 5 Length
0109	0d	_		Frame B sub-ROI y2	24E1	420d	STANDBY only	Slope 6 Length
0111	0d			Frame B sub-ROI h2	24E9	0083h	STANDBY only	Slope 1/2 Gain
0119	0d			Frame B sub-ROI x2	24F1	038Fh	STANDBY only	Slope 3/4 Gain
0121	0d			Frame B sub-ROI w2	24F9	0FBFh	STANDBY only	Slope 5/6 Gain
0129	0d	_	ion	Frame B sub-ROI y3	2501	1F9Fh	STANDBY only	Slope 7 Gain
0131	0d		Definition	Frame B sub-ROI h3	2559	4804h	Any	Defect Avoidance Threshold
0139	0d	Any	Def	Frame B sub-ROI x3	2561	0006h	Any	Defect Avoidance Enable
0141	0d	⊲	e B	Frame B sub-ROI w3	25C1	0003h	CONFIG or IDLE	Encoded Sync Config
0149	0d	_	rame	Frame B sub-ROI y4	25C9	000Ah	CONFIG only	LVDS Power-Down
0151	0d		Fr	Frame B sub-ROI h4	2619	000Bh	CONFIG only	Output Bank Select 2
0159	0d	_		Frame B sub-ROI x4	2D89	0000h	CONFIG only	FLO/MSO Polarity
0161	0d			Frame B sub-ROI w4	4001	4100h	Any	Chip Revision Code
0169	0033h	_		Frame B Decimation	4009	0011h	Any	Chip ID Code MSB
0171	0d			Frame B Video Blanking	4011	0080h	Any	Chip ID Code LSB
0179	800d	_		Frame B Integration Lines	4019	0000h	Any	Set Sensor State
0181	0d			Frame B Integration Clocks	4021	0000h	CONFIG or IDLE	OTP Address
0189	10d	_		Frame B Black Level	4029	0000h	CONFIG or IDLE	OTP Write Data
0191	001Fh			Frame B Gain	4031	0000h	CONFIG or IDLE	Command_Done_FB
					4041	0000h	CONFIG or IDLE	OTP Read Data
						00001		

4061

0000h

CONFIG or IDLE

Soft Reset



Notes - SPI State (the Sensor State from which the register can be set):

- 1. "Any": Can be written from any state (including RUNNING)
- 2. "CONFIG or IDLE": These registers can be changed in IDLE or CONFIG states
- 3. "CONFIG only": Sensor must be in CONFIG state to set these registers.
- 4. Only Register 4018h and 4060h may be set when the sensor is in STANDBY state.

Notes - Decimal, hexadecimal, binary values:

- 1. "b" denotes a binary number, a series of bits: MSB is on the left, LSB is on the right.
- 2. "h" or "hex" denotes a hexadecimal number (Base 16, 1-9, A-F). The letters in a hex number are always capitalized
- 3. "d" denotes a decimal number.
- 4. Note that "0" and "1" are the same value in all number base systems and sometimes the base notation is omitted.

The KAC-12040 features an embedded microprocessor by Cortus.

## Absolute Maximum Ratings

For Supplies and Inputs the maximum rating is defined as a level or condition that should not be exceeded at any time. If the level or the condition is exceeded, the device will be degraded and may be damaged. Operation at these values will reduce Mean Time to Failure (MTTF).

### **SUPPLIES**

Parameter	Value
AVDD_LV, VDD_DIG	-0.25 V; 2.3 V
AVDD_HV, Vref_P, VDD_LVDS	-0.25 V; 4 V
DC Input voltage at any input pin	-0.25 V; VDD_DIG +0.25 V

## **CMOS** INPUTS

Parameter	Symbol	Min	Typical	Max	Unit
Input voltage low level	VIL	-0.3		0.35 VDD_DIG	V
Input voltage high level	VIH	0.65 VDD_DIG		VDD_DIG +0.3	V



## **Operating Ratings**

### **INPUT CLOCK CONDITIONS**

Parameter	Min	Typical	Max	Unit
Frequency for Clk_In1 and Clk_In2	45	48	50	MHz
Duty cycle for Clk_In1 and Clk_In2	40	50	60	%
RESETN	10			ns
TRIGGER pin minimum pulse width	20			ns

TRIGGER must be active at least 4 periods of PLL1 (~12.5 ns at 320 MHz) to start a capture cycle. The polarity of the active level is configurable by SPI (Register 01D8h Bit 0), the default is active high (ie pin = VDD\_DIG = trigger request).

#### **OPERATING TEMPERATURE**

Operating Temperature TOP -50 +80 °C 1	Description	Symbol	Minimum	Maximum	Units	Notes
	Operating Temperature	ТОР	-50	+80	°C	1

Notes:

1. Under conditions of no condensation on the sensor

## **CMOS IN/OUT**

CMOS IN/OUT characteristics

Parameter	Symbol	Min	Typical	Max	Unit
Output voltage low level	Vol			0.45	V
Output voltage high level	Vон	VDD_DIG-0.45			V
Input Hysteresis voltage	VTH		0.25		
Pull-up resistor value for RESETN pin	R <sub>PU</sub>	62			kΩ
Pull-down resistor value for TRIGGER pin	R <sub>PD</sub>	100			kΩ
Current on ADC_REF pin	IADC_REF		100		μA



### **SUPPLIES**

LVDS is supply         VVD LVDS         3.15         3.30         3.63         V           Pixet high voltage supply         AVDD_LVV         3.40         3.50         3.60         V           Pixet low voltage supply         Vref_P         2.71         2.80         2.88         V           Pixet low voltage supply         AVDD_LVV         1.71         1.80         1.88         V           Analog power supply         AVDD_LVV         1.71         1.80         1.88         V           Digital power supply         VDD_Dig         1.9         2.00         2.10         V           AVDD_LVV - Vref_P	Parameter	Symbol	Min	Typical	Max	Unit	Comment
Pixel low voltage supply         Vire P         2.71         2.80         V           Pixel low voltage supply         AVDD_LV         1.71         1.80         1.89         V           Digital power supply         VDD_Dig         1.9         2.00         2.10         V           Digital power supply         VDD_Dig         1.9         2.00         2.10         V           AVDD_HV - Vref P          0         0.5         V         V           Power in STANDBY State          10         mW             Current in STANDBY State          <	LVDS IO supply	VDD_LVDS	3.15	3.30	3.63	V	
Analog power supplyAVDD_LV1.711.801.89VDigital power supplyVDD_Dig1.92.002.10VDog tal power supplyI1.92.002.10VAVDD_HV-Vref_PIIIIIPower in STANDBY StateIIIMWICurrent in STANDBY StateIIIMWICurrent in STANDBY StateIIIMMICurrent in STANDBY StateIIIMMIMADD_LVDSIIIMMIIMVDD_LVDSIIIMMIIMVDD_LVDSIIIMMIIPower in CONFIG StateIIIIIICurrent in CONFIG StateIIIIIIIVDD_LVDSIIIIIIIIVDD_LVDSIII	Pixel high voltage supply	AVDD_HV	3.40	3.50	3.60	V	
Analog power supplyAVDD_LV1.711.801.89VDigital power supplyVDD_Dig1.92.002.10VDog tal power supplyI1.92.002.10VAVDD_HV-Vref_PIIIIIPower in STANDBY StateIIIMWICurrent in STANDBY StateIIIMWICurrent in STANDBY StateIIIMMICurrent in STANDBY StateIIIMMIMADD_LVDSIIIMMIIMVDD_LVDSIIIMMIIMVDD_LVDSIIIMMIIPower in CONFIG StateIIIIIICurrent in CONFIG StateIIIIIIIVDD_LVDSIIIIIIIIVDD_LVDSIII							
Digital power supplyVDD_Dig1.92.002.10VAVDD_IV- Vief_P0.5VVPower in STANDBY State10mWCurrent in STANDBY State10mMVCurrent in STANDBY State<<0.5	Pixel low voltage supply		2.71	2.80	2.88	V	
AVDD_HV - Vref_P0.5VDewer in STANDBY State10mWCurrent in STANDBY State10mWCurrent in STANDBY State<0.5	Analog power supply	AVDD_LV	1.71	1.80	1.89	V	
Power in STANDBY StateImage: stand	Digital power supply	VDD_Dig	1.9	2.00	2.10	V	
Current in STANDBY StateImage: stateImage: stateVDD_LVDS<	AVDD_HV – Vref_P			0.5		V	
Current in STANDBY StateImage: stateImage: stateVDD_LVDS<							
VDD_LVDS<<<< </td <td>Power in STANDBY State</td> <td></td> <td></td> <td>10</td> <td></td> <td>mW</td> <td></td>	Power in STANDBY State			10		mW	
AVDD_HV<0.5.mAAVDD_LV<0.5.	Current in STANDBY State						
AVDD_LV </td <td>VDD_LVDS</td> <td></td> <td></td> <td>&lt; 0.5</td> <td></td> <td>mA</td> <td></td>	VDD_LVDS			< 0.5		mA	
Vinef_PmAVDD_DIGAmAmAPower in CONFIG State330mWCurrent in CONFIG StatemAVDD_LVDSmACurrent in CONFIG StatemAVDD_LVDSmAMUDD_HVUmAMUDD_HVUmAMUDD_HVUmAMUDD_HVUmAMUDD_DIGmAMUDD_DIGmAMUDD_DIGmAMUDD_DIGmAMUDD_DIGmAMUDD_DIGmAMurei IDLE StatemAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDSmAMUDD_LVDS in standard LVDS modemAVDD_LVDS in Sub-LVDS modemAVDD_LVDS in Sub-LVDS modemAVDD_LVDS in Sub-LVDS mode </td <td>AVDD_HV</td> <td></td> <td></td> <td>&lt; 0.5</td> <td></td> <td>mA</td> <td></td>	AVDD_HV			< 0.5		mA	
VDD_DIGMAMAPower in CONFIG StateImage: State St	AVDD_LV			< 0.5		mA	
Power in CONFIG State         Image: Construct of the state         Image: Construct of the state         Image: Construct of the state           Current in CONFIG State         Construct of the state         Construct of the state         Construct of the state           AVDD_HV         Construct of the state         Construct of the state         Construct of the state         Construct of the state           AVDD_HV         Construct of the state         Construct of the state         Construct of the state         Construct of the state           Power in IDLE State         Construct of the state         Construct of the state         Construct of the state         Construct of the state           Current in IDLE State         Construct of the state         Construct of the state         Construct of the state         Construct of the state           Current in IDLE State         Construct of the state         Construct of the state         Construct of the state         Construct of the state           Current in IDLE State         Construct of the state         Construct of the state         Construct of the state         Construct of the state           Current in RUNNING State         Construct of the state         Construct of the state         Construct of the state         Construct of the state           VDD_LVDS in standard LVDS mode         Construct of the state         Constate         Construct of the state	Vref_P			< 0.5		mA	
Current in CONFIG StateImage: stateImage: stateImage: stateVDD_LVDSImage: stateImage: stateImage: stateImage: stateAVDD_LVImage: stateImage: stateImage: stateImage: stateAVDD_LDGImage: stateImage: stateImage: stateImage: stateVDD_LDGImage: stateImage: stateImage: stateImage: statePower in IDLE StateImage: state <td< td=""><td>VDD_DIG</td><td></td><td></td><td>4</td><td></td><td>mA</td><td></td></td<>	VDD_DIG			4		mA	
Current in CONFIG StateImage: stateImage: stateImage: stateVDD_LVDSImage: stateImage: stateImage: stateImage: stateAVDD_LVImage: stateImage: stateImage: stateImage: stateAVDD_LDGImage: stateImage: stateImage: stateImage: stateVDD_LDGImage: stateImage: stateImage: stateImage: statePower in IDLE StateImage: state <td< td=""><td></td><td></td><td></td><td></td><td></td><td></td><td></td></td<>							
VDD_LVDS<<<mAAVDD_HV<0.5	Power in CONFIG State			330		mW	
AVDD_HV<<MAAVDD_LV<	Current in CONFIG State						
AVDD_LV<<0.5mAVref_P<0.5	VDD_LVDS			< 0.5		mA	
Vref_P </td <td>AVDD_HV</td> <td></td> <td></td> <td>&lt; 0.5</td> <td></td> <td>mA</td> <td></td>	AVDD_HV			< 0.5		mA	
VDD_DIG145mAPower in IDLE StateImage: State Sta	AVDD_LV			< 0.5		mA	
Image: system         Image: s	Vref_P			< 0.5		mA	
Current in IDLE StateImage: stateImage: stateImage: stateVDD_LVDSImage: state< 0.5	VDD_DIG			145		mA	
Current in IDLE StateImage: stateImage: stateImage: stateVDD_LVDSImage: state< 0.5							
VDD_LVDS<<MAAVDD_HV20mAAVDD_LV<	Power in IDLE State			410		mW	
AVDD_HV20mAAVDD_LV<	Current in IDLE State						
AVDD_LV<<<mAVref_P<	VDD_LVDS			< 0.5		mA	
Vref_PMAMAVDD_DIGIma145MAVDD_DIGImaImaImaPower in RUNNING StateImaImaImaCurrent in RUNNING StateImaImaImaVDD_LVDS in standard LVDS modeImaImaImaVDD_LVDS in Sub-LVDS modeImaImaImaImage: NVDD_LVDS in Sub-LVDS modeImaImaImaImage: NVD_LVDS in Sub-LVDS modeImaImaImaImage: NVD_LVD Image: NVD_LVD Image: NVD_LVDImage: NVDImage: NVDImage: NVD_LVD Image: NVDE_LVDImaImaImaImage: NVDE_LVD Image: NVDE_LVDImage: NVDE_LVDImage: NVDE_LVDImage: NVDE_LVDImage: NVDE_LVD Image	AVDD_HV			20		mA	
VDD_DIGImage: stateMAPower in RUNNING StateImage: stateImage: stateImage: stateCurrent in RUNNING StateImage: stateImage: stateImage: stateVDD_LVDS in standard LVDS modeImage: stateImage: stateImage: stateVDD_LVDS in Sub-LVDS modeImage: stateImage: stateImage: stateVDD_LVDS in Sub-LVDS modeImage: stateImage: stateImage	AVDD_LV			< 0.5		mA	
VDD_DIGImage: stateMAPower in RUNNING StateImage: stateImage: stateImage: stateCurrent in RUNNING StateImage: stateImage: stateImage: stateVDD_LVDS in standard LVDS modeImage: stateImage: stateImage: stateVDD_LVDS in Sub-LVDS modeImage: stateImage: stateImage: stateVDD_LVDS in Sub-LVDS modeImage: stateImage: stateImage	Vref_P			< 0.5		mA	
Current in RUNNING State         Image: mark standard LVDS mode         Image: mark standard LVDS mode         Image: mark standard LVDS mode         Image: mark standard standard LVDS mode         Image: mark standard stan	1			145		mA	
Current in RUNNING State         Image: mark standard LVDS mode         Image: mark standard LVDS mode         Image: mark standard LVDS mode         Image: mark standard standard LVDS mode         Image: mark standard stan							
VDD_LVDS in standard LVDS mode         164         mA           VDD_LVDS in Sub-LVDS mode         104         mA           AVDD_HV         74         mA           AVDD_LV         12         mA           Vref_P         26         mA	Power in RUNNING State			1.5		W	
VDD_LVDS in Sub-LVDS mode         104         mA           AVDD_HV         74         mA           AVDD_LV         12         mA           Vref_P         26         mA	Current in RUNNING State						
VDD_LVDS in Sub-LVDS mode         104         mA           AVDD_HV         74         mA           AVDD_LV         12         mA           Vref_P         26         mA	VDD_LVDS in standard LVDS mode			164		mA	
AVDD_LV         12         mA           Vref_P         26         mA				104		mA	
Vref_P         26         mA	AVDD_HV			74		mA	
Vref_P         26         mA	AVDD_LV			12		mA	
VDD_DIG 396 mA				26		mA	
	VDD_DIG			396		mA	

Notes:

1. Voltages relative to VSS. Current measurements made in darkness.

2. PLL2 = 320 MHz, Max frame rate (i.e., no row or frame wait time). These average power values will decrease at lower frame rate either from reduced PLL2 or low power state during Line and Frame blanking.

3. Sub-LVDS active.



# SPI (Serial Peripheral Interface)

The SPI communication interface lets the application system to control and configure the sensor. The sensor has an embedded slave SPI interface. The application system is the master of the SPI bus.

Name	Sensor I/O direction	Description
CSN	I	SPI chip select - Active low, this input activates the slave interface in the sensor
SCK	I	SPI clock – toggled by the master
MISO	0	SPI master serial data input – slave(sensor) serial data output
MOSI	I	SPI master serial data output – slave(sensor) serial data input

Parameter	Min	Тур	Max	Unit
SPI SCK	5	25	50	MHz
Duty cycle on SPI SCK	40	50	60	%

# **CLOCK POLARITY AND PHASE**

CPOL(Clock POLarity) and CPHA(Clock PHAse) are commonly defined in SPI protocol such as to define SCK clock phase and polarity. The KAC-12040 defaults to expecting the master to be configured with CPOL=1 (the base value of the clock is VDD\_DIG) and CPHA=1 (data is valid on the clock rising edge).

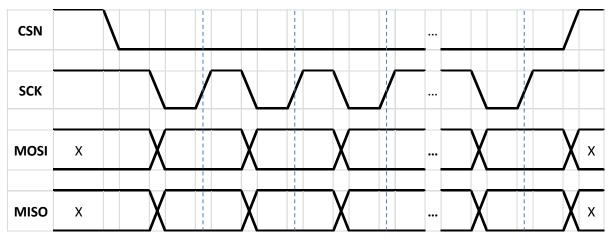


Figure 23: CPOL = 1 and CPHA = 1 configuration



## **SPI PROTOCOL**

	byte	byte	byte	byte
	0	1	2	3
CSN				
Sclk	8 cycles	8 cycles	8 cycles	8 cycles
	16	bit	16	bit
MOSI	Addres	s word	Data to	o write
	MSB	LSB	MSB	LSB

Figure 24: SPI Write byte order

	byte 0	byte 1				byte 2	byte 3	
CSN								
Sclk	8 cycles	8 cycles				8 cycles	8 cycles	
MOSI		bit ss word LSB						
MISO			Wait time 1.5 usec		16 Read MSB	bit data LSB		

Figure 25: SPI Read byte order

There is a delay during readback between presenting the address to be read on the MOSI and being able to read the register contents on the MISO. This delay is not the same for all registers. Some are available immediately, some require a longer fetch time. The 1.5 µsec shown in Figure 14 is the maximum time to fetch a register's value when in CONFIG state (the recommended state for changing registers). Some registers can be adjusted during RUNNING state (see the register summary on page 34). If performing a readback during RUNNING state, the delay could be as long as 4.5 µsec depending on when in the row the request was sent and the sensor's microcontroller activity at that moment.

Note that readback does not provide the actual register value being used, but reflects the next value to be used. All new register writes are placed in a shadow memory until they can be updated into the active memory. This active memory update occurs at the start of the next frame or upon entering the state listed in the Register Summary table on page 34. Register reads access this shadow memory not the active memory. For instance if the sensor is in RUNNING mode and you adjust the LL in register 200h. You can read back and confirm that your register change was received by the sensor; however, the LL will not change since register 200h can only be changed in CONFIG state. If you change the sensor state to CONFIG and then back to RUNNING, then the new LL will take effect.



# **SPI** INTERFACE

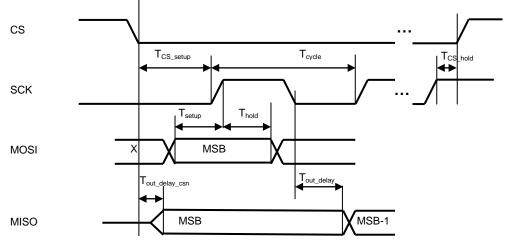


Figure 26: SPI timing chronogram

# **SPI** TIMING SPECIFICATION

Symbol	Min Value	Max Value
T <sub>cycle</sub>	20 ns	200 ns
T <sub>setup</sub>		2.9 ns
Thold	0.8 ns	
T <sub>cs_setup</sub>		2.5 ns
T <sub>cs_hold</sub>	0.1 ns	
T <sub>out_delay_csn</sub>	3.1 ns	4.7 ns
Tout_delay	4.9 ns	8.7 ns



# **LVDS Interface**

The data output can be configured to follow standard TIA/EIA-644-A LVDS specification or a low power mode compatible with common Sub-LVDS definition used in FPGA industry. (Please refer to the *KAC-12040 User Guide* for more information).

Unless otherwise noted, min/max characteristics are for T = -40 °C to +85 °C, output termination resistance RL = 100  $\Omega$  ± 1%, Typical values are at VDD\_LVDS =3.3 V.

Use register 2449h to select standard or Sub-LVDS. This document assumes that Sub-LVDS is active for all power measurements. Standard LVDS can increase the average power consumption as much as 200 mW in the case of minimum horizontal and vertical blanking.

### **STANDARD LVDS CHARACTERISTICS**

Parameter	Symbol	Min	Typical	Max	Unit
Differential output voltage	VOD	250	355	450	mV
VOD variation between complementary output states	ΔVOD	-20		+20	mV
Common mode output voltage	VOCM	1.235	1.259	1.275	V
VOCM variation between complementary output states	Δνοςμ	-25		+25	mV
High impedance leakage current	IOZD	-1		+1	μA
Output short circuit current: when D+ or D- connected to ground when D+ or D- connected to 3.3V	IOSD	2.9 12.25		4.3 30.47	mA
Output capacitance	CDO		1.3		рF
Maximum Transmission Capacitive Load Expected (for 260 MHz LVDS clock)				10	pF

# **SUB-LVDS** CHARACTERISTICS

Parameter	Symbol	Min	Typical	Max	Unit
Differential output voltage	Vod	140	180	220	mV
VOD variation between complementary output states	$\Delta V_{OD}$	-5		+5	mV
Common mode output voltage	V <sub>OCM</sub>	0.88	0.9	0.92	V
VOCM variation between complementary output states	ΔV <sub>OCM</sub>	-10		+10	mV
High impedance leakage current	lozd	-1		+1	μA
Output short circuit current: when D+ or D- connected to ground when D+ or D- connected to 3.3V	losd	1.4 10.21		2.2 30.47	mA
Output capacitance	C <sub>DO</sub>		1.3		рF
Maximum Transmission Capacitive Load Expected (for 260 MHz LVDS clock)				10	рF

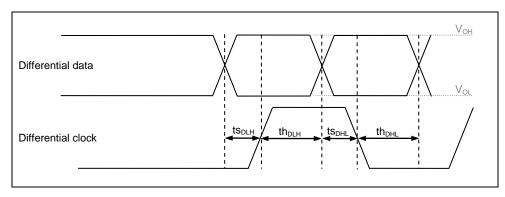
Parameter	Min	Typical	Max	Unit
LVDS_CLK	50	160	160	MHz
Duty cycle on LVDS_CLK		50		%



## IN-BLOCK LVDS TIMING SPECIFICATION

The table below gives LVDS timing specification for one group of LVDS for nominal frequency of 260 MHz. There is no skew specification between groups.

Parameter	Symbol	Value	Typical	Max	Unit
Minimum time between data change and clock rising edge	<b>ts</b> DLH	600			ps
Minimum time between clock rising edge and data change	th <sub>DLH</sub>	600			ps
Minimum time between data change and clock falling edge	<b>ts</b> DHL	600			ps
Minimum time between clock falling edge and data change	thdhl	600			ps
Maximum differential skew between the 7 data pairs	<b>t</b> skd		200	700	ps



#### Figure 27: LVDS timing chronogram

## INTER-BLOCK LVDS TIMING SPECIFICATION

Parameter	Min	Typical	Max	Unit
Inter-block skew		6	12	LVDS Clock period



# Storage and Handling

## **STORAGE CONDITIONS**

Description	Symbol	Minimum	Maximum	Units	Notes
Storage Temperature	TST	-40	+80	°C	1
Humidity	RH	5	90	%	2

#### Notes:

- 1. Long-term storage toward the maximum temperature will accelerate color filter degradation.
- 2. T=25 °C. Excessive humidity will degrade MTTF.

# ESD

- This device contains limited protection against Electrostatic Discharge (ESD). ESD events may cause irreparable damage to an image sensor either immediately or well after the ESD event occurred. Failure to protect the sensor from electrostatic discharge may affect device performance and reliability.
- 2. Devices should be handled in accordance with strict ESD procedures. Devices are shipped in static-safe containers and should only be handled at static-safe workstations.
- 3. See Application Note *Image Sensor Handling Best Practices* for proper handling and grounding procedures. This application note also contains workplace recommendations to minimize electrostatic discharge.
- 4. Store devices in containers made of electroconductive materials.

### **COVER GLASS CARE AND CLEANLINESS**

- 1. The cover glass is highly susceptible to particles and other contamination. Perform all assembly operations in a clean environment.
- 2. Touching the cover glass must be avoided.

3. Improper cleaning of the cover glass may damage these devices. Refer to Application Note *Image Sensor Handling Best Practices*.

### **ENVIRONMENTAL EXPOSURE**

- Extremely bright light can potentially harm image sensors. Do not expose to strong sunlight for long periods of time, as the color filters and/or microlenses may become discolored. In addition, long time exposures to a static high contrast scene should be avoided. Localized changes in response may occur from color filter/microlens aging.
- 2. Exposure to temperatures exceeding maximum specified levels should be avoided for storage and operation, as device performance and reliability may be affected.
- 3. Avoid sudden temperature changes.
- 4. Exposure to excessive humidity may affect device characteristics and may alter device performance and reliability, and therefore should be avoided.
- 5. Avoid storage of the product in the presence of dust or corrosive agents or gases, as deterioration of lead solderability may occur. It is advised that the solderability of the device leads be assessed after an extended period of storage, over one year.

### **SOLDERING RECOMMENDATIONS**

- 1. For manual soldering the soldering iron tip temperature is not to exceed 370 °C. Higher temperatures may alter device performance and reliability.
- Re-Flow soldering method is not recommended. Wave soldering may be employed, however solder should not touch the cover glass.



Mechanical Information

### **COMPLETED ASSEMBLY**

#### Notes:

- 1. See Ordering Information for marking code.
- 2. No materials to interfere with clearance through package holes.
- 3. Imaging Array is centered at the package center.
- 4. Length dimensions in mm units

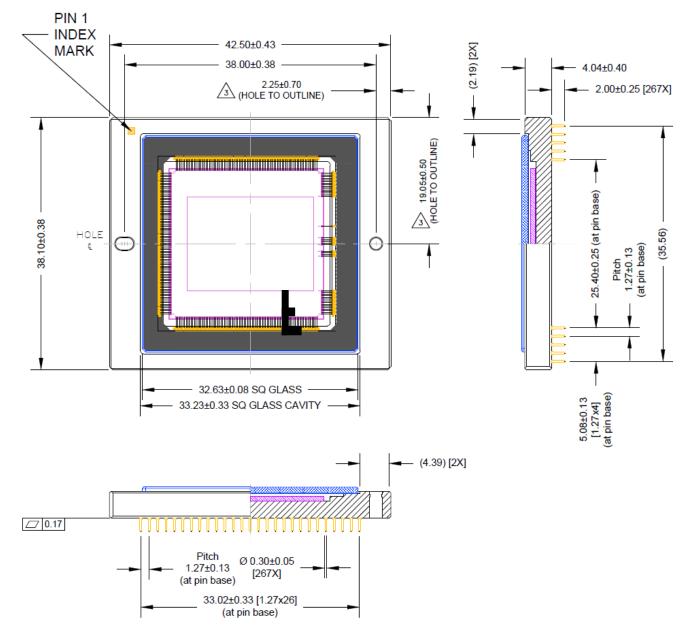


Figure 28: Completed Assembly (1 of 5)



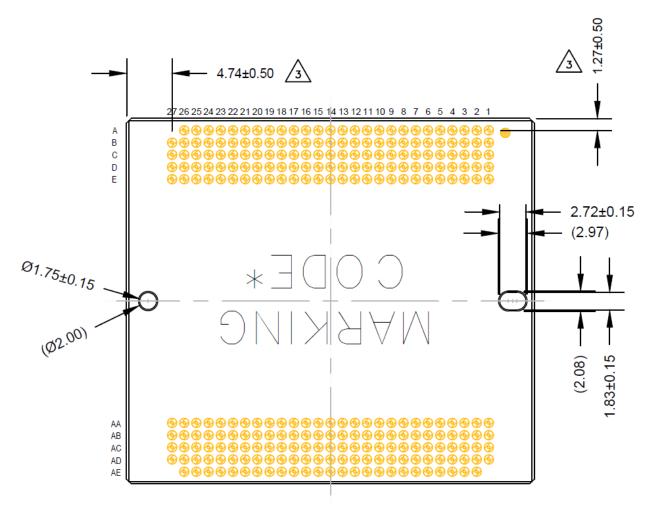


Figure 29: Completed Assembly (2 of 5)



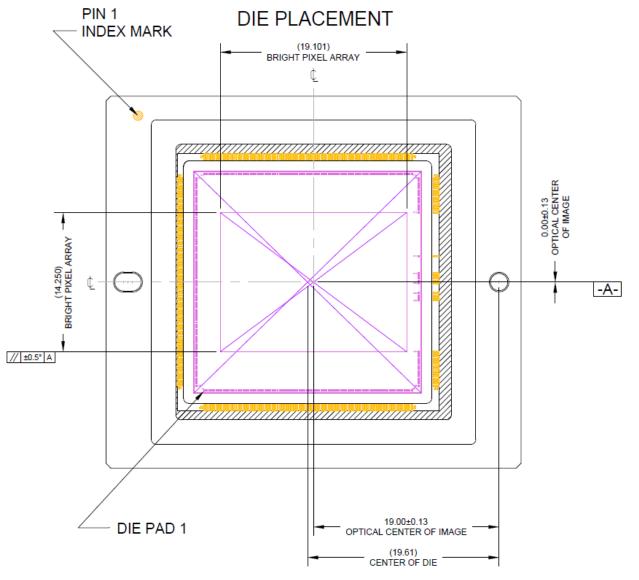


Figure 30: Completed Assembly (3 of 5)



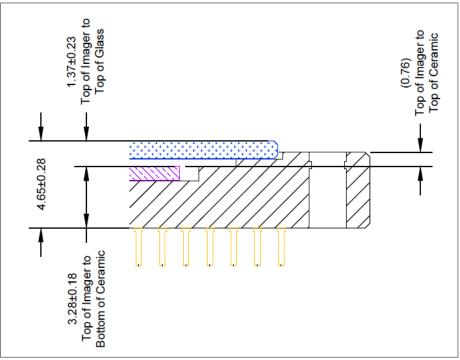


Figure 31: Completed Assembly (4 of 5)

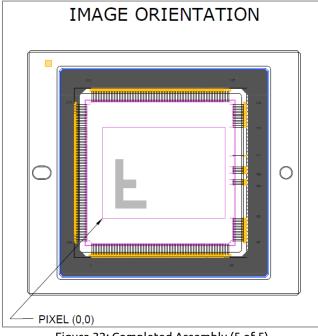


Figure 32: Completed Assembly (5 of 5)



# MAR (MULTI-LAYER ANTIREFLECTIVE COATING) COVER GLASS

#### Notes:

- 1. Units: IN [MM]
- 2. A-Zone Dust/Scratch Spec: 10 µm Maximum
- 3. Index of refraction: 1.5231

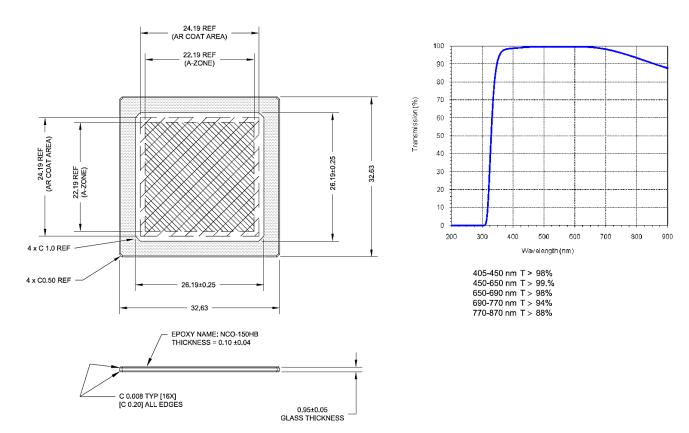


Figure 33: MAR Cover Glass Specification



# **Quality Assurance and Reliability**

## **QUALITY AND RELIABILITY**

All image sensors conform to the specifications stated in this document. This is accomplished through a combination of statistical process control and visual inspection and electrical testing at key points of the manufacturing process, using industry standard methods. Information concerning the quality assurance and reliability testing procedures and results are available from ON Semiconductor upon request. For further information refer to Application Note *Quality and Reliability*.

### REPLACEMENT

All devices are warranted against failure in accordance with the *Terms of Sale*. Devices that fail due to mechanical and electrical damage caused by the customer will not be replaced.

### LIABILITY OF THE SUPPLIER

A reject is defined as an image sensor that does not meet all of the specifications in this document upon receipt by the customer. Product liability is limited to the cost of the defective item, as defined in the *Terms of Sale*.

### LIABILITY OF THE CUSTOMER

Damage from mishandling (scratches or breakage), electrostatic discharge (ESD), or other electrical misuse of the device beyond the stated operating or storage limits, which occurred after receipt of the sensor by the customer, shall be the responsibility of the customer.

### **TEST DATA RETENTION**

Image sensors shall have an identifying number traceable to a test data file. Test data shall be kept for a period of 2 years after date of delivery.

### MECHANICAL

The device assembly drawing is provided as a reference.

ON Semiconductor reserves the right to change any information contained herein without notice. All information furnished by ON Semiconductor is believed to be accurate.

# Life Support Applications Policy

ON Semiconductor image sensors are not authorized for and should not be used within Life Support Systems without the specific written consent of ON Semiconductor.



# **Revision Changes**

Revision Number	Description of Changes
1.0	Initial Release
1.1	<ul> <li>Correction to Figure "Number of LVDS pairs (ports) used vs. bit depth"</li> </ul>
1.2	<ul> <li>Correction to Figure "Number of LVDS pairs (ports) used vs. bit depth"</li> <li>Update to the responsivity model and some specification values</li> <li>Improved low temperature operating temperature specification</li> <li>Update of register summary table</li> </ul>
1.3	<ul> <li>Revised Clk In spec to be 45MHz -&gt; 50 MHz</li> <li>Updated branding</li> </ul>

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